RADIO'S LIVEST MAGAZINE

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Radio-Craft

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NO. 145

to Build
Ultra-Modern
-Tube Superheterodyne

See Page 202

Reactivating Oxide-Filament Tubes—A Modern Reflex Radio Set
Tuning Meters—A Portable S.-W. Test Lab.—Tube Adapters

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FOR THE SERVICE MAN · DEALER · RADIOTRICIAN

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ONE-TUBE REFLEX LOUDSPEAKER SET. High-gain potentialities existent in the new tube types become dynamic in a novel receiver design that bids fair to attract widespread attention.

CHURCH SOUND INSTALLATIONS. Technicians today must be prepared to successfully handle a sound installation in any one of the 200,000 churches in the U.S. The problems in this profitable field are discussed.

THE SPECIALTY TESTER. Service Men long have wanted this instrument, which solves the problem, "What was the resistance of that burned-out resistor?"; also, it will indicate capacity values from 250 mmf. to 14 mf.

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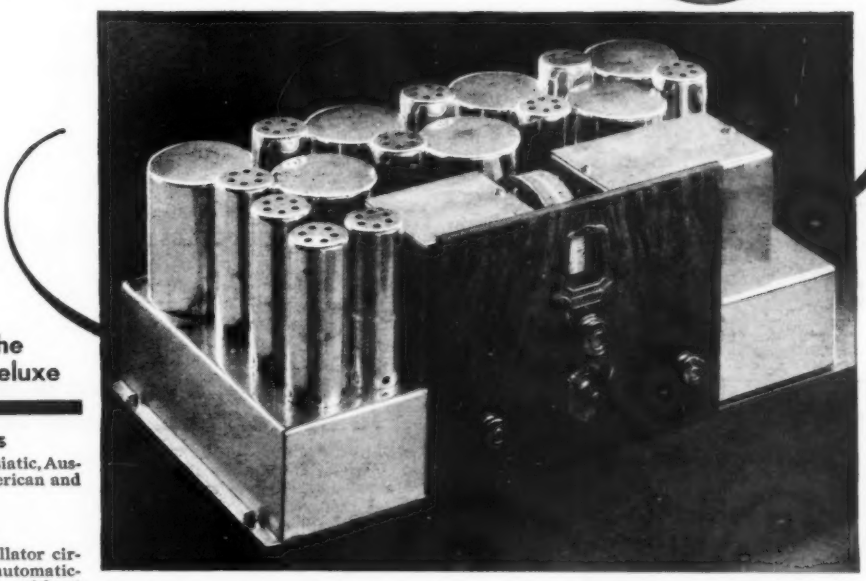

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Includes major European, Asiatic, Australian, African, South American and Domestic voice stations.

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Radio—the Field With a Future

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Dept. 2KX, National Radio Institute
Washington, D. C.

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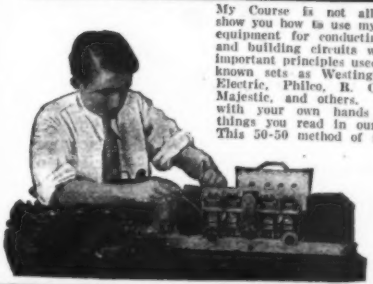

My Course is not all theory. I'll show you how to use my special Radio equipment for conducting experiments and building circuits which illustrate important principles used in such well-known sets as Westinghouse, General Electric, Philco, R. C. A., Victor, Majestic, and others. You work out with your own hands many of the things you read in our lesson books. This 50-50 method of training makes learning at home easy, interesting, fascinating, intensely practical.

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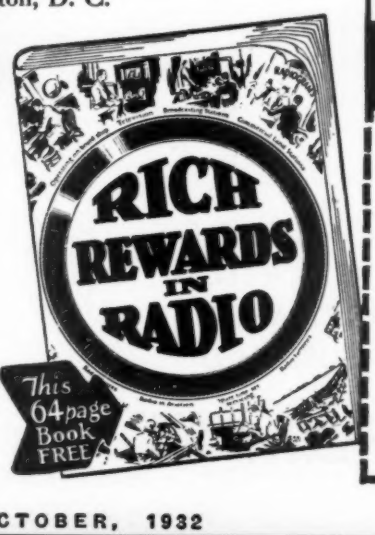

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OF NEW DATA HAVE BEEN ADDED TO THE
OFFICIAL REFRIGERATION SERVICE MANUAL

To bring this useful service manual right up-to-date, Mr. L. K. Wright, the Editor of the OFFICIAL REFRIGERATION SERVICE MANUAL, has added a wealth of entirely new material on electric refrigerators. This new section, which totals 96 pages, has contained therein dozens of new models of recent manufacture which have been placed on the market during the past few months and in addition, information regarding improvements on older models. As usual every refrigerator has been accurately described from the viewpoint of servicing—diagrams to illustrate the essential parts, and simplified so that repairs can easily be made.

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THE idea of electricians, radio service men and other mechanically inclined men, servicing refrigeration units is self-evident and the thought has occurred to perhaps untold thousands ever since electric refrigeration started. Yet nothing was done, because the average man knows little or nothing about refrigeration. Compared with servicing a radio set or wiring a home for electricity, the servicing of a refrigerator is absurdly simple, once you get the hang of it.

The OFFICIAL REFRIGERATION SERVICE MANUAL has been edited by L. K. Wright, who is an expert and a leading refrigeration authority. He is a member of the American Society of Mechanical Engineers, American Society of Refrigeration Engineers, The National Association of Practical Engineers, etc.

In this Refrigeration Manual every page is profusely illustrated; every refrigerator part is carefully explained; diagrams are furnished of every known machine; special care is given to the servicing end. The tools needed are illustrated and explained; there are trouble shooting charts, and other service data.

Remember there is big money in the refrigeration servicing business. There are thousands of firms selling refrigerators every day and they need to be cared for often. Eventually there will be more refrigerators than radios. Why not increase your earnings with a full or spare time business by servicing refrigerators.

Here are some of the important chapters:

Introduction to the Refrigeration Servicing Business
History of Refrigeration
Fundamentals of Refrigeration
Description of All Known Types of Refrigeration.
Service Tools and Shop Equipment
Motors
Trouble Shooting
Unit Parts, Valves and Automatic Equipment
Makes and Specifications of Units
Manufacturers of Cabinets
Refrigerants and Automatic Equipment
and Many other Important Chapters.

Several thousands of copies of the OFFICIAL REFRIGERATION SERVICE MANUAL have been sold; and there still remains the greatest opportunity for thousands more to learn how to make more money in a short time through openings in this new field.

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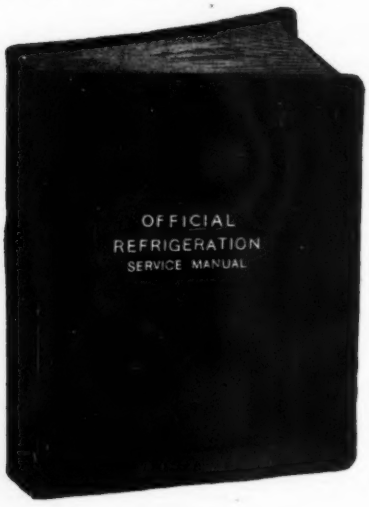
Complete Service Data



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WHAT a radio! One complete 16-tube chassis with one dual-ratio dial—new Super-Heterodyne circuit with a range of 15 to 550 meters . . . . STAT-OMIT Tuning Silencer . . . . New Class "B" Push-Push Power Amplifier . . . . Color-Lite Tuning . . . . Full band automatic Volume Control . . . . Duplex Duo-Diode Detection . . . . Dual-Ratio Single Dial . . . . **No** Trimmers, No Plug-in Coils, No Tuning meter or Neon light required . . . . Fractional Microvolt Sensitivity . . . . Dual Powered (2 separate power Transformers) . . . . Absolute Tone Fidelity . . . . 18 Tuned Circuits . . . . New Mercury Rectifier . . . . Full-Floating Variable Condenser . . . . Low Operating Cost . . . . and many other sensational new features. The new Midwest 16-tube set actually uses less current than previous sets of 8 and 10 tubes. A bigger, better, more powerful, more selective, finer toned radio than you've ever seen before . . . . offered at an amazingly low price direct from the big Midwest factory. Mail the coupon or send name and address on a postal for all the facts.

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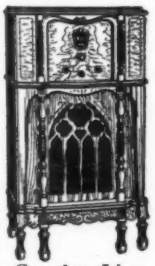

Complete Line of New Consoles

The big new Midwest catalog shows gorgeous line of artistic consoles in the new six-leg designs. Mail the coupon now. Get all the facts. Learn how you can save 30% to 50% on a big powerful radio by ordering direct from the factory.

Deal Direct With Factory!

Don't be satisfied with less than a Midwest 16-tube A. C. radio. A receiver covering only the regular broadcast waves is *only half a set.* Improvements in short-wave programs have made ordinary broadcast sets obsolete. The Midwest gives you regular, foreign, police and amateur broadcasts in one single dial set. No converter or any extra units required.

Remember, you buy DIRECT FROM THE MAKERS. No middlemen's profits to pay. You get an absolute guarantee of satisfaction or money back. You try any Midwest 30 DAYS before you decide to keep it. Then, if you wish, you can pay in small monthly amounts that you'll scarcely miss. Mail coupon for full details or write us a postal.

SAVE UP TO 50%

TERMS as low as $5.00 DOWN

Remember!

Every Midwest set is backed by a positive guarantee of satisfaction or your money back. 30 DAYS FREE TRIAL in your own home makes you the sole judge. Midwest, now in its twelfth successful year, offers bigger, better, more powerful, more sensitive radios at lower prices than ever before. The coupon or a postal brings you big new catalog and complete information. Mail it NOW!

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Using the New AIR CELL BATTERY

Two sensational values: a 6-tube super-het for standard-wave reception and a 9-tube ALL-WAVE, both using the new AIR CELL "A" battery that never needs recharging. Low factory prices. Coupon brings details. Mail it NOW!

Read These Letters From Midwest Owners

Just two of the *thousands* of letters praising Midwest Radios.

Gets France, Spain, Italy, Japan
"Have received foreign short-wave stations such as FYA, France; EAQ, Madrid, Spain; 12 RO, Rome, Italy; and last but not least, JIAA, Tokyo, Japan. I really think the Midwest set is a miracle."
A. F. GRIDLEY, Sarasota, Fla.

W8XK—W3XAL—W1XAZ—W2XAF
I am very much satisfied in every way with my Midwest radio. I heard Sydney, Sunday 3.00 A. M. also W8XK; W3XAL; W1XAZ; W2XAF; in the evening. On the regular band have some 55 stations so far.
Aug. Balbi, 1427 Myra Ave., Los Angeles, Calif.

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Get the Midwest catalog. Learn the facts about Midwest 9, 12 and 16-tube ALL-WAVE sets. Learn about our sensationally low factory prices, easy payment plan and positive guarantee of satisfaction or money back. Don't buy any radio until you get the big new Midwest catalog. Just sign and mail the coupon, or write us a postal.

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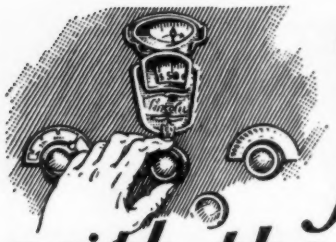
MIDWEST RADIO CORP.

DEPT. 111 (Est. 1920) CINCINNATI, OHIO



Tune the NEW way *for* World-Wide Reception *with the* LINCOLN DE·LUXE·SW 33

SILENT TUNING···

The weakest carrier wave is registered on the signal indicator and can be tuned with precision and perfect silence without disturbing atmospheric noises.

SIGNAL INDICATOR

A meter directly above the dial indicates, not only the weakest signal, but allows the operator to tune into a signal perfectly. Guess work is entirely eliminated. Comparative signal strength is indicated.

UNDISTORTED HIGH AMPLIFICATION·

Three stages of push pull with new system of twin-grid detection allows tremendous undistorted amplification of the high gain I. F. amplifier. The handling power of this system seems to be unlimited and tremendous volume on weak signals can be had if desired.

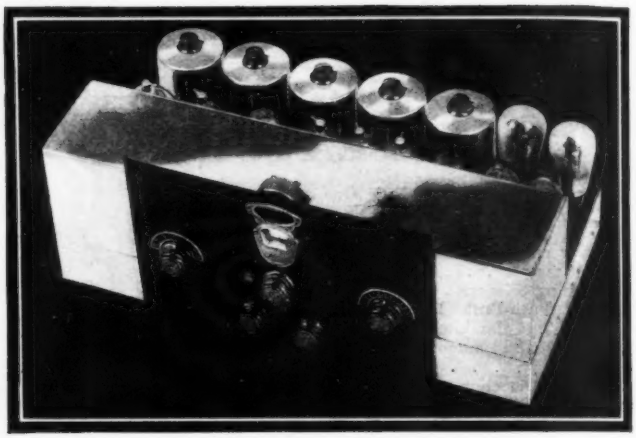
AUTOMATIC VOLUME CONTROL·····

There are two paramount advantages in good automatic volume level. First, in tuning from weak to strong signals; and secondly, in holding a steady volume level on fading stations which is so common in short wave reception. The effect of this new Lincoln feature is so efficient that a near-by stroke of lightning registers only a muffled sound in the speaker; it has the same effect on all sharp electrical interference.

NEW FIDELITY

Twin-grid detection preceded by push pull input I. F. transformers and followed by two stages of transformer coupled push pull stages, produces an undistorted register of a wide band of frequencies, giving a perfectly balanced output with realism hard to associate with radio.

All of the new reactions in the SW-33 model are what we all have wanted for years,—they are here for you today—thanks to Lincoln's foresight in radio possibilities.



THE NEW DEVELOPMENTS

have made the new DeLuxe SW-33 just about as ideal a receiver as one could hope to own. The use of five variable mu tubes controlled by the new twin grid second detector and followed by two transformer coupled push pull stages has opened the gates to new ideas of enjoyable distant reception.

The signal indicator locates carrier waves which are difficult to hear; many times the carrier is not being used or modulated as is the case in transatlantic phone. The signal indicator registers these silent carriers and enables you to be accurately tuned, ready for the voice to be heard.

WHEN THIS CARRIER IS TUNED, ATMOSPHERIC NOISES ARE REDUCED TO A MINIMUM, AUTOMATICALLY.

Distant stations can be tuned silently, and volume then brought up to desired strength (volume control does not affect sensitivity). Perfect volume level on short wave stations is another great asset in the new Lincoln. If you have ever tuned in a foreign short wave station, or even many of our short wave stations in the U.S.A., you will appreciate the great value of uniform volume level.

The performance of Lincoln equipment has been known the world over for years. Its use by Polar Expeditions, broadcasting stations, both domestic and abroad, U. S. Naval Station operators, and hundreds of super critical DX fans, has proved Lincoln's exceptional merit.

Complete equipment consists of chassis, power equipment, auditorium type speaker and complete set of laboratory tested tubes. Chassis is finished in highly polished nickel over copper and presents a handsome appearance. Precision laboratory construction is employed throughout, and every receiver is tested on distance before shipment.

Write for description of new developments and new sales plan which overcome the present defects in the present custom built radio merchandising.

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Radio-Craft
FOR THE SERVICE MAN · DEALER · RADIOTRICIAN
"Takes the Resistance Out of Radio"

Editorial Offices: 96-98 Park Place, New York, N. Y. HUGO GERNSBACK, Editor Vol. IV, No. 4, October, 1932.

RADIO IN EUROPE

An Editorial by HUGO GERNSBACK

FOR the purpose of making an intensive study of radio and television conditions as they now prevail in the foremost European countries, I recently visited Germany, France, England and Belgium.

Every facility was placed at my disposition, and in the course of my travels I visited every important radio and television laboratory on the continent. I also had numerous occasions to personally operate a number of radio receivers in all the countries visited.

In my opinion, the present radio conditions abroad may be best summed up in a single word—"Deplorable." The general quality of the European programs is, as a rule, mediocre. There are good features, and in some cases excellent ones, to be sure, but they are few and far between. What music there is, as a rule, is good, but European broadcasts suffer from entirely too many talks. Surprising as it may seem, there is no such thing as a continuous program from early morning (5.45 A.M.) until after midnight (1 A.M.)—even by large stations—such as American listeners are accustomed to. Even the big stations go off the air for hours at a time. There is also not the snap and go to the average European stations that we are accustomed to in America. They take things easy, and a few minutes wait—anywhere from 3 to 10 minutes—between programs means nothing to the European broadcaster.

But when it comes to radio reception in Europe, the conditions—to the American—are not only deplorable but chaotic. It is practically impossible to listen to a radio program in the large cities, such as in Berlin, Paris, or London. The amount of man-made interference is simply terrific, and at present no steps seem to have been taken to remedy this sad situation. Time and again I have tried to listen to programs in these various cities, but found it absolutely impossible to do so, particularly at night; but even in the daytime, listening often became impossible. In the country districts, conditions are somewhat better so far as man-made static is concerned, but all to no avail, because of the present European radio chaos.

In the United States we have the Federal Radio Commission, which polices the various wavelengths so efficiently that broadcasters must remain on their allotted frequency. In Europe, there is no central authority to enforce this prime radio rule, that is, that radio stations must stay on their assigned wavelength. Berne (Switzerland) is supposed to regulate the stations, but in practice, broadcasters do much as they please.

The result is that there is a terrific amount of interference between the stations where wavelengths overlap. This naturally gives rise to heterodyne whistles, and only rarely is a program free from such annoyances. Of course, we have heterodyne whistles on some of our stations in the United States, too, but generally these stations are in what is termed the "graveyard" wavelength, below 250 meters. In Europe, there are heterodyne whistles on the best frequencies, and it is seldom that you can select a wavelength that is free from this annoyance. This, of course, accounts for the fact that the average receiver in Europe consists of but three tubes. When you talk to Europeans of an eight, ten or twelve tube receiver, they cannot conceive of such a thing. Indeed, if you had such a receiver in Europe it would be useless. The amount of interference that the higher sensitivity would bring in would make listening quite impossible. I listened to a few six and eight tube superheterodynes in some radio laboratories, and it was absolutely impossible to keep the set on for more than a few minutes as the din and racket was so nerve racking that it became positively hopeless to listen to it.

Indeed, as a general thing I found that Europeans who had been in the United States and who had owned or used radio sets here, do not own radios in Europe now. The explanation usually was that after being accustomed to American radio reception, they found themselves unable to listen to European programs with their eternal talks, their accompanying noises, and preferred to be without radios rather than listen to the assorted din.

I have never been a friend of our American radio practice whereby programs are sponsored by advertisers, and where advertising blurbs clutter the air all too frequently. Of course, in Europe there is little advertising on the air; practically all broadcasting stations being owned by the various governments. However, after listening night after night, for three weeks, to the European radio pandemonium, I am certain that American radio listeners, if they had been treated to a like experience, would forget their present antagonistic attitude toward our radio programs and come to the conclusion—as I have—that after all is said and done, American radio, today, is so far above that of Europe, that a comparison becomes silly.

As I have mentioned before, the average European radio set is composed of three tubes. In fully 80 percent of the sets, regeneration by the tickler method is still being used. This means that the sets produce the same sort of howls and cat music that we had to contend with up to about 1928. Of course, since that time, we have not had sets that whistle and howl, but in Europe, such sets are still the usual thing. The condition is aggravated also by the fact that all European sets must tune up to the higher wavelengths of over 2,000 meters, as some of the best stations are broadcasting on these frequencies. This makes elaborate switching arrangements necessary, which switches have not been worked out very happily in Europe so far.

Speaking broadly, it may be said that a great majority of the sets are of the regenerative and tuned radio-frequency type. Only very recently have superheterodynes been introduced, but so far, only a very small percentage of the receivers sold are what might be termed modern sets, such as we know in the United States.

As to the general appearance of the sets, they are of the 1928 vintage, according to American standards, and they are practically all of the table-model type. While it is possible that someone is building consoles in Europe—I saw hundreds of sets and operated dozens of them—I never glimpsed at a single console during my whole stay in Europe. All the sets I saw, with but one exception, had separate loudspeakers, such as were in vogue in the United States in 1926-28. In Europe, the radio set is still just a radio set, and no one seems to know what radio furniture is, such as we know the term here.

Summing up, it may be said that radio in Europe today is just about five years behind us, and it has been so ever since it started abroad. It is also quite likely to remain so, due to present economic conditions, and the slower adaptability of Europeans to new ideas.

BUILDING AND OPERATING AN ULTRA-MODERN SUPER.

By CLIFFORD E. DENTON

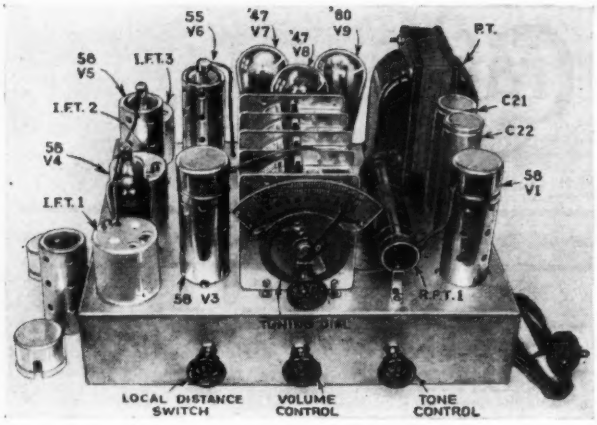


Fig. A
Front view of the completed receiver. The designations refer to the schematic circuit below.

WHILE there are many good receivers available today, at the suggestion of RADIO-CRAFT, the author constructed a 9-tube receiver of the superheterodyne type, utilizing all the new tubes that would be practically beneficial; a band-pass antenna system, to provide sufficient selectivity for those located in metropolitan areas; automatic volume control; and a diode detector. The sensitivity of this receiver is in the neighborhood of 1 microvolt-per-meter, which should more than satisfy the most ardent DX fan, and at the same time enable distant stations to be received with ease.

If the instructions given in this article are followed, then the entire set may be built for about twenty dollars, assuming nothing in the "junk box" but a few tools.

The Circuit

A study of the circuit diagram of Fig. 1 reveals that a total of 9 tubes are used, including the rectifier. Starting at the antenna, a band-pass tuning unit feeds the signal into the receiver proper. The output of the first R.F. tube is loaded by a choke, L4, which resonates at a frequency just below the broadcast band. Coupling to the first detector, or modulator tube, is made via a small, mica fixed condenser, C24. With this resonant-choke arrangement, maximum output is secured from the R.F. stage, V1, over practically the **entire** broadcast band.

Oscillator tracking is accomplished by means of a series of padding condensers, C3A and C4 in the diagram. However, best results will not be secured unless care has been taken in the construction of the coils.

An interesting feature of this receiver is the fact that the oscillator is coupled to the first detector both conductively and inductively; conductively by means of the common resistor in the cathode leads of the oscillator and first-detector; and inductively through the physical location of the two sets of windings, that is, the oscillator and first-detector secondaries. Both of these coils are wound on the same tube and separated by about ½ inch.

Extremely high gain is obtained in the I.F. stages by virtue of the fact that both the primary and secondary of each I.F. transformer is tuned; and since there are three such transformers, there are six tuned circuits in the I.F. stages alone. In view of the fact that the receiver is equipped with A.V.C., the sensitivity of the various tuned stages varies with signal strength. This is so in the first R.F. and first I.F. stages; the second I.F. stage is not in the A.V.C. connection and consequently works at high sensitivity at all times. A local-distance switch is used, in series with a resistor, R14, to decrease the gain of the second I.F. stage.

The new Duplex-Diode Triode is used as the second detector. (See description of this tube in the Sept. issue of RADIO-CRAFT. —*Editor.*) This tube, the 55, is fed from the I.F. secondary, and has both plates connected together in a half-wave connection; a center tapped I.F. transformer must be used for full-wave detection, and since such transformers are not as yet available, the half-wave connection of this tube was used. This is also desirable since a greater voltage is available for A.V.C. action, also secured from this same tube. Hence, this tube is a combination diode detector, A.V.C. tube, and since it is equipped with an additional grid and plate, it also acts as an audio amplifier. An especially interesting feature is the fact that the diode portion of the tube is resistance-coupled into the triode portion, and the latter transformer-coupled into the push-pull output stage.

A tone control, consisting of C20 and R13 is shunted across the secondary of the push-pull transformer to facilitate the accentuation of the lower register, if so desired.

The Automatic Volume Control

Automatic volume control is obtained in a novel manner. Tube V6, the Duplex-Diode Triode, supplies the voltage. Note that the cathodes of both the first R.F. and first I.F. tubes return, through grid bias resistors R2 and R8, directly to the cathode of V6. Therefore, the bias voltage appearing across R11 supplies the triode portion of V6. The voltage across R10 supplies the potential for the tubes oper-

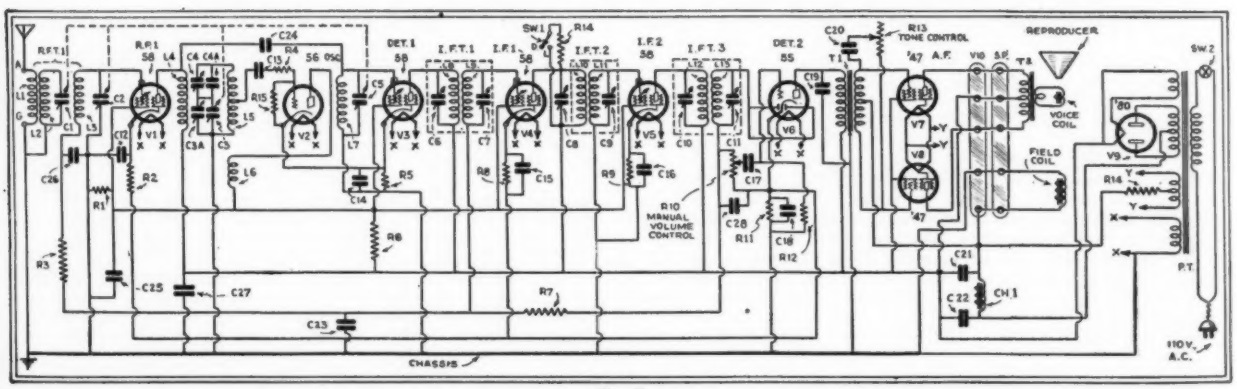


Fig. 1
Complete schematic circuit of the receiver. Note the use of the new 55 in the second-detector circuit.

RADIO-CRAFT always attempts to bring before its readers the latest devices and the latest application of these devices. To accomplish this, we have prevailed upon Mr. Denton to construct the receiver described here. This receiver uses, as may be seen, type 58 tubes in the R.F. and I.F. stages; the new type 55 Duplex-Diode Triode as combination second detector, first A.F., and A.V.C. tube; and two pentodes in a push-pull connection.

A single evening's test convinced us of the tremendous sensitivity and selectivity of the set. Between the hours of 9 and 10 P.M., over 50 stations were logged, and without any interference of any sort. Local stations were tuned out in a half a degree on the dial, and a distant station tuned in on the next channel!

We believe that this is the most up-to-date receiver ever described that may be constructed at home.

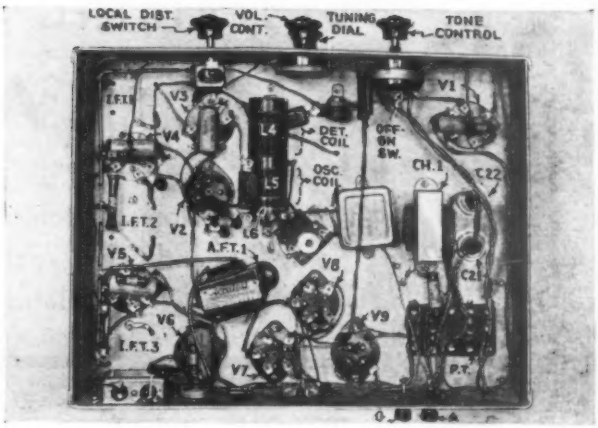


Fig. B
Under-chassis view of the receiver. The simplicity is evident.

ating under the influence of A.V.C. Manual control of volume is obtained by varying the arm of the potentiometer R10.

The time of delay of the A.V.C. action is about .1-second, the exact amount depending upon the accuracy of the time-delay circuit composed of R7 and C23. Additional isolation in the form of R3 and C26 decouples the first R.F. stage from the remainder of the set.

Coil Data

Full constructional details of the coils are given in Fig. 2. Note that the antenna coil in Fig. 2A is wound in honeycomb fashion at one end of the tubing and is held in place by means of collodion and small pieces of adhesive tape. The three coils on this form are all wound in the same direction.

The first-detector and oscillator coils are wound on a second form as shown in Fig. 2B, a half-inch separating the two coils. Care must be taken to place the plate winding of the oscillator *on* that end of the grid coil nearest the end of the tubing, as shown. The list of parts at the end of this article gives the name of the manufacturer of the coils, in the event the constructor desires to purchase them.

Winding data for the I.F. transformers are given in Fig. 3. Each of the windings (both primary and secondary) consist of 800 turns of No. 36 D.S.C. wire on a wooden core one-half inch in diameter. Further spacing details may be secured from Fig. 3.

Construction

The set is simple to construct although there are nine tubes employed; simply use care in mounting and soldering all electrical connections. Chassis specifications are given in Fig. 4, but must be changed if material other than that specified is used. A cadmium-plated chassis may be purchased fully drilled and folded in accordance with the above specifications. Made of steel and with welded joints, it presents an attractive appearance.

Mount the sockets, band-pass coil, electrolytic condensers, four-gang tuning condenser, and the power transformer on top of the chassis as shown in the photographs in Figs. A and B. Place all tube shield-bases on their respective sockets, and after this is done, mount the intermediate-frequency transformers. Be sure that the third I.F.T. has its control-grid lead running *down through the chassis to the plates* of the 55 tube. All other material is mounted under the chassis. The fixed condensers and the resistors are supported solely by their leads; the two 1. mf. bypass condensers and the push-pull transformer are bolted down.

(*Continued on page* 237)

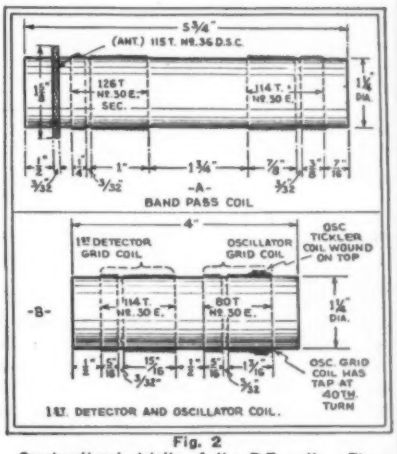


Fig. 2
Constructional details of the R.F. coils. The band-pass unit is shown at A, and the detector-oscillator unit at B.

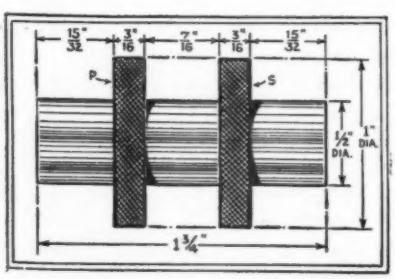


Fig. 3
Winding details of the I.F. transformers.

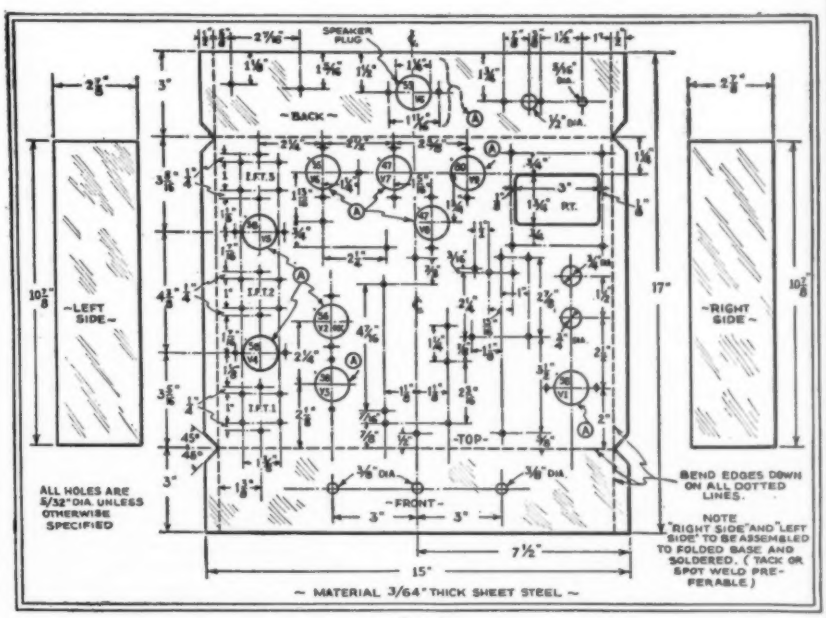

SPEAKER PLUG
-BACK-
-LEFT SIDE-
-RIGHT SIDE-
-TOP-
-FRONT-
ALL HOLES ARE 5/32" DIA. UNLESS OTHERWISE SPECIFIED
BEND EDGES DOWN ON ALL DOTTED LINES.
NOTE "RIGHT SIDE" AND "LEFT SIDE" TO BE ASSEMBLED TO FOLDED BASE AND SOLDERED. (TACK OR SPOT WELD PREFERABLE)
15"
7½"
17"
- MATERIAL 3/64" THICK SHEET STEEL -

Fig. 4
Complete chassis layout. The location of the parts is indicated by their respective cut-outs.

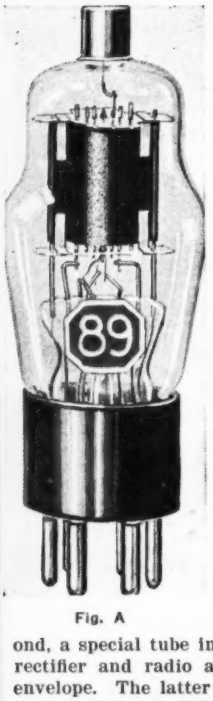


Fig. A

DESCRIBING
THE LATEST TUBES

A new universal automotive output tube and a special combination amplifier and power rectifier complete this month's description of the latest tubes used in radio work.

By LOUIS MARTIN

THIS month there are two tubes that we will describe. The first of these is of the automotive type; the second, a special tube incorporating a power rectifier and radio amplifier in a single envelope. The latter tube, while it is not commercially available at this time, indicates quite clearly what the trend is.

The 89

The 89 is a triple-grid power amplifier tube of the heater-cathode type recommended especially for use in automobile receivers or for other types of mobile service. The triple-grid construction of this tube, with external connections for each grid, makes possible its application as (1) a class A power-amplifier triode, (2) a class A poweroutput pentode, and (3) a class B power output triode.

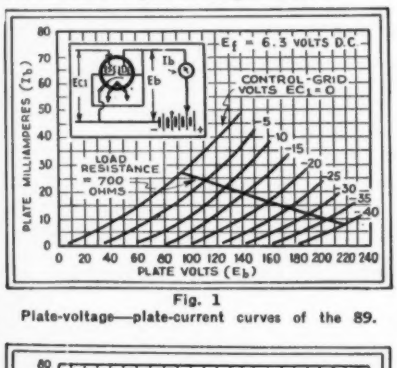


Fig. 1
Plate-voltage—plate-current curves of the 89.

The three-fold application of the 89, shown in Fig. A, to audio power-amplifier circuits is made possible by different connections of the three grids incorporated in the tube's structure. Thus, one arrangement of grid electrodes provides a triode for class A service with a low amplification factor, a low plate resistance, and a high mutual conductance; while another provides a triode with an amplification factor so high that negative grid bias is not required for its operation as a class B amplifier. A pair of 89's so connected in a class B output stage is capable of supplying a large amount of power with relatively low plate voltage and with unusual overall economy of power consumption. A third arrangement of the grids makes possible the use of the 89 as a class A power output pentode capable of giving large power output with relatively small signal voltage input.

In class A service, the grid of the tube is maintained negative with respect to the cathode by an amount such that some plate current flows at all times, and such that the grid takes no appreciable current during the most positive swing of the signal voltage. These operating conditions are obtained when the normal bias *without* signal gives sufficient operating plate current to permit the application of a peak signal equal to twice the bias value without reducing the plate current below a certain predetermined minimum value under the load conditions employed, or without swinging the grid positive. Thus, the value of grid signal voltage which can be applied to any given type of tube is limited and this results in limited power output. *Theoretically*, the maximum plate circuit efficiency for class A operation is 50 percent, assuming a sine wave input signal. The *actual* plate circuit efficiencies, however, are of the order of 20 percent for triodes and 40 percent for pentodes.

Characteristics of the 89

Figure 1 shows a family of plate voltage—plate-current curves of this tube when operated as a triode, that is, when the suppressor and screen-grids are connected directly to the plate. The applied signal is then fed between the control grid and the cathode in the conventional manner. In Fig. 2 is shown class A operation when this tube is used as a pentode, that is, with the suppressor or third grid connected directly to the cathode and the screen-grid to some positive voltage higher than the cathode.

In class B service, the tube is operated so that the plate current is practically zero with no grid excitation. When a signal of sufficient magnitude is applied to the grid, there will be no plate current flow over a substantial part of the negative half-cycle. In other words, plate current flows only during the least negative excursions of the signal voltage. A considerable amount of second and even higher order harmonic distortion is thus introduced into the power output of a single tube. However, with two tubes in a balanced push-pull circuit, the even harmonics are eliminated from the power

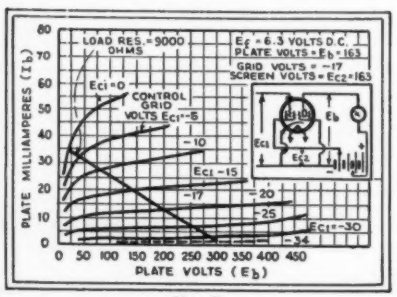

LOAD RES. = 9000 OHMS
Ef = 6.3 VOLTS D.C.
PLATE VOLTS = Eb = 163
GRID VOLTS = −17
SCREEN VOLTS = Ec2 = 163
CONTROL GRID VOLTS Ec1 = 0
PLATE MILLIAMPERES (Ib)
PLATE VOLTS (Eb)

Fig. 2
Curves illustrating pentode, class A operation.

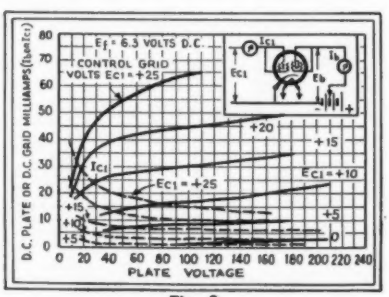


Fig. 3
Curves of the 89 as a class B amplifier.

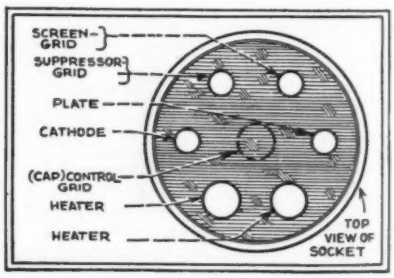


Fig. 5
Socket connections of the new 89.

output. In such a circuit, therefore, two tubes may be employed as class B amplifiers to supply virtually undistorted output.

In class B service it is possible to drive the grids of the two amplifier tubes positive to a certain amount and still obtain reasonably undistorted output, provided that sufficient input power is available to supply the grid current required by the positive grids. This power is conveniently supplied by a class A power amplifier feeding the grids of the output tubes through a push-pull transformer having proper characteristics. Usually this transformer has a step-down ratio.

By designing class B amplifier tubes with a sufficiently high mu-factor, it is possible to operate them with zero grid bias, and so dispense with biasing resistors whose effect would be to produce considerable loss in sensitivity because of degenerative effects. Since provision for grid bias is unnecessary with such tubes, the entire voltage of the rectifier is available for plate supply.

Distinguishing features of this class of service are that very high output of good quality may be obtained with fairly small tubes operating at relatively low plate voltage; and that unusual overall economy of power consumption is possible because the plate current is very low when no signal is applied to the grid. To give these advantages, the class B amplifier circuit requires the use of two tubes in a balanced output stage preceded by a driver stage capable of supplying considerable undistorted power and the use of a power supply capable of maintaining good voltage regulation regardless of the variation of average plate current with signal intensity. It should be noted that the distortion present in the high power output of class B amplifiers is usually negligible, but is always somewhat higher for the ordinary range of signals than that obtained with class A amplifiers employing much larger tubes capable of operating at the same maximum power output.

Figure 3 shows a family of plate-voltage—plate-current curves of the tube when operated as a class B amplifier. In the same curve, in dotted lines, the grid-current characteristics are also shown. Class B connection may be secured by connecting the suppressor grid directly to the plate and the screen directly to the control grid. The input signal is then applied in a conventional manner as in all triode tubes. An inspection of the curve will show that high grid-current flows when the plate voltage is low.

Figure 4 is an interesting set of curves inasmuch as they indicate how the power output and plate current varies when two type 89 tubes are connected for class B operation, which in turn are driven by a single 89 in a class A connection. At A, the transformer ratio between the class A driver and class B stage is 2.4; the output load impedance from plate to plate being 9,400 ohms. In B of the figure, the input transformer ratio is 3.7 and the output impedance from plate to plate is 13,600 ohms. It will be seen that in the first instance, the total harmonic distortion is about 8 percent and at B, the harmonic distortion is about 5 percent for maximum signal input.

The base of the 89 is of the small 6-pin type as shown in Fig. 5. Its pins fit the standard six-contact socket which may be installed to operate the tube either in a vertical or in a horizontal position. For horizontal operation, the socket should be positioned with one filament pin opening vertically above the other.

The D.C. resistance in the grid circuit of the 89 operating as a class A amplifier (either with triode or pentode connection) should not exceed 1.0 megohm if self-bias is used. Without self-bias, the resistance should not exceed 0.5 megohm. The use of resistances higher than these may cause the tube to lose bias due to the grid current with the result that the plate current will rise to a value sufficiently high to damage the tube.

The direct-current requirements of class B circuits are subject to fluctuation under operating conditions. The power supply, therefore, should have as good regulation as possible to maintain proper operating voltages regardless of the current drain. For this purpose, a high voltage "B" battery or a suitably designed "B" eliminator may be employed. If the latter is used, a rectifier tube of the mer-

(*Continued on page* 235)

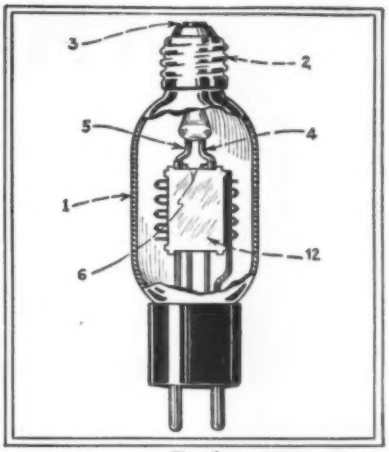


Fig. 6
External view of the amplifier-rectifier.

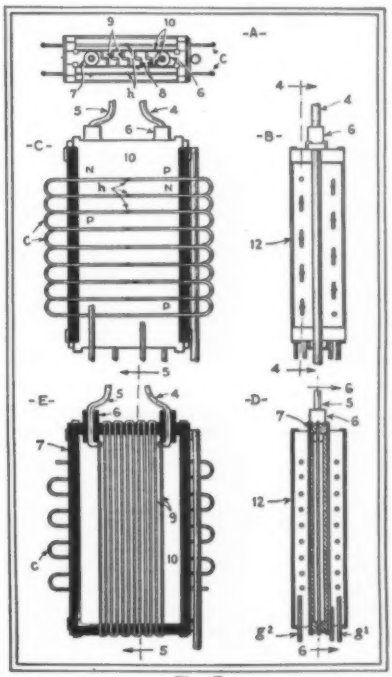


Fig. 7
Details of the internal construction of the tube.

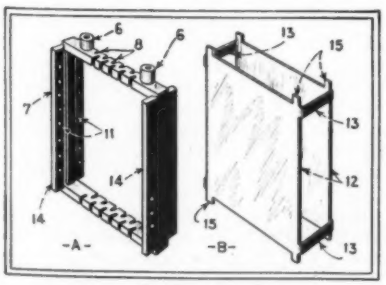
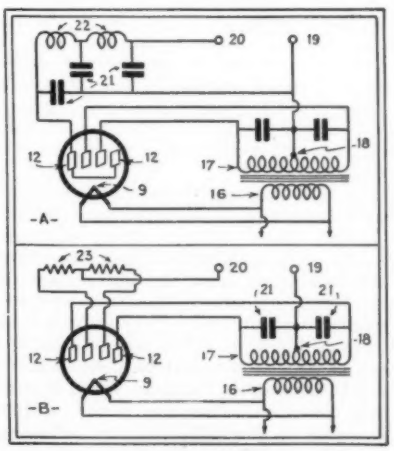


Fig. 8, above.
Details of the grid support and plates.

Fig. 9, right.
Schematic circuit showing how the tubes may be used in a power rectifier circuit. With the proper connection of the grids, which are self biasing, the tube may be connected to a source of A.C. and operated directly without a separate rectifier tube.



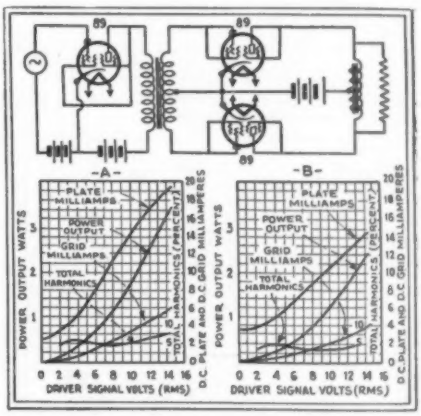


Fig. 4
Variation of power output and plate current.

THE NEW "K" TUBE

The efficient use of relays demands a vacuum-tube that has been designed specifically for such purposes. The author describes such a tube in the article below.

By C. BRADNER BROWN

Fig. A
The "K" tube, designed especially for relay work.

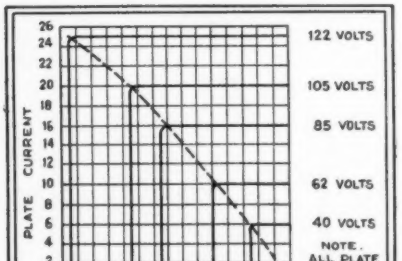


Fig. 1
A family of grid-voltage—plate-current curves.

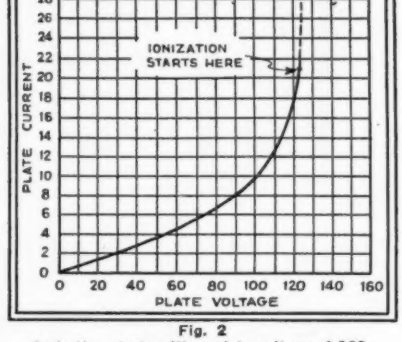


Fig. 2
Ionization starts with a plate voltage of 120.

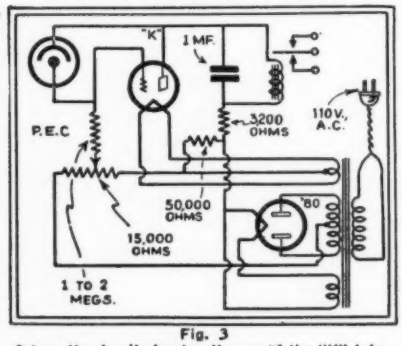


Fig. 3
Schematic circuit showing the use of the "K" tube.

THE new "K" tube was developed in response to the need for a vacuum tube which would have more favorable characteristics for relay operation. The author having done considerable experimenting with photo-cell amplifier and associated relay equipment has come to the conclusion that the present high-vacuum tubes are not eminently suitable for relay operation. The main reason for this is found in the inherent action of a high-vacuum tube. When the grid voltage is changed, the plate current is changed, the amount depending on the change in grid voltage. This change in plate current, however, follows a smooth curve and therefore a point will be found at which the relay is pulled in with not quite enough force to operate the armature. This point is called the "threshold" of operation and when held at this current, the relay will chatter, resulting in the burning or pitting of the contacts. Even when operated at the point of grid voltage cut-off, this condition will be found unless a considerable grid shift is used to operate the tube.

The thyratron tube offers a much better characteristic for relay operation as the action of the grid is to turn the plate current on—more in the manner of a snap switch. The type "K" tube is an adaptation of the thyratron tube using neon gas instead of the customary mercury vapor. The conducting gas is the secret of its peculiar characteristics which will be discussed later. It was constructed using the elements of the old type CG1162, because this tube has a circular construction using a cylindrical plate which considerably simplifies the calculation of electrostatic field strengths and surface densities. The operation of the gas filled three-element tube is as follows:

The electrons emitted from the filament are accelerated toward the plate by the positive plate voltage. These electrons collide with molecules of neon, knocking loose electrons from the outer rings of the atomic structure. These released electrons are also attracted to the plate and, in turn, break up other molecules of the gas increasing the electron flow until the gas is thoroughly ionized. When it is in this condition, there are plenty of free electrons in the tube, and the resistance is very low which will result in the flow of large plate currents. It is necessary to use a filament which will withstand the bombardment of the positive ions which are necessarily released to be attracted to the filament. The coiled tungsten filament used in the type "K" tube is admirably suited for this duty as it is very rugged and not affected noticeably by the positive ions.

If the grid is made sufficiently negative, the plate current will be reduced to zero as all electrons emitted by the filament will be returned to the filament by the negative grid. The tube will be inactive in this condition, but current will begin to flow again as soon as the grid voltage is reduced to the cut-off point. When the current starts to flow, the tube will ionize and its resistance reduce to almost zero. Once started however, the grid is powerless to stop the flow of current as the electrons *are being released by the gas itself and not by the filament.* The ionization can only be stopped by reducing the plate voltage to a value sufficient to stop ionization. If alternating or pulsating voltage is used on the plate, this condition is fulfilled at every cycle or alternation. The output of a full-wave rectifier particularly adapts itself to the operation of the type "K" tube and was used in the experimental relay hookups which were set up to test the operation of the tube.

The remarkable characteristics of the type "K" tube are shown in the form of a family of plate-voltage—plate-current curves in Fig. 1. The sharp cut-off action when the tube is operated with a negative grid bias is extremely valuable when the operation of a relay is to be considered. The corners are naturally rounded off slightly when used with rectified A.C., but this action is not noticeable in actual tests. It can be seen that a sudden increase of plate current from zero to some positive value takes place when the negative grid bias is reduced past a certain critical point. It is this "breakover" action which prevents the chattering of relay contacts and allows the use of much more insensitive equipment than normally is the case. Fig. 2 shows the ionization characteristics in the form of a plate-voltage—plate-current curve for zero grid bias. The current increase to above 200

(*Continued on page* 245)

A PORTABLE PHONO-RADIO RECEIVER-RECORDER

By DR. F. NOACK*

● The idea outlined in this description of a novel German development, introduced at Leipzig Spring Fair, may be applied in many ways.

AT the recent German Radio Exposition were shown by the German industry several sets for the home-manufacture of phonograph records. With only one exception, these were so designed that only an amateur radio enthusiast could manipulate them—since they consisted mainly of the components requisite to making such records, and were not designed for convenient operation by the layman.

This exception was the equipment introduced under the name Schacktograph; it is illustrated in Fig. A in finished form and is the instrument shown at the succeeding Leipzig Spring Fair.

The Schacktograph incorporates a radio receiver chassis which includes a 3-stage A.F. amplifier capable of delivering an output of 2.5 watts; the output of this section of the instrument feeds into a magnetic reproducer which also is part of the assembly; then, there is the turntable and its associated motor; finally, there is a microphone and a recording-reproducing head. (To meet the demands of German power systems the Schacktograph is designed to operate at 110, 127 or 220 volts, A.C.)

*Berlin Correspondent.

By means of a switching system it is possible to arrange the circuit for a number of services, to wit: (1) Disc recording of radio programs; (2) disc reproduction of radio programs; (3) disc recording of impromptu programs; (4) disc reproduction of impromptu programs; (5) radio program reproduction.

In this particular machine the cutting needle may be steel, diamond or sapphire. The magnetic reproducer may be conveniently disconnected from the circuit, and the output run into a dynamic reproducer, if desired; the field excitation current being supplied at terminals on the receiver. The magnetic unit built into the chassis is used not only to reproduce the radio programs picked up by the set, but also as a means of monitoring the audio output of the A.F. amplifier portion of the assembly, before or during the recording of impromptu programs (such as singing or talking by the operator or his friends, etc.).

A device of this nature achieves a versatility combined with utility that certainly places it in a class by itself. In this instance of bringing "the mountain to Mohammed," the only requisite is a 110-volt power line.

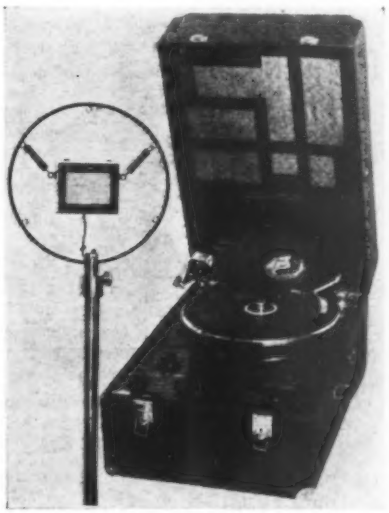


Fig. A. The Shacktograph receiver-recorder.

R. A. FESSENDEN DIES

● In the passing of Professor Reginald Aubrey Fessenden, who died July 22 at Hamilton, Bermuda, at the age of 65 years, the technical world mourns the loss of one of its most colorful pioneers.

Although of Colonial American ancestry—the family came to Cambridge, Mass., in 1628—Professor Fessenden was born in Milton, Province of Quebec, Canada.

His experimental work with radio dates from the early '90s, when he first investigated "Hertzian waves" and at that time held, against the opinion of other investigators, that these radiations, in their method of propagation, bore a direct relationship to light; and that they traveled as "continuous waves" and not as a "whiplash" action. He accomplished the radio transmission of speech by a commutator method in 1900, at Cobbs Point, Md.; in 1903 the distance had been increased to 25 miles through the use of the continuous arc. He instituted radio broadcast programs from Brant Rock, Mass., in 1906; a high-frequency alternator of his design made this practicable.

Professor Fessenden's inventions include the loop antenna, silicon iron for transformers, high frequency alternator, the heterodyne principle, automatic relay radio, induction telephony by means of a copper diaphragm and exciting coil, Empire cloth, submarine telephone, echo-type depth finder, iceberg and submarine locators, the tracer bullet, ultra-audio sound signalling, and three hundred others.

In the magazine series, "Men Who Made Radio,"* appeared a short biography and cover illustration, in colors, of Professor Fessenden.

A man of strong will, he fought to maintain his convictions and often against almost overwhelming odds. He created a furore in the commercial world when, in 1925, he propounded the following oft-quoted Law of Invention:

"No organization engaged in any specific field of work ever invents any important development in that field, or adopts any important development in that field until forced to do so by outside competition."

A man of high ethics, Professor Fessenden contributed greatly to placing modern invention on a plane by itself as a field worthy of the best efforts of an individual; in this connection he is quoted as follows: "Invention is an art; just as much so as painting, or doing fine machine work."

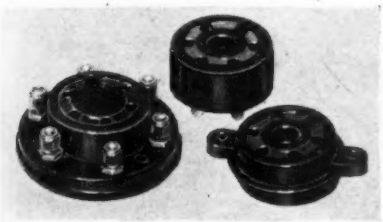
In Memorium, we again, humbly, quote Professor Fessenden: "We shall be remembered for a hundred things, our literature, our surgery, our medicine, our chemistry, our wars. But most of all we shall be remembered for our inventions."

* RADIO-CRAFT, January 1930, page 309.

THE LATEST RADIO EQUIPMENT

NEW NA-ALD SOCKETS

THREE sockets, all of the 6-prong type are now available for consumer use from the Alden Manufacturing Co. The type illustrated at the extreme left is for baseboard mounting; that in the upper center

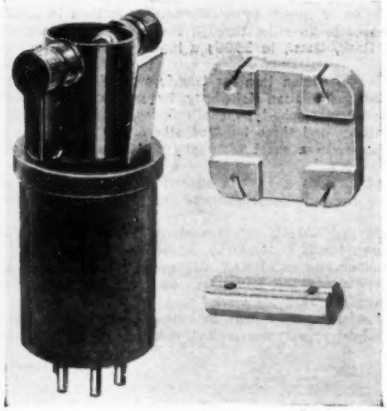


New Na-Ald sockets.

for subpanel mounting; and that at the extreme right has a special flange suitable for subpanel mounting with two screws—that shown in the center having but a single hole. They are known respectively as types 486, 426, and 436.

LYNCH ANTENNA KIT

THE September issue of this magazine contained an article on short-wave collectors. This article stressed certain types of construction for the antenna in order to efficiently receive short waves. Another recommendation was the use of an antenna coupler connected as shown in the article—Fig. 2 at C—and pictorially illustrated below. Reduction of noise pickup is accomplished by transposing the lead-in by means of transposition blocks constructed as shown in the article and pictured below. All of the above apparatus are now available from the Lynch Manufacturing Co.



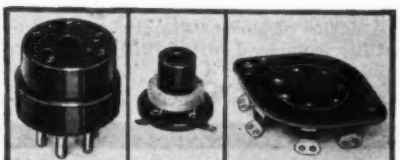
Lynch antenna kit.

RADIO CITY ADAPTERS

NEW tubes mean new adapters, if test equipment is to be kept up to date. Shown below, at the extreme left, is an example of a 5-hole to 5-prong adapter suitable for use in tube testers, analyzers, radio sets, in fact anywhere where tubes are used. Manufactured by the Radio City Products Co., they have been especially designed to accommodate the Service Man who must have the proper adapter at a low cost. They are made in all types and sizes for all the new tubes including the new 7-prong tubes which will be announced shortly.

R.F. CHOKE FOR THE 82

THE recently announced 82, mercury-vapor rectifier, functions because of the ionization of the mercury in the tube. As the voltage applied to the tube increases, the mercury starts to ionize, slowly at first and then more rapidly until the maximum current flows. This rapid increase in current generates high-frequency currents which must be prevented from entering the radio set if efficient operation



Left, adapter; center, R.F. choke; right, 7-prong socket.

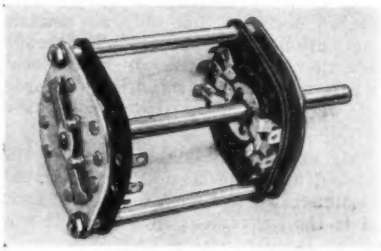
is to be secured. Manufacturers of the 82 recommend a small R.F. choke to be inserted in series with the "B" minus lead between the power transformer and the filter unit. Such a choke, manufactured by the Sun Radio Co., shown above at the center, is rated at 2 milli-henries and is capable of carying a current at 250 ma., the rated value of the 82 rectifier.

NEW 7-PRONG SOCKET

ALTHOUGH the new 7-prong tubes have not been announced at this writing, they are "just around the corner," but manufacturers have already produced the socket in anticipation of future business, as may be seen by reference to the photograph above (at the extreme right). The socket illustrated has many desirable features: the terminals are well spaced; riveted to a single piece of Bakelite, and a lasting, firm contact is insured by the fact that additional springs keep the contact surface tight against the tube pins at all times.

OAK S.-W. SWITCH

OAK MANUFACTURING CO., by perfecting an all-wave switch that is noiseless in operation, has solved the problem of "switching noise" for short-wave receivers and converters. This new switch

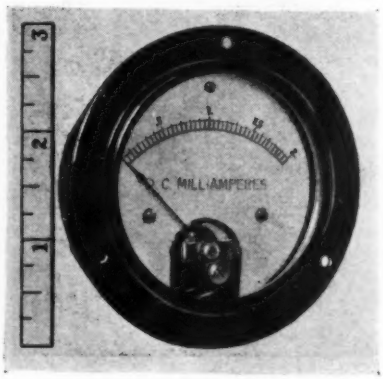


New Oak all-wave switch.

is said to have a lower contact resistance than any other radio switch now available—actually less than .001-ohm. The contact is of the double surface-cutting type as may be seen by reference to the photograph above, and is silver plated to minimize deterioration. Other important advantages are a well-defined stop action, Bakelite dielectric, and rust-proof cadmium-plated frame.

NEW BEEDE METERS

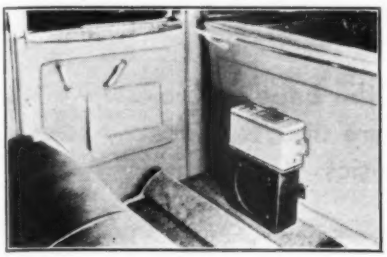
THE Beede Instrument Co. announces a new line of meters for the Service Man, one of which is illustrated below. They are low priced and some of the features include a D'Arsonval moving coil movement; large, laminated permanent magnets; new method of gap spacing which minimizes the collection of dust; and sapphire bearings. They are made in various sizes and sensitivities to accommodate every purpose.



New Beede meter.

CROSLEY ROAMIO

MANY new and distinctive mechanical features have been incorporated in the new Crosley Roamio automobile receiver depicted below. This small receiver



The Crosley Roamio.

is contained in a gray-silver metal case of a size small enough for easy installation either on the floor in front, between the driver and passenger; below the dash; in the rear, either fitted snugly against the driver's seat or between the passengers in front of the rear seat.

A new type antenna, slung under the chassis of the car by two clamps, eliminates the removal of the upholstery in order to secure a satisfactory antenna. The tubes used are as follows: A '36 as a combination oscillator and detector; a '37 as A.V.C. and second-detector; a '39 as first I.F. amplifier; a '36 as second I.F. amplifier; a '37 as first A.F. amplifier; and a 41 output stage.

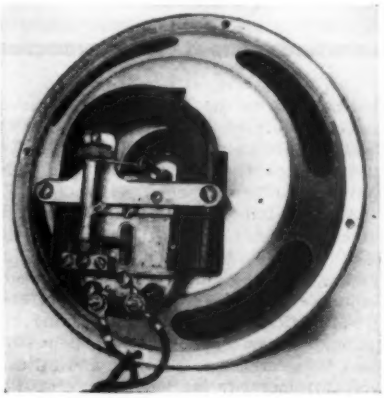
"OHM-SPUN" RESISTORS

"OHM-SPUN," is the name given to a woven, new type resistor unit, manufactured by the States Co. It is a light, non-inductive resistor. Instead of winding the resistance wire on an insulating form, it is spun in a plane, as shown below, so the current in adjacent wires flows in opposite directions, thus reducing the inductance to practically zero.

They are available in two types; the type A for frame and the type B for flat-surface mounting. They are 6-inches square and available in sizes up to 1,000 ohms, ¼ ampere each, in the type A mounting.

MAGNETIC SPEAKER

MAGNETIC speakers seem to be coming into their own again, to judge from the number of receivers, especially of the automotive type, that are making their appearance. As an aid toward portability, the new unique magnetic speaker, shown here, fills a long-felt want, for it is only 6¼ inches in outside diameter and but 3 inches deep. An unusual type of rotary drive between the armature and the cone is employed, which places the resonant frequency of the driving mechanism outside of the audible range. The pole pieces are of laminated iron and the rotary drive shaft is insulated electrically and acoustically at both ends by



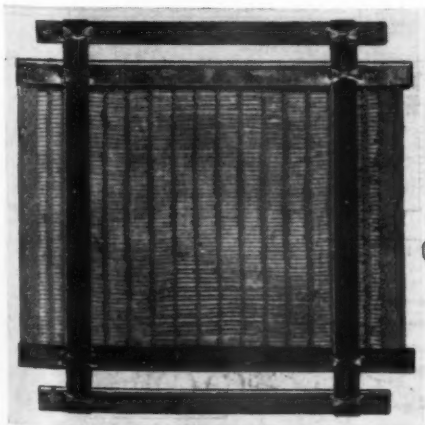
A small magnetic speaker.

rubber washers, to insure uniform reproduction.

This speaker, a product of the Best Manufacturing Co., is specially desirable in portable receivers, or in any location where space is at a premium.

A UNIQUE VOLUME CONTROL

BECAUSE the ear is more sensitive at the middle than at either end of the audio register, it is desirable to vary the volume of a radio receiver in such a manner that the low and high notes are not distorted. This feature is accomplished in the new device manufactured by the Allen-Bradley Co., shown below.

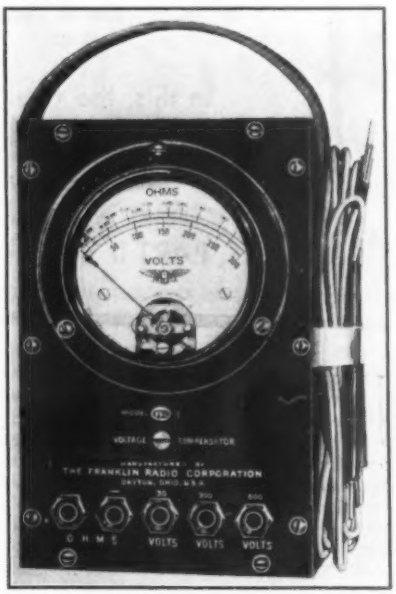


The Ohm-Spun resistor.

Here's a real volume control.

VOLT-OHMMETER

THE Franklin Radio Corporation has recently announced their new model 1 Voltmeter-Ohmmeter. Mounted in a laminated bakelite panel, and housed in



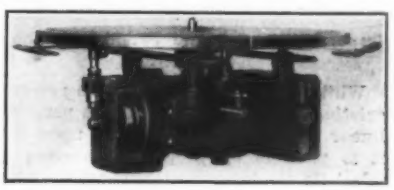
An volt-output meter.

a hardwood finished case, this instrument is designed for voltage ranges of 3, 30, and 600 volts, and ohmmeter ranges of 0-10,000 and 0-100,000 ohms. When used as a voltmeter, the instrument has a sensitivity of 1,000 ohms per volt.

All necessary pin-jacks and push buttons are mounted on the front panel; a 4½-volt battery for the ohmmeter is housed in the case itself. Of course, provision is made for battery-voltage compensation.

TWO-SPEED MOTOR

CHANGING from 33-1/3 to 78 R.P.M. is accomplished by a single lever in this phonograph motor produced by the General Industries Co., Inc. They are made



A two-speed phonograph turntable.

in two types, the model D for 10- and 12-inch records, and the model H for 16-inch records. All gears are enclosed and run in oil.

Two additional features are the facts that (1) it will not overheat in closed cases and cabinets, and (2) absolute uniformity of speed under severe variations of voltage fluctuations and record drag is maintained.

A LAPEL MICROPHONE

In this, the final of a series of articles concerning the Lapel Microphone, electrical characteristics and application are discussed, for the first time, by Dr. Saxl, authority on physics and radio.

PART II

By DR. IRVING J. SAXL*

IN our last article in RADIO-CRAFT we have seen how the new lapel microphone makes possible free movement of the speaker before the audience; non-obvious pickup of his voice; and allows sound amplification to be thus introduced in places, and with individuals, where microphonic amplification would not be feasible heretofore.

We have a tiny button, almost invisible, controlling the energy of many watts in loudspeakers all over an auditorium or convention hall. How is the transformation of the sound impulses into electrical impulses from the lapel microphone achieved?

We know already that the sound waves, striking the membrane of the microphone, produce a variation in the resistance of this microphone because the carbon granules filling the space between the carbon button and the actual membrane is pressed more or less together. If a voltage is applied between membrane and carbon button, the carbon granules will be able to conduct more current than if no pressure takes place. A sound wave is nothing but a succession of variations of the pressure falling upon the lapel microphone. These variations, which are almost proportionate to the sound energy produced, are converted through this microphone as a variation of electrical impulses.

Dependency of Sound From Microphone Current

While there is, at present, no receiver existing—the action of which is perfectly linear to sound waves falling upon it—it is possible by proper engineering to select its electro-acoustical characteristics in such a way that a satisfactory sound reproduction is received. The first part of this transformation from sound into electrical energy takes place in the microphone itself, and from here, naturally, the first control of the sound is affected.

Consider, for instance, the electrical resistance of the microphone. By choosing proper carbon grains it can be varied from less than one hundred to several hundred ohms. This resistance has to be matched to the impedances of the input transformer, as shown in Fig. 1, the best performance taking place if the impedances at the output and input sides have approximately equal values.

Another important item is the amount of current sent through the microphone. While it would seem probable, at first glance, that the amount of current would influence only the total quantity of electro-acoustical impulses transmitted through the microphone, it will be readily seen by reference to Figs. 2A and 2B that not only quantitative but definite *qualitative changes of the transmitted sound take place*, in accordance with the amount of current transmitted through the microphone.

Figs. 2A and 2B show the frequency response characteristics of this lapel microphone taken both at a current of 10 and 20 milliamperes passing through the microphone. It will be readily understood that the total acoustical energy produced with 20 milliamperes is differently distributed over the entire frequency range than it would be with 10 milliamperes. It is especially remarkable that the production of the highest frequencies, about 6,000 to 7,000 cycles, and the lowest frequencies, have increased several times more than the maximum energy output, which takes place at 3,000 cycles per second. This means, that by passing more

(*Continued on page* 238)

Fig. A

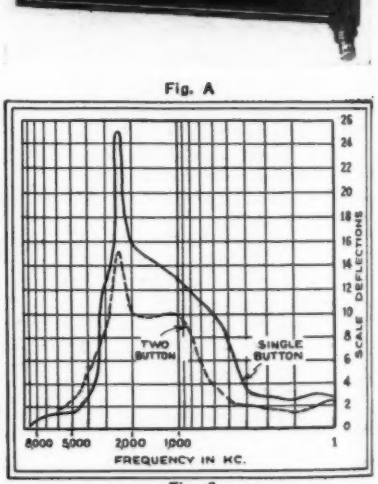


Fig. 3

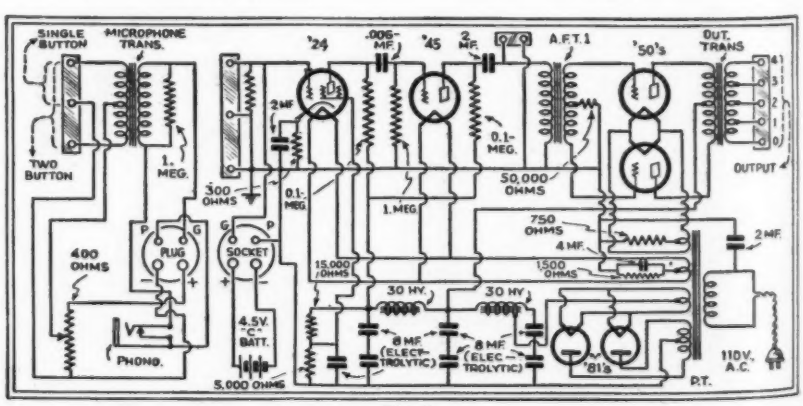


Fig. 1

*Consulting Physicist.

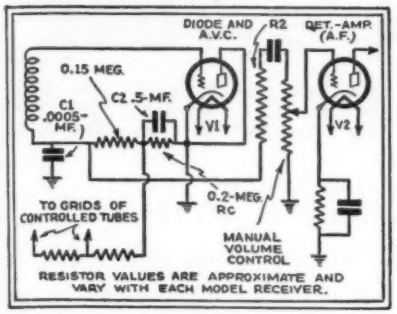


Fig. 5

Circuit showing the method of operating a diode detector and A.V.C. arrangement in one tube.

OPERATION AND SERVICE OF AUTOMATIC VOLUME CONTROL SYSTEMS

This is the final article dealing with the operation and service of automatic volume control systems.

By W. S. WILLIAMS

THE September issue of this magazine contained the first part of an article dealing with the theory of operation of A.V.C. systems. In this issue, the practical application of A.V.C. systems will be discussed.

The hookup of a complete receiver using A.V.C. is shown in Fig. 4. This receiver was built by the author and all resistor and condenser data are given. In other types of receivers the bias voltage and the plate resistor for the volume-control tube will probably have to be changed, and the best means of arriving at the proper values is by trial.

We have considered the necessary parts, the adjustment of the various operating potentials and some of the troubles encountered, but have said little about how the automatic volume control worked. The A.V.C. tube operates under conditions similar to a power detector. That is, it is biased to cut-off, or until no plate current flows. When a tube operating under this condition has an alternating voltage impressed upon its grid, the bias is reduced by an amount equal to the applied voltage during the positive half of the cycle, and plate current flows; the *amount* of current that flows being proportional to the impressed voltage. Now, on the negative half of the cycle, since the tube is already biased so that no current flows, an increase of negative voltage cannot produce any change in plate current. This results in an intermittent flow of current, which in the case of the power detector, produces a voltage drop across the primary of the first audio transformer. In the case of the A.V.C. tube, this current, flowing through the plate resistor, produces a voltage drop. This voltage is added to the normal bias afforded by the cathode resistor and decreases the sensitivity. Now, with the volume-control tube operating under the same conditions as a power detector, as soon as any signal is impressed on its grid the sensitivity would be reduced and the output would be very low. But as said before, it is biased to cut-off and a little beyond, and current does not begin to flow until the signal is strong enough to produce a satisfactory output.

Diode A.V.C. Systems

The second type of A.V.C. to be considered is the diode or two-element detector. In this system, a single tube serves both as detector and volume control. A conventional three-element tube is usually used, one element being connected to the cathode, while the other, either the plate or the grid, serves as the anode (plate) of a simple rectifier. In some instances the plate and the grid are connected together and used as one element.

Since the tube functions as a rectifier there is no gain, in fact there is a distinct loss which requires some amplification before it is applied to the output stage. In many instances, two or more stages are used before the output stage. The advantages claimed for this system are better fidelity, ability to handle a greater input, and the fact that the detector also functions as a volume control. This last sounds like doing the same work with fewer tubes, but this is offset by the A.F. amplification needed. As for fidelity, it is doubtful whether a power detector introduces more distortion than the extra audio stages. Besides, when a tone control is used, fidelity becomes almost an unknown quantity. With reference to its power handling ability, one cynical observer, after viewing a hookup

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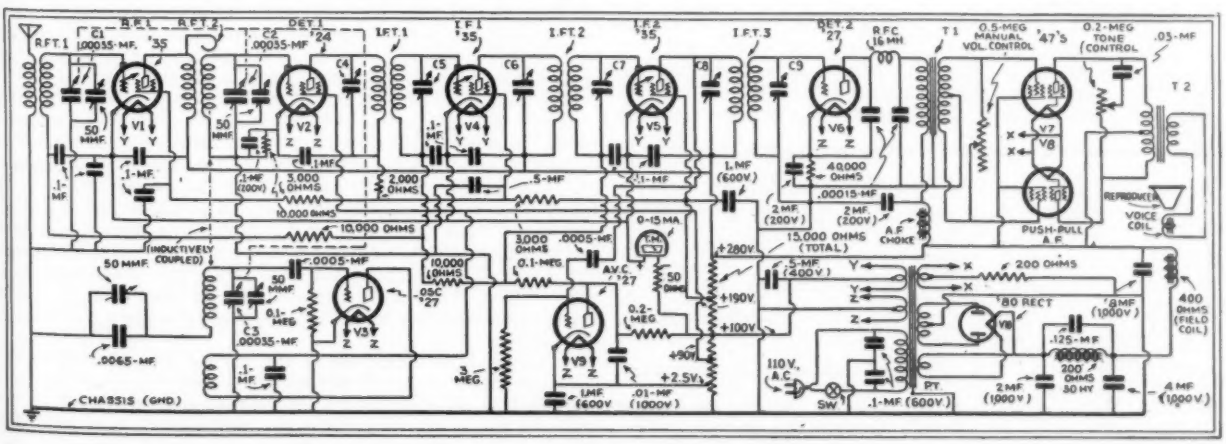
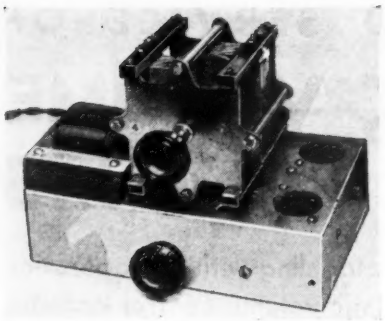


Fig. 4

The antenna and R.F. coil secondaries are wound with approximately 110 turns of No. 28 enameled wire; oscillator-grid coil, 85 turns of same size wire; tickler, 50 turns No. 32 enameled. All forms 1¼ ins. in diameter. The oscillator and grid coils are wound end to end (grids adjacent); the spacing between windings is ⅜ in. The antenna coil and R.F. chokes consist of 400 turns of No. 38 wire, scramble wound on a bobbin ½-in. in diameter. Tuning condensers, .00035-mf. The I.F. transformers were made at home by winding chokes as described and shunting them with 75 mmf. tuning condensers; the primary and secondary are spaced about 1. inch. The fixed section of the oscillator trimmer condenser has a capacity of about .00065-mf.; the variable section, a capacity of 200 mmf.



A very compact unit; it may be located in almost any convenient spot.

A MODERN "3-TUBE" REFLEX RECEIVER

Here's a description of a "three-tube" receiver which revives the once much-used reflex system.

By H. G. CISIN, M.E.

THE PZ pentode and other more recently announced tubes offer many fascinating possibilities to experimentally-inclined radio fans. These tubes may be used advantageously in T.R.F. circuits, superheterodynes and practically all other standard circuits. Realizing this, it was decided to try out some of the new tubes in the "reflex" circuit; several different combinations were tried, with very excellent results.

It seems as though the reflex circuit has been unjustly neglected by radio engineers. Five or six years ago, the writer had occasion to test a number of the reflex sets designed by David Grimes, and even at that time, remarkable results were obtained with the circuit. Taking advantage of the great advances and improvements in vacuum tubes, it is now possible to design a three-tube reflex, capable of outperforming a modern four-tube T.R.F. receiver and superior to the old reflex sets having six and seven tubes.

In the reflex circuit described in this article, a PZ pentode serves the dual functions of first R.F. and power A.F. amplification; the first R.F. portion of the circuit being untuned.

A tuned circuit is used to couple the two R.F. stages. A variable-mu type of pentode, the 58, is used in the second stage. This is coupled by means of a tuned impedance to the detector. The inductance is the secondary of a standard R.F. coil; the capacity is furnished by the second .00035-mf. section of the dual variable condenser.

One of the new general-purpose, type 56 tubes is used as a detector. The A.F. output of the detector is reflexed (fed back) through the type PZ pentode, which is thus made to function doubly—in the first instance, as an R.F. tube, and, finally, as the A.F. output tube.

The 2,500-ohm field coil of the dynamic reproducer serves as the filter choke; this inductance is bypassed by two 4 mf. electrolytic condensers; the capacity specified is sufficient for reducing the hum to a negligible value.

The screen-grids, cathodes, etc., of all the tubes are bypassed by small fixed condensers wherever necessary, thus assuring smooth operation.

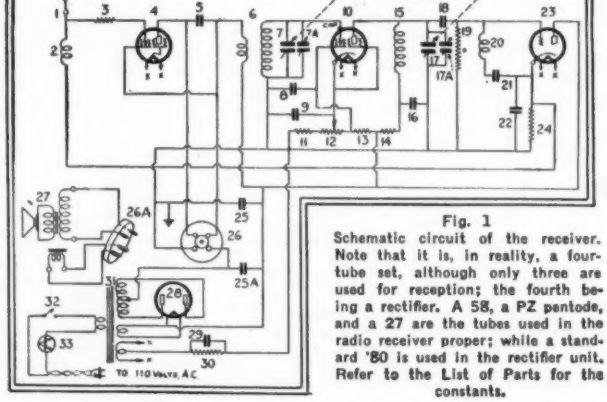


Fig. 1
Schematic circuit of the receiver. Note that it is, in reality, a four-tube set, although only three are used for reception; the fourth being a rectifier. A 58, a PZ pentode, and a 27 are the tubes used in the radio receiver proper; while a standard '80 is used in the rectifier unit. Refer to the List of Parts for the constants.

Volume is controlled by means of a 10,000 ohm "tapered" potentiometer connected in the cathode circuit of the type 58 tube. A fixed series resistor 11, limits the *minimum* negative grid bias.

An Amperite is employed in series with the primary winding of the power transformer. This keeps operation normal regardless of line-voltage fluctuations and also serves as a protection to the tubes and other components of the receiver.

Constructional Data on the 3-Tube Reflex

Six socket holes are drilled in the aluminum sheet before the latter is bent to the size specified for the chassis. The hole for the flush-mounted power transformer is also cut at this time.

After the chassis is formed, the wafer sockets are mounted. One of these, the speaker connection socket 26, is mounted on the rear chassis wall.

The power supply transformer, 31, is mounted next and then the dual variable condenser 7, 17 and the antenna binding post, 1. No ground post is necessary, as no external ground is used. The trimmer condensers, 7A, and 17A, are fastened in the insulated framework at the side of the dual variable condenser.

The chassis is next turned upside down and the R.F. choke 2, and the three electrolytic condensers 25, 25A and 29 are fastened on the rear chassis wall. Fixed condensers 21 and 16, R.F. choke 20 and the potentiometer-and-switch unit 12, 32 are mounted on the front chassis wall. The R.F. coils 6 and 15 are mounted on the underside of the chassis; dual condenser 8, 9 is mounted on the side wall. Finally, the various fixed condensers, flexible resistors, etc. are fastened in position. These are soldered as close as possible to the part with which they function. This reduces and simplifies the wiring.

Having completed the assembly, the set is now ready for the wiring. A break is made in one of the leads from the primary of the power transformer 31, and the switch 32 and Amperite socket 33 are connected in series in this circuit. Diagonally opposite socket terminals are used for the Amperite, (viz: one small prong terminal and one filament prong terminal).

The PZ tube uses a 5-prong socket. The control-grid connection is made at the terminal separated from the other four. Of these other terminals, the two center ones are for filament connections. Looking at the underside of the socket, the plate terminal connection is at the left, while the screen-grid is at the right.

Looking at the underside of the 6-prong type 58 tube socket, the two large terminals are for the filament leads. The next terminal to the right is for the cathode connection and to the right of this is the suppressor-grid terminal, which is connected back to the cathode *externally*: i.e., at the socket terminals. To the left of the filament terminals is the plate terminal and the remaining terminal is for the screen-grid connection. The control-grid connection is made at the cap of the tube.

The 56-tube socket is connected the same way as a '27. The

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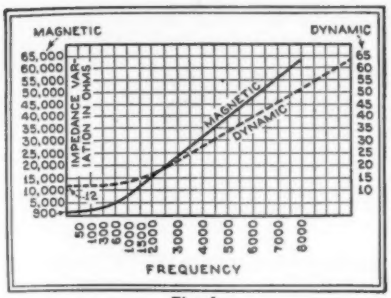

Fig. 4
Impedance variations of magnetic and dynamic speakers.

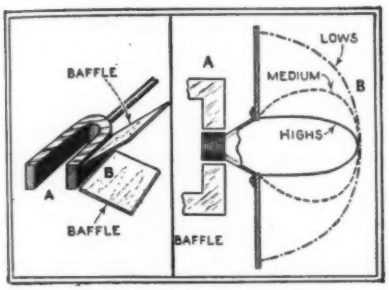

Fig. 5, left. A tuning fork baffle.
Fig. 6, right. Energy distribution from a dynamic speaker.

SOME INTERESTING LOUDSPEAKER FACTS

Some preliminary loudspeaker data is given in this first of a series of articles which will deal with the use of multiple speakers in radio sets.

By ELI M. LURIE

MANY technical articles have been written on various types of radio loudspeakers. To read and understand this avalanche of information would require considerable effort and time. In order to obviate such a tiresome procedure, this article has been prepared. Although it is by no means a complete thesis on the subject, it does contain the basic principles behind the modern loudspeaker. Some of the material has been compiled from notes, while most of the information has been taken directly from engineering journals dating as far back as 1925.

We generally associate the dynamic loudspeaker with recent progress but though the speaker of today is really new, the basic idea is not. As early as 1898 Sir Oliver Lodge, the renowned English scientist, was issued a patent on this identical thing. As a matter of fact, it was he who made the first permanent magnet dynamic type of speaker. In Fig. 1 is shown a sketch of his idea. A bar-magnet is used to supply a constant field. The voice coil is suspended as shown and when a fluctuating current passes through the coil it moves right and left. Thus, a real dynamic speaker was engineered over thirty years ago. This same principle is used in the Telautograph writing machine, the device used in banking institutions, hotels etc. for the reproducing of one's handwriting at a remote location.

One of the greatest handicaps in the design of a loudspeaker is inherent resonance. At the present time there is no type of speaker that does not have some resonance. By this is meant the tendency of the different parts of the speaker to continue to vibrate after the signal has stopped.

In the magnetic type of loudspeaker the cone driven by a *prime mover* or vibrating reed, usually a brass or steel rod of about 3/64-inch in diameter. A permanent magnet is generally the polarizing agent. Now the prime mover or reed is a tuned mechanical system that has a definite resonant point. This resonant point can be varied by damping or limiting the vibration of the reed, but if damping is used in order to either raise or lower the resonant point above or below the audible-frequency spectrum, there will naturally be a considerable loss in intensity.

Referring to the *diaphragm* or cone of the speaker, resonance will usually take place at the fundamental frequency, and will always be accompanied by a series of overtones which will vary in regard to the fundamental; in other words, they may be two, three, five or six times the frequency. In fact, there may not only be even harmonics, but any odd number as 3.6 times the fundamental frequency. This, naturally, aggravates the resonant condition, and in a way is similar to the reed in that they are both tuned resonant systems; the difference being that the resonant positions of one will not coincide with those of the other. If each is allowed to vibrate separately, it is possible to draw two resonant curves of the situation as in Fig. 2.

Harmonics in Speakers

In any electrical circuit where close coupling between two separate circuits exists, a resonant curve of the complete unit will have two distinct humps. This phenomenon is analogous to the combining of both the reed and cone, which is generally done by fastening the reed securely to the apex of the cone by means of a nut and thread. The resonance curve drawn for this combination will be that of a composite vibrating system and will have two humps similar to that of the two tightly coupled electrical circuits. The harmonics have not been shown; the actual effect of these would be to superimpose little ripples on the main curve.

Thus it can be readily understood why, up to the present time, we have been unable to obtain flawless and perfect reproduction. Perhaps it would be unwise to say that perfect reproduction is impossible, but it is true that, until we learn how to entirely eliminate the natural effects of resonance, perfect reproduction is very improbable. Efforts have been made to design a speaker with a continuance of resonant positions, the ultimate idea being to approach the ideal by making use of *resonant points all over the complete spectrum*, so that the aperiodicity of the speaker will depend on the resonant points and not on the vibration that is normally associated with the cone when in unresonant oscillation. The big problem with such a speaker is to cause the amplitude of vibration on the resonant positions to conform to the applied input in such a manner that a linear condition between input voltage and power output is the result. Otherwise, the amplitude of some of the resonant positions will be much greater than others, with the resulting non-linear response characteristic. The more resonant points that are used and the closer together these points are grouped, the closer the ideal condition will be approached, provided of course,

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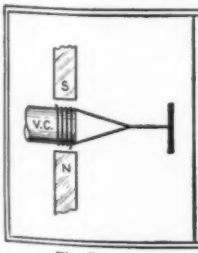


Fig. 1

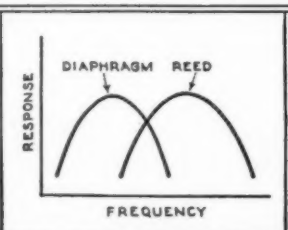


Fig. 2

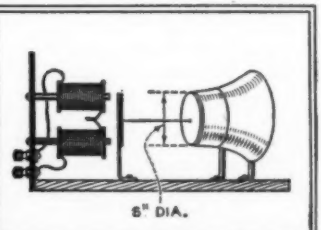


Fig. 3

THE WHY AND WHEREFORE OF
TUNING METERS

By L. VAN DER MEL

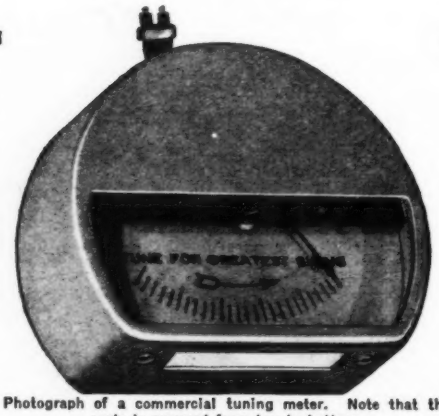
A TUNING METER, as the name implies, is a device used to indicate when the radio receiver is tuned to the peak of the signal (tuned exactly to resonance). It is not necessary that the device itself be a voltmeter or a milliammeter; all that is required is some form of indication when the signal has been fully tuned in. Tuning meters are used in sets where the selectivity is high, and therefore some means is necessary to determine exactly when the carrier is tuned in, so good quality may be secured. (Especially useful for short-wave work.)

In receivers of the superheterodyne type, a hissing noise may be heard when tuning between stations, which disappears when tuned to resonance because of the demodulation of the noise by the stronger carrier. If, in such receivers, a tuning meter is employed to indicate exact resonance, clear reception is insured. In any event, they are used in modern receivers, and a knowledge of where they may be used and "why" is essential for good servicing.

The D.C. Meter

Most people are familiar with the action of the ordinary D.C. milliammeter used so extensively in modern radio service work. It is commonly known that they will not read A.C. Why? Refer to the diagram of Fig. 1. Here is shown an A.C. voltage applied to a D.C. meter. During the positive halves of the cycle, the pointer of the meter tends, let us say, to move up the scale to the right, and during the negative half of the cycle it tends to move off the scale to the left. If the frequency of the voltage is very low, the pointer of the meter is able to follow the variations of the impressed voltage, although no steady reading can be obtained. As the frequency of the voltage is increased, the mechanical inertia of the pointer prevents its following the rapid variations in the applied voltage so that its reading (if it reads any value at all) *is the average value of the impressed voltage or current through the meter.* In other words, *any D.C. meter reads average values only.*

The average value of the sine wave shown in Fig. 1 is obviously zero, and hence the meter reads zero. Even if the frequency is so low that the pointer is able to follow the variations faithfully, the *average reading* of the meter is zero. This leads us to the conclusion that just because a D.C. meter reads zero does not necessarily mean that no current is flowing through it—in fact, if the peak value of the A.C. through a D.C. meter is great enough, the meter may burn out while it is reading zero!



Photograph of a commercial tuning meter. Note that the scale is reversed from top to bottom.

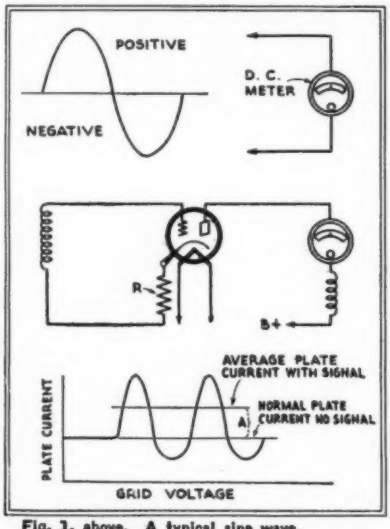
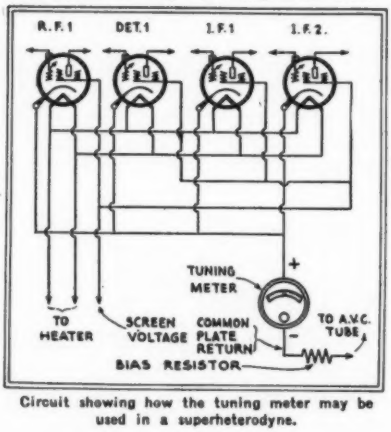


Fig. 1, above. A typical sine wave.
Fig. 2, center. Where the tuning meter may be placed.
Fig. 3, below. Why the tuning meter reads.



Circuit showing how the tuning meter may be used in a superheterodyne.

Where to Use the Tuning Meter

If the above is true, where, then, may such a meter be connected in a radio set in order to indicate the signal strength? Suppose it is connected in the plate circuit of an amplifier as indicated in Fig. 2. If the amplifier is of the class A type, the plate current variations through the meter will increase above and decrease below the value of current flowing when no signal is received; furthermore, the *amount* of increase is equal to the amount of decrease, and hence the signal variations have no effect upon the normal reading of the meter with no signal received. Obviously, the plate circuit of a class A amplifier is no place for a tuning meter.

If it cannot be placed in the usual amplifier, let us see what happens in the detector circuit. If the value of resistor R in Fig. 2 be increased so the tube is now a detector (which is the same as connecting the meter in the detector circuit), the amount of increase above normal current (the increase being due to a signal) is greater than the decrease below normal (due to the negative half-cycles of the same signal). The plate current now flowing through the meter, because of the signal, is shown in Fig. 3. It is seen that the plate current reading of the D.C. meter has increased by an amount A because of the signal. Thus we may conclude that a D.C. milliammeter may be inserted in the plate circuit of the detector (power detector assumed) and the meter may be called a "Tuning Meter."

Further, it should be pointed out that if the detector circuit in which the meter is inserted is of the grid-leak and grid-condenser type, the reading of the meter will decrease with an increase in signal strength which, of course, is contrary to the action of power detector.

The Tuning Meter in A.V.C. Circuits

An examination of some of the more recent commercial receivers employing automatic volume control (A.V.C.) will indicate that the tuning meter, if any, is connected in the plate circuit of the amplifier circuits, usually the I.F. amplifier in superheterodynes—which at first

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BUILDING YOUR OWN A. F. CHOKE COILS

By C. H. W. NASON

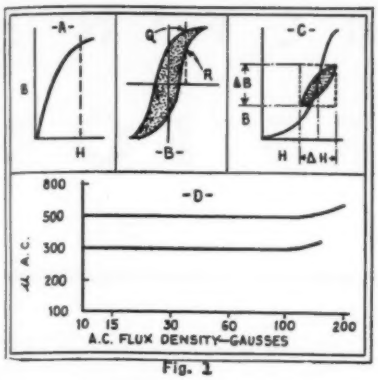

Fig. 1

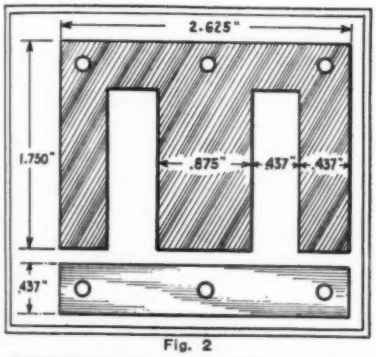

Fig. 2

FILTER chokes designed for a given inductance require calculations far outside the abilities of the average amateur. Before discussing actual simplified design methods it might be well, however, to discuss briefly the factors involved. Certain relations are involved in the design of the magnetic circuit which are briefly as follows:

Factor	*Equation*	*Unit of Expression*
H (the magnetic force)	4 pi N I	Gilberts per Cm.
Φ (the flux)	u H A	Maxwells
B (the flux density)	Φ/A	Gausses
u (the permeability)	B/H	Numeric value.

In these equations:
l = the magnetic path in cms.
A = the cross section of the core in sq. cms.
N = the number of turns.
I = the current in amperes.
pi = 3.1416

If we pass a gradually increasing current through the winding surrounding a magnetic material the relation of B to H is as shown in Fig. 1A. If the current through the winding is now decreased it will not follow the original path of the curve but will take another course as shown in Fig. 1B. Thus for a given value of H a twofold value of B corresponding to Q and to R in Fig. 1B are found. (This is due to the residual magnetism in the core and the condition is known as "hysteresis." Naturally, the smaller this loop the better is the core material for magnetic purposes.)

If with a fixed D.C. magnetizing force a small A.C. voltage is superimposed, a minor hysteresis loop appears as is indicated in Fig. 1C. It is from the minor loop that we determine the permeability of the core material to alternating currents, and it is $u_{a.c.}$ that we employ in calculating the in-

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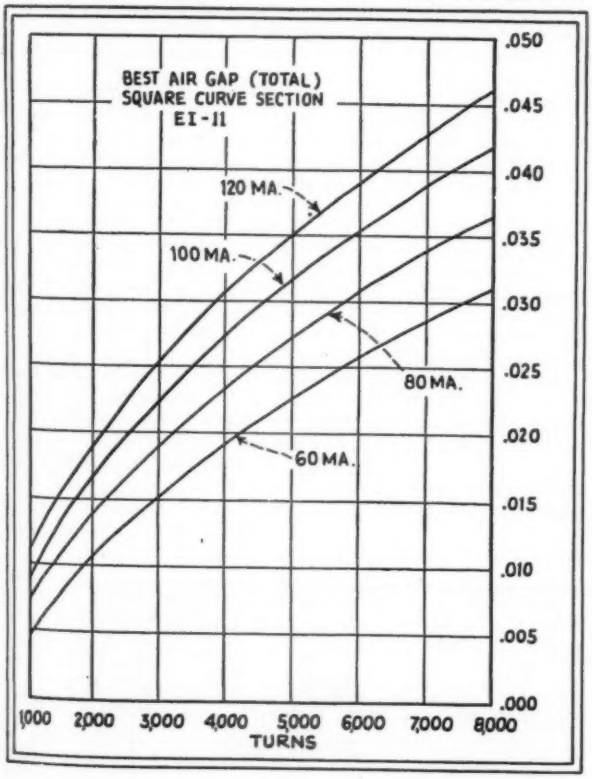
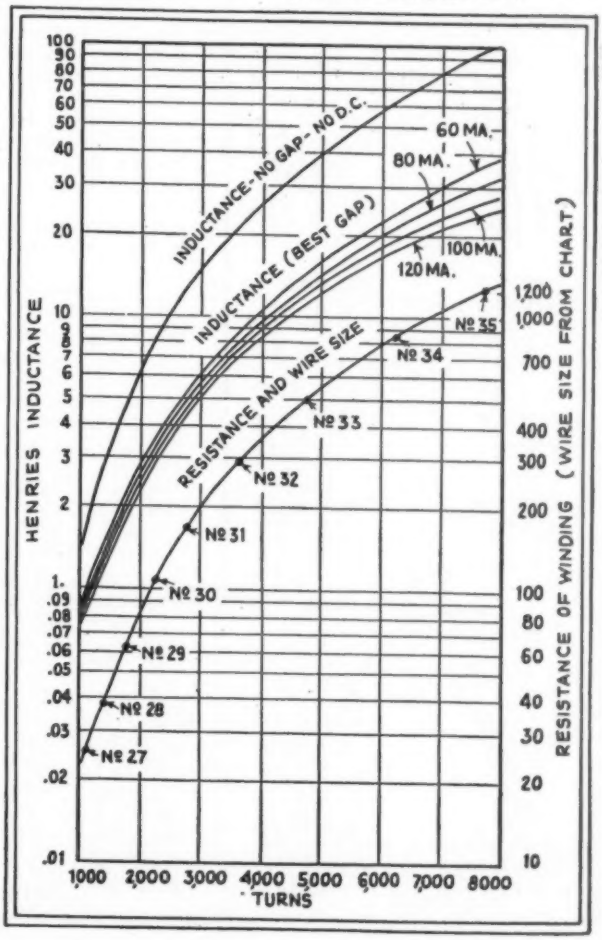


Fig. 3

For a given inductance and D.C. through the core, the number and size wire may easily be determined by means of the chart above. For an inductance of 20 henries, with 120 ma. through the core, 6,700 turns of No. 35 wire should be used on the core indicated in Fig. 2.



CONSTRUCTING ADAPTERS FOR

By F. L. SPRAYBERRY*

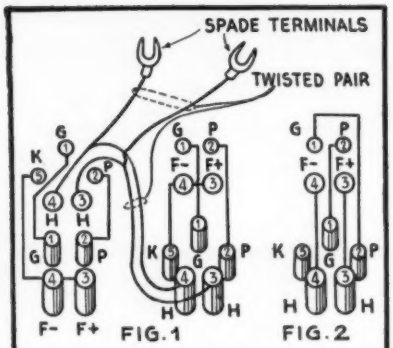


FIG. 1

FIG. 2

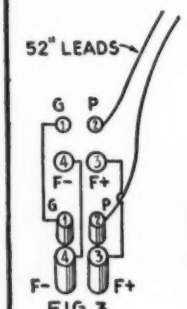


FIG. 3

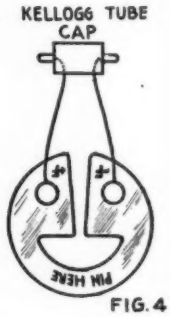


FIG. 4

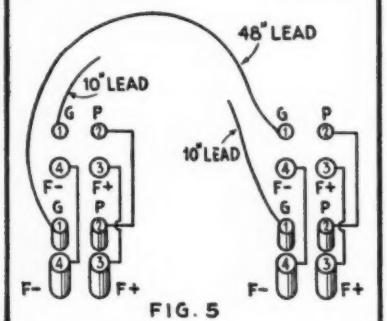


FIG. 5

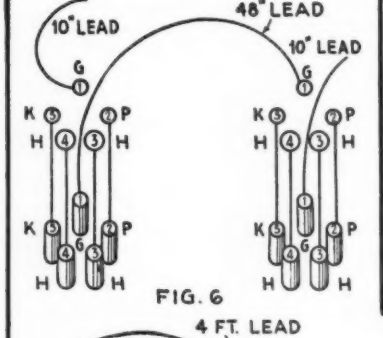


FIG. 6

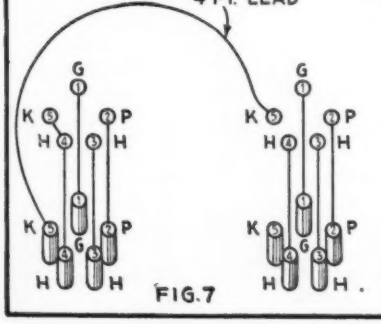


FIG. 7

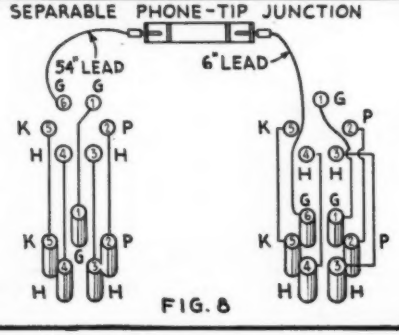


FIG. 8

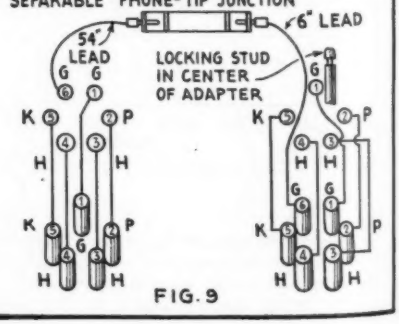


FIG. 9

THE average Radio Service Man views with dismay the ever changing design of vacuum tubes. Each season sees new types differing in design from previous ones. For a long time the standard tube was of the three-element type, the base of which used four prongs. However, the public demand was for A.C. operation, and the four-prong tubes available at that time were not suited for operation on raw A.C.—this was particularly true of detector stages. Therefore, the five-prong tube emerged and became a commercial product. At least one was used in every A.C. operated receiver.

The next major change in design was the screen-grid tube. First there was the four-prong '22 type which was to be operated on batteries. It was not long until the companion tube for A.C. operation was released; this was known as the '24. The next change in radical tube design was the power pentode or '47.

The present season sees a variety of new tubes announced, several of which have a six-prong base. There will also be a tube having a seven-prong base. The next will be a ——?

Radio testing equipment is, as a rule, designed to test the circuits of tubes in general use at the time that equipment is being manufactured. The manufacturers, of course, have no way of knowing the types of tubes that will appear in the future. As a consequence of the continued change in tube design over a period of several years, many fine pieces of radio test equipment have become obsolete, as they will not test the more modern tube circuits.

Many Service Men think that it is necessary to junk this equipment and purchase new equipment. This is not strictly true. Two courses are open to owners of old test equipment. This equipment may be rewired (old testers may easily be rewired and brought up to date), or; the owner of such equipment may use standard adapters making it possible to test the new tube circuits. Rewiring is usually more expensive and for that reason many Service Men prefer to use adapters.

The purpose of this article is to inform owners of old test equipment just how they may bring their equipment up to date by using adapters. We shall cover each well-known line in the following order: Jewell, Weston, Supreme, Hickok, Dayrad and Readrite. Data is given as to the type of adapter required for each of the most popular testers of these makes. As these adapters are commercially available, they are best described by reference to their regular (Na-Ald) stock numbers.

Adapters for Jewell Testers

No. 942, Fig. 1, is used with the 133 and 133A testers. It permits testing from a five-hole socket. It is to be attached to the regular test plug.

No. 945J, Fig. 2, is used with models 133, 133A and 137. The plate and grid connections are crossed, the cathode being open. The adapter is placed in a five-hole socket and the test plug in the adapter. Refer to the instruction booklet for these testers for the further use of this adapter.

No. 952J, Fig. 3, may be used with all Jewell models. It is a UX adapter having a split-plate circuit with two 52-in. external leads for connecting a pair of headphones or a milliammeter in series with the plate circuit.

No. 949K, Fig. 4, is a UX adapter hav-

* Service Consultant, National Radio Institute, Washington, D. C.

Twenty-nine out of one hundred eighteen adapters are described in this first of a series of practical articles.

TEST EQUIPMENT

ing 8-in. leads from both filament terminals. Used to test Kellogg "overhead filament" type tubes. It is inserted in the tube socket of the receiver. The filament leads are then connected to the overhead filament contacts of the receiver. May be used with Jewell Models 133, 133A, 198, 199, 408, 409, 444, 577, 578, 579, 581 and 660.

No. 932, Fig. 5, is a twin-adapter for testing the circuits of types '22, '32 and '34 tubes, using old-type 198, 199 and 581 testers. One adapter is placed on the test plug and the other in the tube socket of the receiver. The control-grid circuit is completed with the lead provided for that purpose.

No. 992, Fig. 6, is a similar adapter to adapter No. 932 except that it is made for types '24, '35, '36, '38, '39, '44 and '51 tubes. It is to be used with old-type 198, 199 and 581 testers.

No. 974, Fig. 7, is for testing the circuits of pentode tubes. This includes the types GA, LA, PZ, '33, 46 and '47. The adapter is of the twin type; one-half goes on the test plug and the other in the tester socket. May be used with the models 198, 199, 408, 409, 577, 578 and 579.

No. 965 DW, Fig. 8, is used for circuit tests on six-prong tubes such as the types 57 and 58. It is a twin-adapter, one-half of which is to be placed on test plug and the other in the tester socket. May be used with models 198, 199, 408, 409, 477, 578 and 579. By breaking the suppressor-grid lead this adapter may also be used to analyze the types 41, 42, PA and PZH tube circuits; connect the short lead to the control-grid stud on the analyzer plug and the long lead to the control-grid pin-jack in the analyzer.

No. 965DSW, Fig. 9, is for testing the circuits of types 57 and 58 tubes. It is a twin-adapter having split suppressor-grid leads which are to be connected to the tip jacks provided on the Jewell 444. Also tests types 41, 42, PA and PZH tube circuits as directed above.

No. 964DS, Fig. 10, changes a six-prong plug to a four-prong plug, used with the Jewell 660. May also be used with any tester having a basic six-prong plug.

No. 965DS, Fig. 11, changes a six-prong plug into a five-prong plug. Used with the Jewell 660 May also be used with any other tester having a six-prong plug.

No. 982, Fig. 12, is used with all tube testers having a 2.5 volt socket. This adapter is made for testing the type 82 and 83 rectifiers in tube testers only; do not use with a set analyzer. It incorporates a switch for changing from one plate to the other and includes the necessary resistor. Use the type '45 socket in the tube checker for the 82 and the type '71A socket for the 83. Used with models 209, 210, 214, 534, 535, 536, 538, 540 and 597.

No. 944JY, Fig. 13, permits the reading of plate current for the 2nd plate of an '80 rectifier. May be used with the following Jewell testers: 198, 199, 209, 408, 409, 577, 578, 579 and 581.

No. 965CG, Fig. 14, may be used with Jewell Models 214, 533, 534, 535, 538, 540 and 597 tube checkers to test the types 41, 42, PA and PZH tubes. The adapter is to be inserted in the '36 socket, or any five-hole socket wired for a filament potential of 6.3 volts.

No. 954 KPC, Fig. 15, is for testing the GA, PZ, '33, 46 and '47 type tube in Jewell 209 and 210 tube checkers. It is to be inserted in a 2 or 2.5 volt (filament) four-hole socket.

No. 965KS, Fig. 16, is for testing the types 57 and 58 tubes in the following Jewell tube checkers: Models 209, 210,

(*Continued on page* 245)

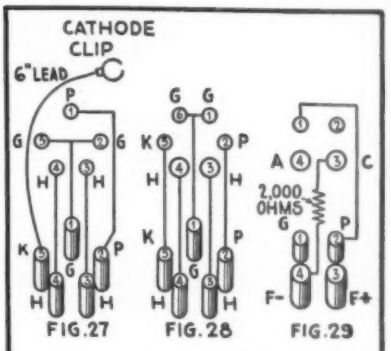


FIG. 27 FIG. 28 FIG. 29

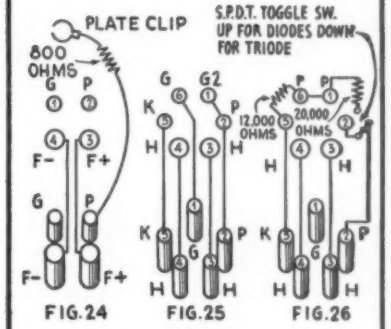


FIG. 24 FIG. 25 FIG. 26

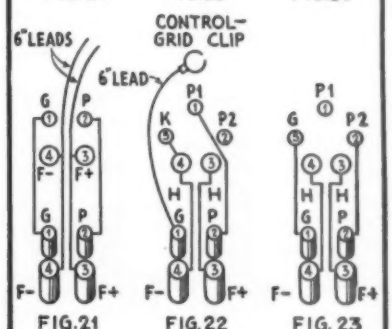


FIG. 21 FIG. 22 FIG. 23

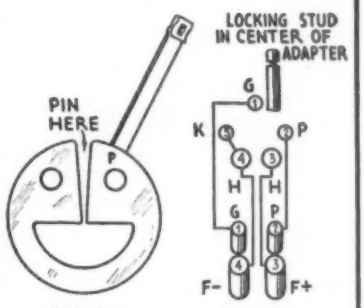


FIG. 19 FIG. 20

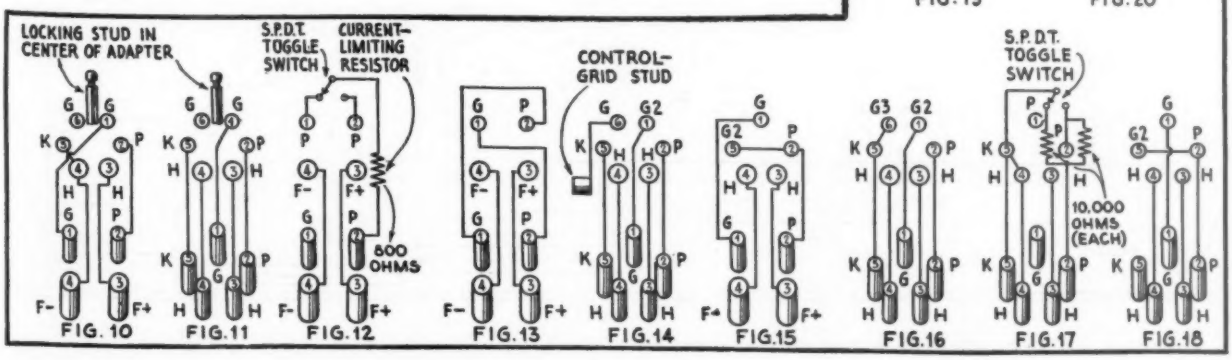


FIG. 10 FIG. 11 FIG. 12 FIG. 13 FIG. 14 FIG. 15 FIG. 16 FIG. 17 FIG. 18

HOW TO REWIRE THE HI-Q 29 FOR THE NEW TUBES

By HUBERT LEE FRAZIER

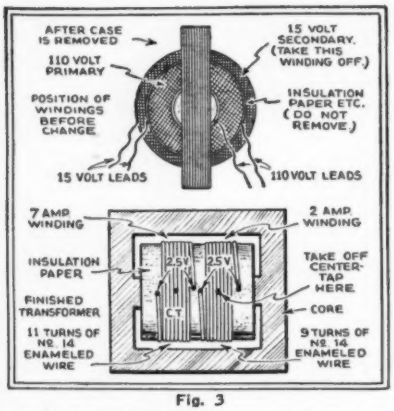


Fig. 3

Constructional details of the new filament transformer.

EVERYONE who owns one of the Hammarlund Roberts Hi-Q 29 receivers is familiar with their all around good performance. The R.F. coils are designed to be used with the high-gain type of radio tubes and the R.F. circuit is sufficiently filtered and shielded to permit the use of the new type tubes without adding any parts to it.

The Hi-Q "Master A.C." model is designed to use the 15-volt type of tube in all but the last A.F. stage. This tube is both difficult to obtain and expensive. Now, by making the changes described here, the operating cost and expense of tube replacement are reduced to a minimum; also, the sensitivity and tone of the receiver are greatly increased. Figure 1 is a complete wiring diagram of the finished receiver and power supply, with all the parts marked for identification.

Disconnect the reproducer and power-supply unit from the receiver chassis and remove both the power-supply unit and the set chassis from the cabinet. Remove the tubes from the power supply unit and the set chassis, and begin the rewiring of the latter unit first.

Rewiring the Receiver

Remove the wires from the socket terminals marked P, F and F, connecting the wires taken from terminals marked F to the terminals marked P and G of Fig. 2. Next, remove the wire that runs from the post marked C on the filament winding of the power pack to the bottom of the "C" bias resistor, and connect the wire that was taken from terminal P on the tube socket to the post marked C on the filament winding of the power compact. Now remove the wires from the posts marked F on the power compact, and run new leads from these posts to the socket terminals marked F. The filament transformer is now disconnected from the A.C. line cord. This transformer is either rewound according to instruction in Fig. 3, or is replaced with a new unit having one 2.5 volt, 2 ampere winding, and one 2.5 volt, 7 ampere winding. Mount the new or rewound filament transformer on the power supply unit and continue with the rewiring.

Connect the wires taken from the posts marked F on the power pack to the 2.5 volt, 2 ampere winding on the filament transformer. Upon tracing these wires they should be found to connect to the fahnestock clips numbered 5 and 6. These are the filament supply leads for the last A.F. tube. Connect the two wires taken from the filament secondary of the 15-volt transformer to the 2.5 volt, 7 ampere winding of the new filament transformer; from the center-tap of this winding connect a wire to the bottom of the "C" bias resistor. The center-tap of the 2.5 volt, 7 ampere winding connects to clip No. 10. Reconnect the primary of the filament transformer and the A.C. line cord.

The power supply unit is now ready for testing. Use a high-resistance type of

(*Continued on page* 235)

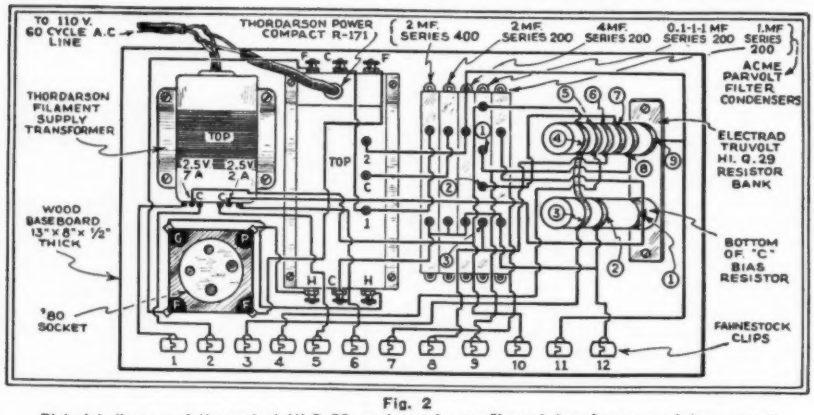

TO 110 V. 60 CYCLE A.C. LINE
THORDARSON POWER COMPACT R-171
2 MF. SERIES 400
2 MF. SERIES 200
4 MF. SERIES 200
0.1-1-1 MF. SERIES 200
1 MF. SERIES 200
ACME PARVOLT FILTER CONDENSERS
THORDARSON FILAMENT SUPPLY TRANSFORMER
ELECTRAD TRUVOLT HI. Q. 29 RESISTOR BANK
WOOD BASEBOARD 13" X 8" X ½" THICK
BOTTOM OF "C" BIAS RESISTOR
'80 SOCKET
FAHNESTOCK CLIPS
TOP
2.5V 7A
2.5V 2A

Fig. 2

Pictorial diagram of the revised Hi-Q 29 receiver. A new filament transformer must be rewound as shown in Fig. 3.

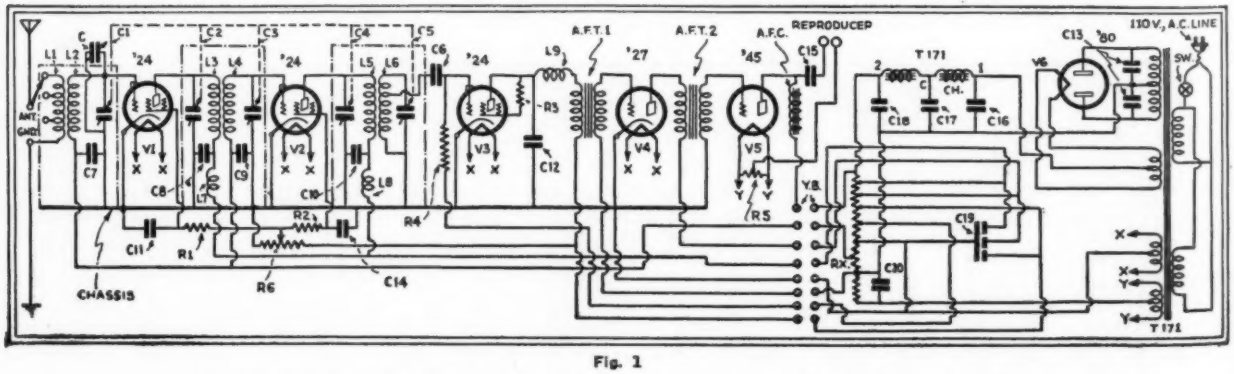
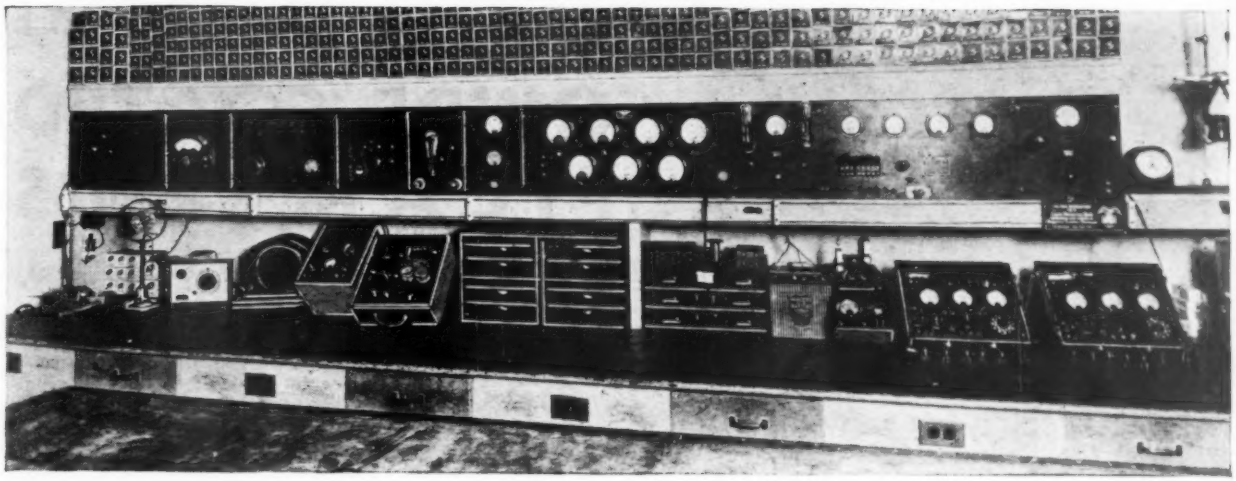


Fig. 1

Complete schematic circuit of the revised Hi-Q 29. The old tubes have been replaced by the types '24 and '27.



PROFITS VIA THE SERVICE LAB.

By M. C. MANCILL

WE have found, after a thorough test over a period of six and one-half years, that a well-equipped service department is an asset to the radio sales department as well as profitable from a servicing standpoint. These statements are based on our operations as the radio service department and RCA Victor sales division of a music house which has been established for 27 years.

This laboratory, illustrated above, was built with the intention of expansion as business increased and as improved equipment was brought out. There are ten (10) panels in use, most of them being either 12 x 12 in. or 12 x 24 in. The connections are shown below. The first panel reading from left to right is blank; it was at one time a tube analyzer, but was replaced with a Sylvania Vis-O-Meter.

Panel No. 2 is a decibel meter which is used to test speech amplifiers and audio systems in radio sets.

Panel No. 3 is a tube rejuvenator which was used in the early days for rejuvenating thoriated-filament tubes; it is still very useful, but to make it even more valuable, six UY sockets have been wired in for preheating, in conjunction with a Weston 565 analyzer (which is kept in the laboratory).

Panel No. 4 is a Leeds and Northrup Wheatstone bridge which is used for checking resistances and for making various other measurements.

Panel No. 5 is a neon glow tube with a self contained power supply delivering about 10 ma. at about 450 volts D.C. This equipment is used for checking bypass and filter condensers.

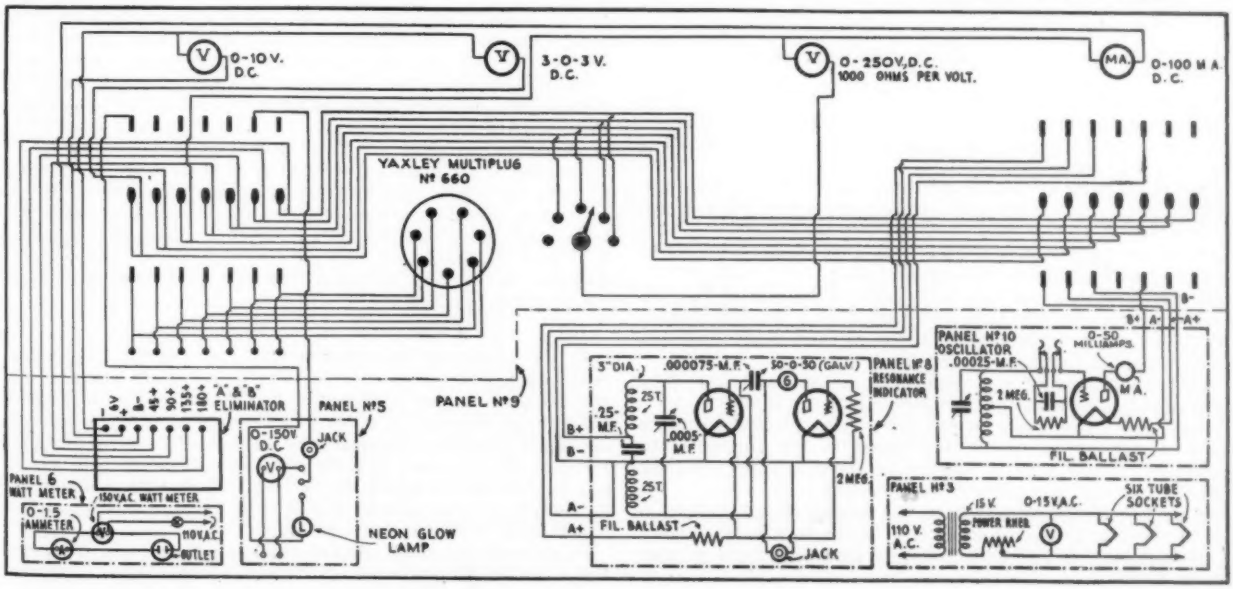
Panel No. 6 contains an 0 to 150V. A.C. voltmeter and an 0 to 5 A. A.C. ammeter. These are used for finding the power consumption or current drain of the set under test. This combination is also very handy for locating intermittent power pack trouble.

Panel No. 7 is a Standard Jewell analyzing panel.

Panel No. 8 is a grid-dip instrument or resonance indicator which is used mostly for locating variable condenser short-circuits.

Panel No. 9 contains an 0 to 10 V. and an 0 to 250 V. D.C. voltmeter; also a 3-0-3 A. D.C. ammeter and an 0 to 100 ma. D.C. milliammeter. To these meters

(*Continued on page 237*)



HOW TO REACTIVATE OXIDE-COATED FILAMENTS

By RAYMOND SHAW

Fig. A
Front view, showing the location of the controls.

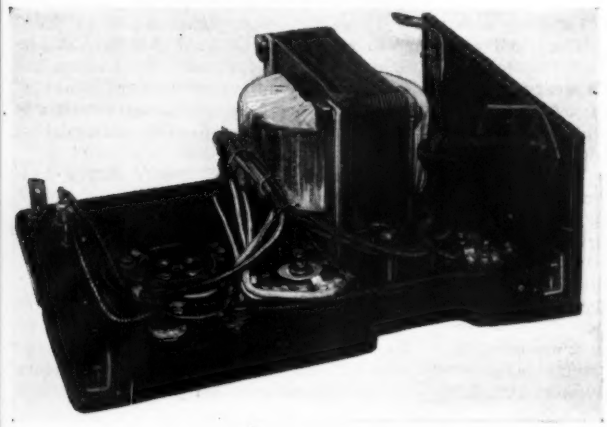


Fig. B
Rear view showing the location of the component parts.

I CAN hear some of you Service Men saying that you have no interest in rejuvenating tubes since their cost has dropped to the present low level; also, that you can make more money selling new tubes—but:

First, how many of you Service Men, on a hurry-up job, have found one or more weak tubes in a radio set and discovered that your stock in that kind of a tube was low? Of course, this should never happen, but it does! How handy it would be to be prepared to build up those weak tubes;

Second, how many of you have taken a job where the customer's finances are low (and many are during this depression) and the job must be done at the lowest possible figure at which you can make a profit or you don't get it?

Third, how many of you would like to cut down your tube replacement costs when reconditioning sets for resale?

Fourth, also, how many of you have an unsatisfied customer whose ninety-day guarantee has expired? My answer is to rejuvenate, if the tube is weak, and impress your customer with the fact that you know a thing or two.

To my knowledge no article on the subject of A.C. rejuvenation has been in print and while it is known to a small number of radiotricians, it has been jealously concealed from Service Men at large.

Rejuvenated Tubes Stand Up

The author has been regularly rejuvenating A.C. tubes for over a year at this writing, and most of the tubes so treated are in service to-day. A charge of twenty-five cents is made for each tube and a clean profit of many dollars has been made during 1931. This has resulted in convincing customers of his ability as a radiotrician and so has sent him many new customers.

The time required to treat most tubes is from ten to thirty seconds and the first figure is sufficient for most tubes; however, it takes longer when treating heater-type tubes. The author wrote to the manufacturer of one of the best known tubes asking if A.C. tubes could be renewed. They replied "that there was no known method of reactivating A.C. tubes as the oxide-coated filaments would not stand high-flashing voltages without burning out." The latter part of their statement is true, but as to the first part, the author had been successfully rejuvenating A.C. tubes for over six months at the date of their letter!

The Theory of A.C. Rejuvenation

The average vacuum tube has four important constants, electronic emission, plate impedance, amplification factor (mu), and mutual conductance. All of these constants are governed by one major element—the condition of the filament. If the physical structure of the tube elements have not been altered and the filament is intact, but does not emit the necessary number of electrons, then the tube cannot correctly function and is generally spoken of as having come to the end of its life.

Now if we could force the filament or cathode to emit the proper number of electrons, then its life would be renewed; this is exactly what we will do. Most Service Men are perfectly familiar with reactivating D.C. tubes wherein no plate or grid voltages are applied; but to reactivate A.C. tubes, high grid- and plate-voltages are necessary and normal filament voltages are used on some tubes while others require twice their normal filament voltage. In A.C. tubes, the abundance of electrons necessary for their operation is due to an oxide coating on the filament or cathode, depending upon the type of filament or heater used in the tube. When this oxide coating is burned away, sufficient electrons are not emitted for satisfactory operation and the tube is useless. Application of high grid and plate voltages will overcome this condition in the majority of cases by reconditioning this oxide coating and so renew the tube's life.

Construction of the Rejuvenator

In an emergency, any high-voltage transformer (preferably one with a 700-volt secondary) can be hooked-up, and the job can be done without switches of any kind. But it will more than repay the cost of parts and time to build a permanent hookup if carefully assembled and housed, either in a portable form or in any small box for shop use as shown in Figs. A and B. An R.C.A. model 16 cabinet will do very well, as an A.C. oscillator may also be placed in it and operated from the transformer used for the rejuvenator.

In the portable outfit shown, the laminations of a defective R.C.A. model 60 power transformer was used. First heat the

While tube reactivation is not new, nevertheless it is generally conceded that the process is not a very successful one with oxide-coated filament tubes. In this interesting article, the author describes how to build and use a device that really works—the editors made exhaustive tests—and its cost is about five dollars!

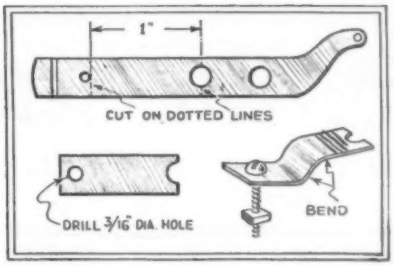


Fig. 3
Breaking apart the phone jack.

cased transformer (turned upside down) over a flame and place it on two wooden strips in a pan covered with an old can. (See RADIO-CRAFT for April, 1931, page 492). Uncase the transformer and as soon as the remaining compound begins to congeal, remove it; now carefully remove the laminations. Place them in gasoline and leave over night to loosen the tar coating; then wipe clean with cloth. If the primary winding is intact, save it. In any case, be sure to count the turns of the 2.5- and 5-volt windings so as to know the turns-per-volt required when rewinding.

The high voltage winding was burned out in my case, so it was rewound with No. 28 enameled wire and tapped for 200, 325, and 750 volts. (See RADIO-CRAFT for Sept. 1931, page 166 on the "Design of Power Transformers.") Then wind on the filament windings, tapping at 1.5, 2.5, 5, and 7.5 volts, as shown in the schematic circuit of Fig. 1.

Special Switches

While high grade push buttons were used in the portable outfit, the author prefers special constructed switches (SW 1 and SW 2 of Fig. 1) as very rapid-acting switches are required. A miniature D.P.D.T. knife switch was used because no D.P.S.T. switch was in the "junk box." Refer to Fig. 2. The cross-bar was removed. Two fiber rods ¼ x 1½ inches were slotted and drilled for rivets made from No. 18 copper wire. These rods made excellent handles when attached to the switch blades; the miniature switch was then mounted on a piece of bakelite. A phone jack was torn down and two of the springs were cut to 1-inch lengths, as shown in Fig. 3.

How to Apply High Voltages

In general, use the lowest possible voltage on grid and plate that will give results. The high voltage to the grid and plate must be flashed; never use voltages so high that the filament turns white and emits a large number of sparks, because it will burn out. Do not flash high voltages long enough or so rapidly as to turn the plate of the tube red. As a general rule, begin using flashing voltages of 1½ times the tube's normal plate voltage and, by steps, raise it to 2 or 2½ times normal voltage. This may sound very difficult but it really is not, for few tubes burn out except where excessive voltages are used and then the operator is the only one to blame. *But do not become discouraged if you burn out the filaments of the first few tubes.*

Table of Voltages to be Used

The author has successfully treated the following tubes at the voltages as given:

Type of Tube	Voltage of Filament	Voltage flashed on Grid and Plate
'12A	5	200-400
'71A	5	200-400
'26	1.5	200-400
'30, '31, '32	2.5	200-400
'45	2.5	400
'50	7.5	400-600

'27, '24, '35, '51; apply 5 V. approximately one minute and test before applying grid and plate voltages; then apply 200-400 volts if necessary.

Some '80's will respond to the treatment; especially where one plate shows full emission and the other, low.

Color the Important Thing

Set your apparatus up in a dark place, or at least shade the tube from strong light, *for you must watch the color changes brought about in the spaces inside the plate and surrounding the grid and filament.* Most tubes show a blue-green color in the above mentioned space when the desired results are obtained. Better to test the tube frequently than to over-do the process. Those tubes that develop this blue-green color are the easiest treated. Some tubes will not develop this color but will show a bluish color not very pronounced and still will show an O.K. test. Experience, only, will fix in your mind the colors required to develop full emission from the tube.

Heater-type tubes of the 2.5 volt group should be treated with 5 volts on the filament from ½ to 3 minutes and then tested

(*Continued on page* 250)

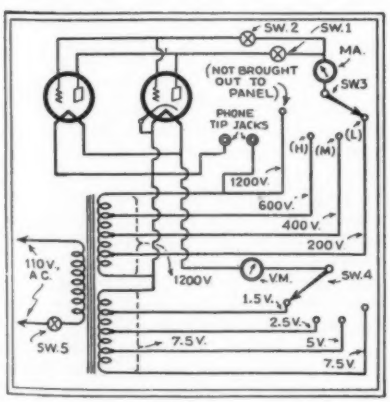


Fig. 1
Schematic circuit of the reactivator.

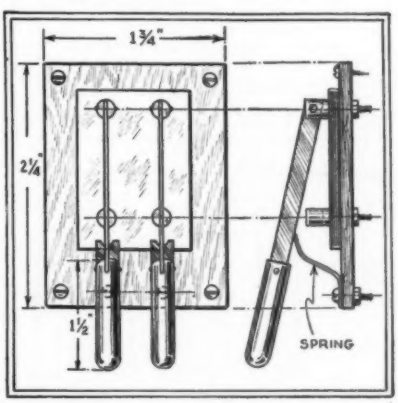


Fig. 2
Construction details of the knife switches.

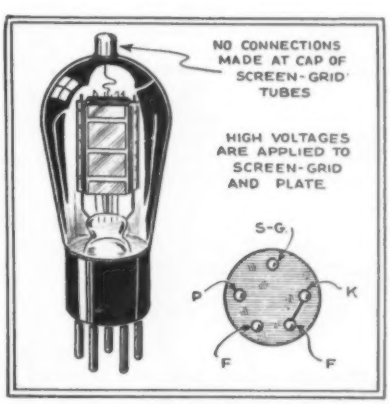


Fig. 4
How the five-prong tubes are reactivated.

THE SERVICE MAN'S FORUM

Where His Findings May Benefit Other Radio Technicians

SERVICE MEN PREFER RADIO-CRAFT

Editor, RADIO-CRAFT:

I have just purchased my July issue of RADIO-CRAFT and I want to say it's a dandy. Just the information and data that most Service Men would like to have.

I am one who buys his radio magazines after a glance through to see "What it has for the Service Man."

The Editorial, more dope on ohmmeters, circuits, latest equipment, two new tubes, how to use a set analyzer, operating notes, the Service Man's Forum, a simple oscillator, RADIO-CRAFT kinks, information bureau and, especially, Mr. Silver's article on audio systems—and, just what I have been waiting for, *a tube table.*

It would make me go plenty far out of the way to get the July issue if I didn't purchase it next door.

I spread the good news around and also have been asked, "Did you get the July issue of RADIO-CRAFT?"

There are three magazines most Service Men buy: RADIO-CRAFT, Radio News and Radio Callbook. Whichever has the most data or, we might say, dope for the Service Man, is taken along to the shop.

I feel that *the majority of Service Men buy* RADIO-CRAFT, if they have not already a subscription. So I can see, in my light, that if you continue putting to press a magazine such as your July issue you certainly will have us all buying your magazine every month.

Remember, radio is our bread and butter and we are not interested in experiments, etc. We want more issues just like your July number.

I write this as a matter of information to you and for my own lookout; also, I can safely say, the lookout of the majority of men who are in the service business.

Nothing is more disgusting than to purchase a radio magazine and find that three-quarters of it are devoted to foolish experiments for some office man in his leisure hours.

Well, I bet there are plenty of us that have your tube chart enclosed in cellophane and hanging over the bench for reference.

AL. MCNEILL,
2315 *E. 75th St.*,
Chicago, Ill.

(Thanks for the bouquet. Although we try to present to the Service Man new and original material on every phase of this branch of radio which might enable him to increase his income, it should be realized that there are thousands of experimenters who have no commercial interest in radio and to whom the thrill of watching experimental ideas put into practice means more than the lure of gold. A few general-interest articles, and miscellaneous kinks serve to keep these folk posted on "doings" of interest to them. Let's all shake hands, in RADIO-CRAFT, "for the good of the order."—*Editor.*)

"RAH! RAH! CORRESPONDENCE SCHOOL!"

Editor, RADIO-CRAFT:

This is to bridge the space between us and to pat Oscar Prescott, of Vinton, Iowa, on the back. The sum total of my radio knowledge was gained through a correspondence course, augmented, of course, by keeping up to date via the monthly issue of RADIO-CRAFT.

A short time ago a Heritage radio set (U. S. Radio & Television Corps.' old model 80) was brought in to me to fix; the thing was deader than any door nail you ever saw, as far as reception was concerned, although the speaker sounded alive. Pulling out the first R.F. tube and sticking the aerial into the plate receptacle of the socket produced music, thus localizing the source of trouble. Finding it was a different thing.

Everything checked O.K. down to the volume control (which is a potentiometer across the input—and in these models with a serial number above 5050 the aerial connects to the aerial). A few minutes thinking gave me my answer—the slider was not insulated from the shaft. The shaft, in turn, made contact to the chassis through a small metal plate used to support the unit. Therefore, the aerial was completely grounded. I questioned the owner as to other Service Men working on it. Here is the rub:

The owner, fearing to trust his radio to a correspondence school man in a village of 1,200 people, had taken it to a high-powered, well advertised service shop in Appleton, 21,000 population, 40 miles away. The owners of the shops are graduates of Wisconsin U. They had replaced the original control with an Electrad unit, but had overlooked that one point, and the set came home worse than when it left. Evidently, they had not tested the set after making the repair.

One of several things I have done was to cut a mounting plate out of heavy fiber and, presto—the job was done.

All this is no argument against the residence school, but just the same, you can get to know your radio from a correspondence course *if you really study the material furnished you.* At least, you can learn fundamental principles and after that, it is up to you. So, therefore, I say, "Rah! Rah! Correspondence School!"

L. W. NYGAARD,
Box 36,
Gillett, Wis.

(And, we might add, three cheers for "horse sense"—for who in this day and age would expect to repair a radio set and then turn it out of the shop without running an operating test on it! Apparently, population hasnt a thing to do with reasoning ability. However, the fact that the receiver left the shop without being tested may have been an accident. It so happens that the Service Men may have *thought* that the set was tested when, in fact, it was not. Accidents will happen, even in the best of families, and no one can be blamed for making a mistake once in a while; *for the man who never makes a mistake never does anything.*—*Editor*)

THE Official Radio Service Mens' Association, sponsored by RADIO-CRAFT, *invites all Service Men who are not members of the Organization to write for an application blank. It is the official service organization of this magazine and is maintained solely for the interests of Service Men. Membership cards are issued upon passing a written examination which is forwarded by mail. Write for yours today. The O.R.S.M.A., 98 Park Place, N. Y.*

Official lapel button of the O. R. S. M. A.

CALL YOUR NEIGHBORHOOD
RADIO DOCTOR
WHEN YOU HAVE RADIO TROUBLE

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References furnished if desired.

Benefit by our eight years experience in repairing.

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SHORT CUTS IN RADIO SERVICE

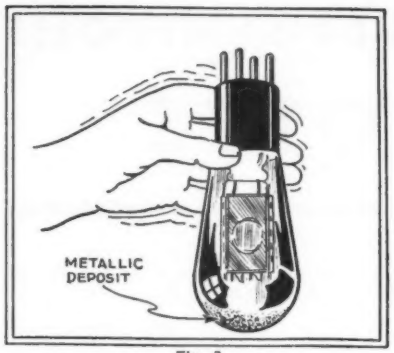


Fig. 1
Shake well before testing.

Prize Award

HOW OLD IS THAT TUBE?

By Leon J. Fox

AT the present time there is no way to tell the age of vacuum tubes. When the Service Man replaces the tubes in a receiver he wants to be sure that they will perform a reasonable length of time in order to satisfy the customer. While there are old tubes in the set that test O.K., the Service Man does not know how long they will last. In many cases he may get a recall and will have to replace these tubes free of charge.

Upon turning the doubtful tube upside down and tapping gently, it will be noticed that there will collect a deposit coating from the filament. After a careful examination of a number of tubes of known age, the quantity of this coating will determine how long the tube has been in operation. The heater-type tubes will show less coating. See Fig. 1.

I have found that it is a good idea to show this defect in the old tube to the customer as a reason for replacement.

CURING RECEIVER HUM

By Randolph Mutschelknaus

NO doubt many Service Men have found sets using electrolytic condensers that will operate all right for half an hour or more, then develop an abnormal hum; and after trying everything the manufacturer recommends to cure this hum, have replaced the electrolytic condenser and thus cured the trouble.

In the event that a replacement unit is not available, the Service Man may be interested in the simple procedure followed by the writer in eliminating this hum and giving the customer service until a new condenser of longer life can be connected into the set.

Drill a small hole in the top of the condenser (through the hard rubber), as illustrated in Fig. 2; turn on the radio set and fill the condenser with distilled water until the hum stops. Then, add enough water to raise the solution level in the condenser about ½-in. or more. Seal the hole in the condenser top with wax and the set is ready for many more months of service.

Be careful not to touch the condenser connections or the water container (if it is metal), while filling, or you may receive a shock.

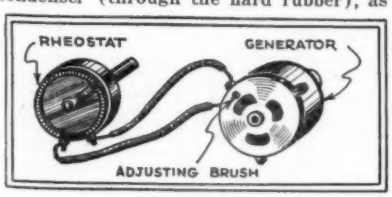


Fig. 4
A very neat device for regulating the charging rate of the car battery.

$10 for Prize Service Wrinkles

Previous experience has indicated that many Service Men, during their daily work, have run across some very excellent Wrinkles, which would be of great interest to their fellow Service Men.

As an incentive toward obtaining information of this type, RADIO-CRAFT will pay $10.00 to the Service Man submitting the best all-around Radio Service Wrinkle each month. All checks are mailed upon publication.

The judges are the editors of RADIO-CRAFT, and their decisions are final. No unused manuscripts can be returned.

Follow these simple rules: Write, or preferably type, on one side of the sheet, giving a clear description of the best Radio Service Wrinkle you know of. Simple sketches in free-hand are satisfactory, as long as they explain the idea. You may send in as many Wrinkles as you please. Everyone is eligible for the prize except employees of RADIO-CRAFT and their families.

The contest closes the 15th of every month, by which time all the Wrinkles must be received for the next month.

Send all contributions to the Editor, Service Wrinkles, c/o RADIO-CRAFT, 98 Park Place, New York City.

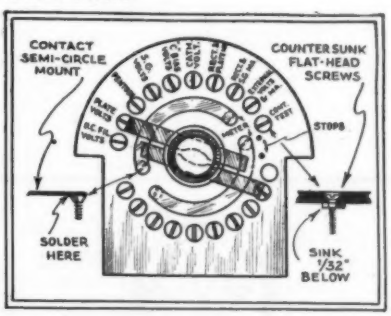


Fig. 3, above.
A home-made switch for that analyzer or tube tester is described.

Fig. 2, right.
Filling the electrolytic condenser with water, similar to a storage battery, may reduce hum.

MAKING A BI-POLAR SWITCH

By G. A. Sherwood

IN Fig. 3 is illustrated the manner in which the writer made a successful bi-polar analyzer switch to meet an immediate need.

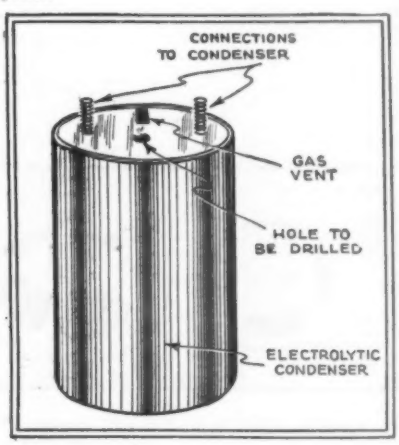
A series-parallel switch; two pieces of copper; a small piece of ⅛-inch bakelite; twenty, ½-inch, 6-32 flat-head brass screws; four, ⅜-in. flat-head brass screws; and two stop pins complete the list of parts. Countersink the flat-head screws 1/32-in. below the panel to prevent the contact blades shorting two of the contact points. The switch made by the writer and illustrated in the figure had over-all dimensions of 4¼ x 4¼ ins.

COMPENSATING CAR-RADIO BATTERY DISCHARGE

By L. Z. Tucker

THE writer, specializing in automotive radio service, has experienced difficulty in keeping car batteries correctly charged, due to the fact that the cars often make long runs, thus making it necessary to adjust the generator (normally adjusted to compensate for drain due to the radio set being operated on short runs) so as not to over-charge the battery.

To make generator adjustment easy for the car operator, I now mount on the cowl a 6-ohm rheostat of heavy-duty type and run two wires to the generator, as shown in Fig. 4. This resistor is connected into the circuit by breaking the field coil lead to the adjusting brush, completing the circuit through the variable resistor; then, advance the third, adjusting brush on the charger to its full limit. After this, the rheostat will afford a charging range of 2 to 20 amperes.



THE ANALYSIS OF RADIO RECEIVER SYMPTOMS

OPERATING NOTES

By BERTRAM M. FREED

THE servicing of radio receivers is becoming more complex with each succeeding day. Manufacturers are releasing new models so quickly that the Service Man has all he can do to keep abreast of the different tube designs, tube coupling devices, and automatic volume control systems. Each problem has its own solution and the only saving feature is the fact that the same faults are found in each model; the latest model receivers providing as much food for thought as their predecessors.

Philco 91

In a number of the new Philco 91 twin-speaker receivers, the complaint of "no control of tone" has been met. The same four-point switch as used in previous Philco models is employed here, as shown in Fig. 1. With the switch in the "brilliant" position, there are no condensers in the plate circuit of the first audio amplifier tube. As the switch is placed in the next position, a condenser is added to the circuit. The third and fourth positions each add a condenser, so that in the "bass" or "deep" position, there are three paralleled condensers connected from the plate of this stage to chassis. However, in these several receivers, as soon as the tone control was switched from "brilliant" to the next step, the tone became so muffled and deep and the volume dropped so much that it was impossible to understand any voice broadcast clearly.

In the first set where this condition was encountered, the tone control was thoroughly checked. Each condenser was tested separately and *each proved satisfactory!* A new tone control was then installed and when this did not help, the original unit was connected into another chassis and it performed in correct manner, creating a puzzle of "how come?"

Upon a happy thought, the Philco '37 tube (this receiver employs automobile-type, 6.3-volt tubes, discounting the '80 rectifier) in the first A.F. stage was replaced and the tone control worked splendidly. Despite the fact that the tube had been checked previously and found to be normal in all respects, it would not function in this stage; in the oscillator stage, it worked perfectly. To satisfy curiosity, the tube in question was placed in a number of receivers of the same model and the same result was obtained, "no control of tone." All in all, six of these tubes have been found. When this condition is encountered, all that is necessary is to switch the sockets of the oscillator and first A.F. tube.

After the Philco 91 has been in operation for a short time, the dial may be found to slip. This receiver makes use of a dial-cord whose tension, or slack, is taken up by a coil spring. It is only necessary to move this spring up one or two steps in the series of notches cut out in the dial to remove any slipping condition. Rosin, or warm tallow from an ordinary candle, rubbed into the cord will also help to eliminate this fault.

Philco 90

The Philco 90 receiver, in the pentode-output models (See the circuit on pg. 160 of the September, 1932, RADIO-CRAFT) has furnished grounds for much distress to many Service Men. The chief complaints with this model include not only fading or intermittent reception, as previously described, but also noisy reception. *Both* of these effects are caused by the A.F. coupling condensers.

This receiver utilizes three of these bakelite encased units: one, coupling the diode detector to the detector-amplifier tube; the second, coupling the plate of the detector-amplifier tube to the grid of the first A.F. tube; and the last, coupling the grid of the pentode to the plate of the first A.F. tube.

Their value is given as .01-mf., but .1-mf. tubular condensers have been used for replacement with much success when the former were not at hand, the value not being critical. In several cases, however, where the .1-mf. condensers were used, it was found necessary to replace the first A.F. grid-leak resistor and sometimes, the pentode grid-leak resistor, with 100,000-ohm units to eliminate the tendency to motor-boat.

In this same model, the second coupling condenser has been found mainly responsible for noisy reception, although, on

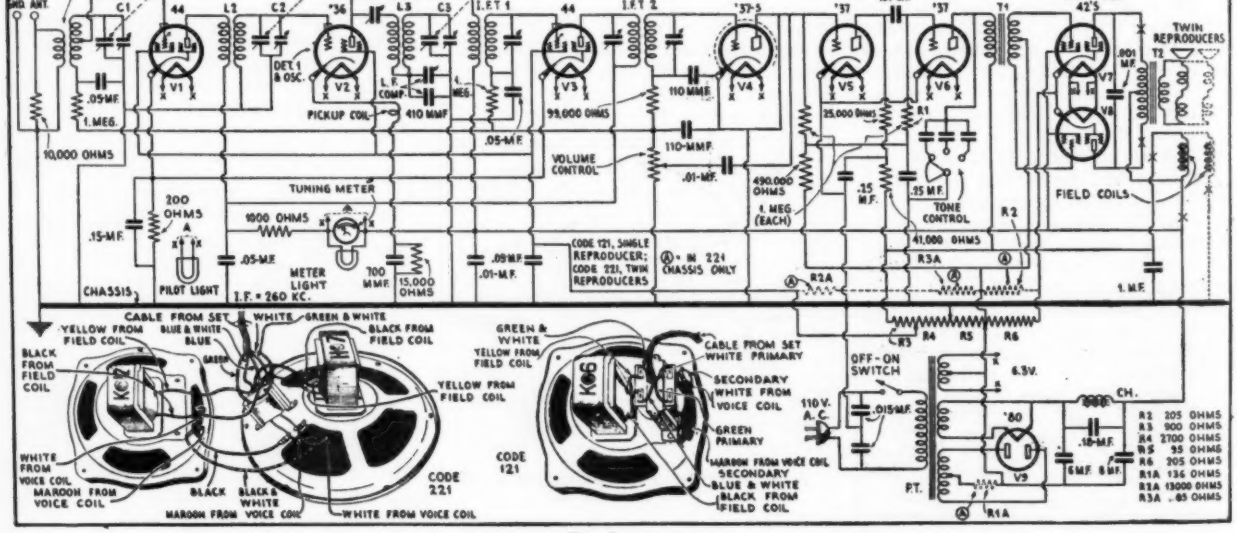


Fig. 1

Schematic circuit of the Philco 91 receiver. The volume control has a value of 350,000 ohms; the tone control has one .05-mf. and two .01-mf. condensers; the field coils, a resistance of 3,200 ohms; and the filter choke, a value of 285 ohms.

two occasions, the third was the offending member. This may be eliminated by applying the soldering iron to the lug of the condenser. Usually, it will be necessary to replace the unit. Any method may be pursued in determining the noisy condenser, although the writer has found that removing each one, in turn, from the circuit, and substituting a new condenser is the most certain and quickest procedure.

Zenith 91, 92

The next receiver to be considered is the Zenith 91, 92 superheterodyne; in chasses below Nos. 373, 334 and 301, 394, respectively, we experienced no end of trouble. The symptoms in this case, are: (a) marked fading; (b) no dip-action on the tuning meter, and; (c) little or no response on the local side of the local-distance switch. A glance at the schematic detail, Fig. 2, will disclose two resistors, R1, R2, which are connected in series across the D.C. voltage output as bleeder resistors, the values being respectively, 2,800 ohms and 3,600 ohms. These carbon units, not capable of withstanding the load placed upon them, are the source of the entire complaint.

These resistors break down under load and the electrical values vary, resulting in the above-mentioned complaint. The only remedy is replacement with resistors whose rating is 10 watts or more. The defective units are easily located, as they are the two largest resistors mounted upon the resistor strip. The other means of identification will be the discolored or burnt portions to be found on the resistors as well as a collection of dust under the units, such as is found under beds and dressers in a bedroom, when the floors have been neglected for a few days. (!)

It must be mentioned that the last lot of this model that was released incorporated wire-wound resistors of sufficient carrying capacity in place of the troublesome carbon units.

Majestic 61

On service calls marked "Majestic 61," a common complaint is "no reception," or "dead." A check will often reveal no plate voltage on the I.F. tube or first-detector and very low plate voltage, if any, on the other tubes.

The use of the ohmmeter will show an almost full short-circuit between the I.F. plate or first-detector plate—as the case may be—and chassis. Unsoldering wires will finally localize the trouble to the primary of either one of the two I.F. transformers. A low-range ohmmeter, capable of measuring the D.C. resistance of the I.F. primary, will show the "short" to chassis to be at the "B+" end of the primary. The circuit is Fig. 3.

This short is caused by the breakdown of the bypass condenser connected from the "B+" side of the primary to the shield can of the transformer. The condenser is sealed in pitch within the riveted shield can of the I.F. transformer, and before the defective unit may be removed from the circuit, it will be necessary to drill out the rivets (so that the shield can may be disassembled), and to melt out the bypass condenser. The transformer and shield are then re-assembled; a bypass condenser of sufficient voltage rating is then connected externally across the correct terminals.

No doubt, it is much simpler to replace the I.F. transformer, but this is a somewhat expensive proposition if a genuine replacement is desired, although an ordinary high-gain I.F. transformer, peaked at 175 kc. may be employed, together with an external bypass condenser.

If the entire shield is carefully insulated from the chassis and the transformer mounted on the chassis by means of insulated bushings or an insulated strap, this will disconnect the shorted condenser from the chassis. The secondary return of the first I.F. transformer does not connect to the shield can but connects to the A.V.C. circuit externally; the trimmer condensers have one side of each connected to the shield can and the other side connected to the plate and grid of the primary and secondary windings, respectively. The grounded sides of each condenser may be disconnected, and reconnected to the "B+" side of the primary and the grid return of the secondary, so that each trimmer is connected across and tunes each respective coil in conventional manner as depicted in Fig. 3.

Servicing this model, recently, two instances were encountered where no signals could be obtained unless the A.V.C. type '24 tube was withdrawn. When this was done, the receiver performed with full sensitivity but without any control of volume. The schematic showed the function of the A.V.C. tube to be that of varying the control-grid bias of the R.F., first detector and I.F. tubes.

As a similar experience with a Radiola 67 (described in the July, 1932 issue of RADIO-CRAFT, page 33, where an inoperative receiver would perform without the A.V.C. '27 tube) inspired further tests, the control-grid circuit of the A.V.C. '24 was checked.

The wire connecting to the control-grid of this tube runs through a hole in the chassis and is soldered to a small terminal on a square, insulated strip directly alongside the socket. A green wire, which should have been soldered to this lug was moving freely about in the joint

(*Continued on page* 251)

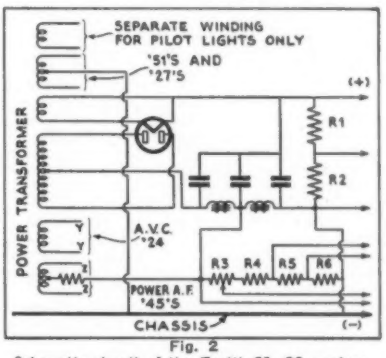


Fig. 2
Schematic circuit of the Zenith 91, 92 receiver, used in chasses Nos. 373, 334, and 301, 394.

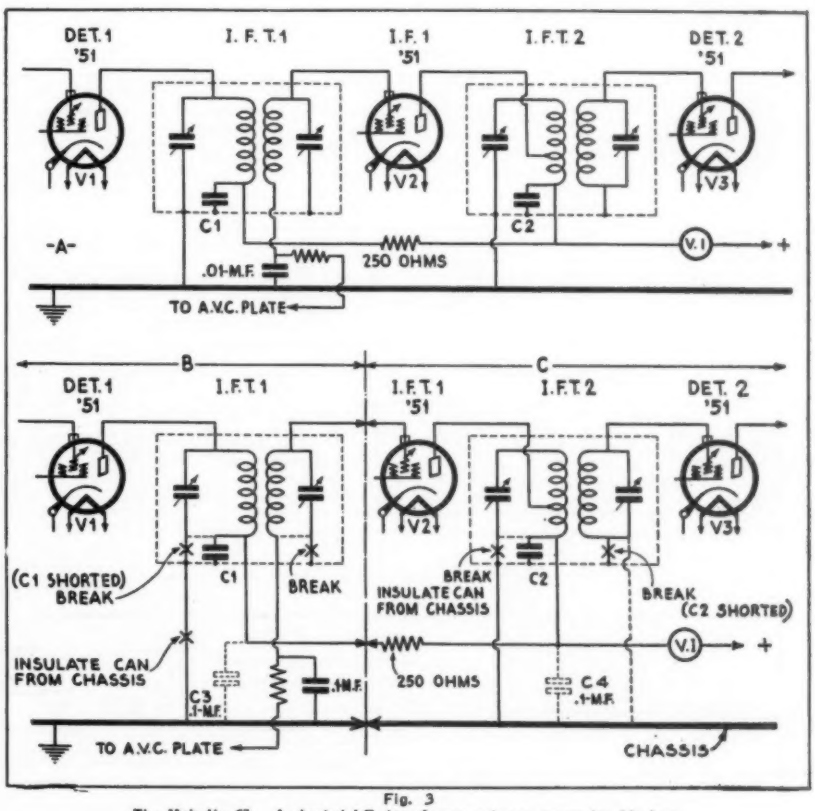


Fig. 3
The Majestic 61. A shorted I.F. transformer primary causes trouble here.

GALVIN MOTOROLA MODEL 61 AUTOMOTIVE RECEIVER

(Incorporating a type 85 duodiode-triode tube as a combination second-detector, automatic volume control and first A.F. amplifier.)

This automotive receiver incorporates the following tube combination: Tube V1, type '36 screen-grid as an R.F. amplifier; V2, type '39 variable-mu R.F. pentode as a combination oscillator and first-detector; V3, type '36 screen-grid I.F. amplifier; V4, type 85 duodiode-triode as a combination second-detector, automatic volume control and A.F. amplifier; V5, type 41 high-mu special automotive pentode second A.F.

High voltages are obtained from the storage battery of the car by means of an interrupter system and step-up transformer, in the manner described in the Sept. 1932 issue of RADIO-CRAFT, pg. 152. 152.

Following are the values of the components of this modern radio set: Condensers C1, C2, C3, tuning condenser gang; C4, C8, coupling condensers; C5, C6, C7, I.F. trimmers; C9, C13, C17 (buffer), .05-mf.; C10, 0.5-mf.; C11, 0.25-mf.; C12, 500 mmf.; C14, .002-mf.; C15, .01-mf.; C16, 1. mf.; C18, C19, 8 mf.; C20, C21, 0.1-mf.

Resistor R1, volume control potentiometer, 0.5-meg.; R2, R4, R7, R8, 0.1-meg.; R3, 75,000 ohms; R5, 5,000 ohms; R6, 500 ohms; R9, 0.2-meg.; R10, 50,000 ohms; R11, 0.7-meg.

It is not recommended that any repairs be made to a defective Elkonode (interrupter.) All such units should be returned to the factory (Galvin Mfg. Corp., Chicago, Ill.), or to the manufacturers of the Elkonode (see label on unit).

An open buffer condenser, C17, will be indicated by failure of the rectifier tube V6 to stay ionized. A purple glow in the tube is an indication of correct operation of this type of rectifier; a shorted C17 condenser will be indicated by spasmodic operation of the Elkonode, as well as failure of V6 to glow. As a general rule, when spasmodic operation of the Elkonode is observed it is an indication that the Elkonode is not feeding into the correct load; it either is underloaded or overloaded; two undesirable conditions.

After the Elkonode has been removed, it may be tested by applying 6 volts to the large terminals, with positive polarity to the brown wire; it is also necessary to connect a 5,000-ohm resistor across the red (or green) and black wires, together with an 8 mf. electrolytic condenser and a voltmeter. With this setup, the Elkonode should consume not more than 2.25 A.; the voltage drop across the 5,000-ohm load should be between 160 and 170 volts, provided the battery voltage is exactly 6.3.

The following precautions should be observed: Do not remove the receiver section of the set from the power pack, with the set turned on; the BR tube, V6, should not be removed from its socket unless the set is turned off. Since the "A" supply is polarized, it is necessary to make certain that the red wire connects to the positive terminal and the white wire to the negative terminal of the battery; do not operate the set with the "A" leads reversed, otherwise the Elkonode will be damaged beyond repair. For this reason, the polarity of the car battery should be double-checked by means of a voltmeter before the set is put into operation. (Reversed connection to the "A" battery will be indicated by low "B" voltage, spasmodic operation of the Elkonode and erratic flashing of V6.)

An ideal place for the reproducer is face-out, with the reproducer flush with the instrument board, but such a position is undesirable because of the space factor. Therefore, the first alternative is to leave it at the same level but to move it back to the bulk-head, for good operation.

If there is not room to mount the reproducer in that position, a second method is to face the reproducer toward the floor-boards, with the front-edge of the reproducer against the instrument board and the side against the side of the car, a position available in most cars.

It generally necessitates two holes being drilled through the instrument board to hold the reproducer and an additional bracket run from the adjacent side of the reproducer to the side of the car. This location is rather new to some installation men, but its acoustic properties are superior because of the additional baffle effect which results from the close proximity to the instrument board.

Following is a listing of the preferred location for automotive radio receivers to be used in the 1931 car models specified:

Ford, Model A: Motor compartment on the left side;

Chevrolet: Motor compartment on the left side or below the cowl on the right inside; (For 1932 model, remove carburetor and air cleaner, temporarily.);

Buick: Below the cowl on the right inside;

Chrysler: Right side of car under cowl;

Pontiac: Left side of motor compartment or right side of car under cowl;

DeSoto: Ditto;

Plymouth: Left motor compartment or right side of car, under cowl;

Cadillac: Ditto;

Lincoln: Center of motor compartment or right side of car, under cowl;

Packard: Ditto; (Light Eight, left side of motor compartment.);

Oakland, V8: Below cowl on right side or right side of car, under cowl;

Studebaker: Left motor compartment or right side of car, under cowl;

Oldsmobile: Right motor compartment or right side of car, under cowl;

Auburn: Right motor compartment or right side of car, under cowl.

(The "bulkhead" is the partition in the car which separates the motor compartment from the driver's compartment.)

Car manufacturers have furnished the following data regarding their provisions, in 1932 models, concerning automotive radio antennas:

Chrysler: Roof antenna with lead-in and provisions for "B" battery box;

Dodge: Ditto;

DeSoto: Ditto;

Plymouth: Ditto;

Reo: Equipped with roof antenna and lead-in;

Rockne: Ditto;

Studebaker: Ditto;

Buick: All models, $6.00 additional for antenna installation;

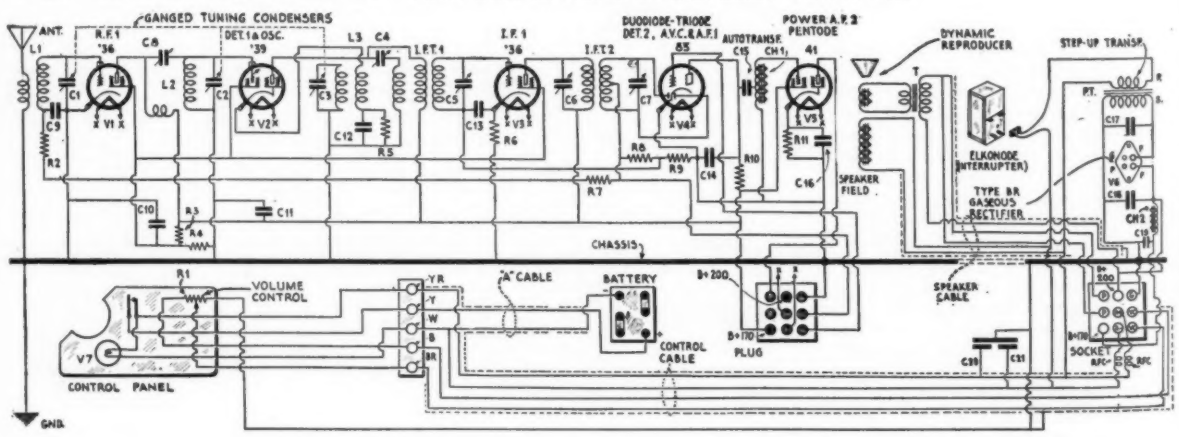
Franklin: Roof antenna but no lead-in;

Ford: Ditto;

Cunningham: All models, additional charge for antenna installation.

Check the proposed antenna for ground by means of a 0-200 V. meter, of 1,000 ohms-per-volt type, and a 200 V. battery. Even on damp days the leakage should not exceed 2 V. Lack of peak resonance on the antenna trimmer indicates a leaky antenna, or one having too great capacity; in general, an antenna screen area of about 9 sq. ft. will be satisfactory.

The dynamic reproducer may be checked for a rubbing voice coil by applying 50 V., 60 cycles, to the two outside or "B" terminals of the output transformer; instead of a clear, steady hum, a rubbing noise will be heard, if the voice coil is not floating entirely clear.



WELLS-GARDNER SERIES 062 AUTOMOTIVE SUPERHETERODYNE

(Incorporating types '36, '37, '38 and '39 tubes; also A.V.C.)

Numerous mail-order houses sponsor this chassis under their particular trade name; consequently, it is better remembered by its circuit connections than by the name of the chassis manufacturer.

In the No. 062 Series automotive receiver the following electrical values are used: Resistor R1, 0.5-meg.; R2, 7,000 ohms; R3, 1. meg.; R4, R13, 0.5-meg.; R5, R12, 0.1-meg.; R6, R7, R8, 2 megs.; R9, 1,000 ohms; R10, 20,000 ohms; R11, 15,000 ohms; R14, 350 ohms; R15, 1,300 ohms.

Condensers C1, C2, C3, equipped with trimmers, constitute the tuning gang; C4 is the oscillator padding condenser; C5, C6, C7, I.F. trimmers; C8, .006-mf.; C9, .02-mf.; C10, C14, 4 mf.; C11, .002-mf.; C12, C18, 05-mf.; C13, C15, C16, 0.1-mf.; C17, .001-mf.

Operating potentials are as follows: Filament potential, all tubes, 6 volts. Plate potential, V1, V3, 177 volts; V2, 173 volts; V4, zero; V5, 54 volts; V6, 159 volts. Screen-grid potential, V1, V3, 80 volts; V2, 76 volts; V4, zero; V5, 77 volts; V6, 165 volts. Control-grid potential, V1, V3, 3 volts; V2, 7 (depending upon dial setting); V4, zero; V5, 6 volts; V6, 15.5 volts. Plate current, normal, V1, V3, 3.6 ma.; V2, 0.9-ma. (depending upon dial setting); V4, zero; V5, 1.2 ma.; V6, 10 ma. All bias voltages must be read from cathode to ground.

Do not check the "A" and "B" potentials at the multi-point socket on the cable head, as the pilot light may be burned out when the switch is turned off. (This is due to the high inductance of the speaker field, which will increase the voltage at the break of the circuit.) Also, when the cable head and multi-point socket are taken off, the connections between the chassis and power unit are open so that readings are not made under load conditions.

To read the voltages at the sockets the chassis box, in most cases, will have to be removed from its mounting. In some instances, the cables, which may be attached to the dash or at other points, will have to be removed. The voltages can be read at the sockets with a long plug or with a pair of long, insulated test prods.

All tubes must be inserted and all units connected. A signal will effect the control voltages on V1, V3 and V5. If signals are received, ground the antenna and remove V4 to take the other readings.

The diode current establishes a drop across a resistor network; this potential is used as an additional bias potential on V1, V3 and V5, giving A.V.C. action.

The full control potential is applied to V1, two-thirds to V3, and one-third to V5. As the signal increases in intensity, the applied control potential is increased, thus giving uniform output as set by the manual volume control. The manual volume control varies the diode A.F. potential applied to V5.

The tone control is mounted on the dynamic reproducer, which derives its field supply from the storage battery in the car.

Aerials are preferred in the following order: Roof antenna; tape antenna on roof (this is not very effective where a grounded chicken-wire mesh remains inside the roof); plate antenna (strung underneath the running board); under-car antenna (A wire fastened from the right side of the rear axle to the lowest point under the motor, then back to the left rear axle, forming a V. At the vertex of the V is a spring to take up the slack. The lead-in is brought up from the vertex end.).

Do not turn the set on until all the wiring connections of the installation have been completed—this is extremely important.

To adjust the antenna trimmer, tune in a weak signal at the high frequency end of the dial with the manual volume control about three-quarters on. On one end of the chassis box is a small metal plate. Remove the two screws holding this plate. Directly under the hole in the chassis box is the antenna trimmer condenser screw. Adjust this up or down until maximum output is obtained.

Noise, in some instances, may be due to weak pickup caused by a poor antenna. The action of the automatic volume control, due to the low pickup, causes the set to operate at maximum sensitivity, thereby increasing noisy reception, due both to external pickup and internal conditions.

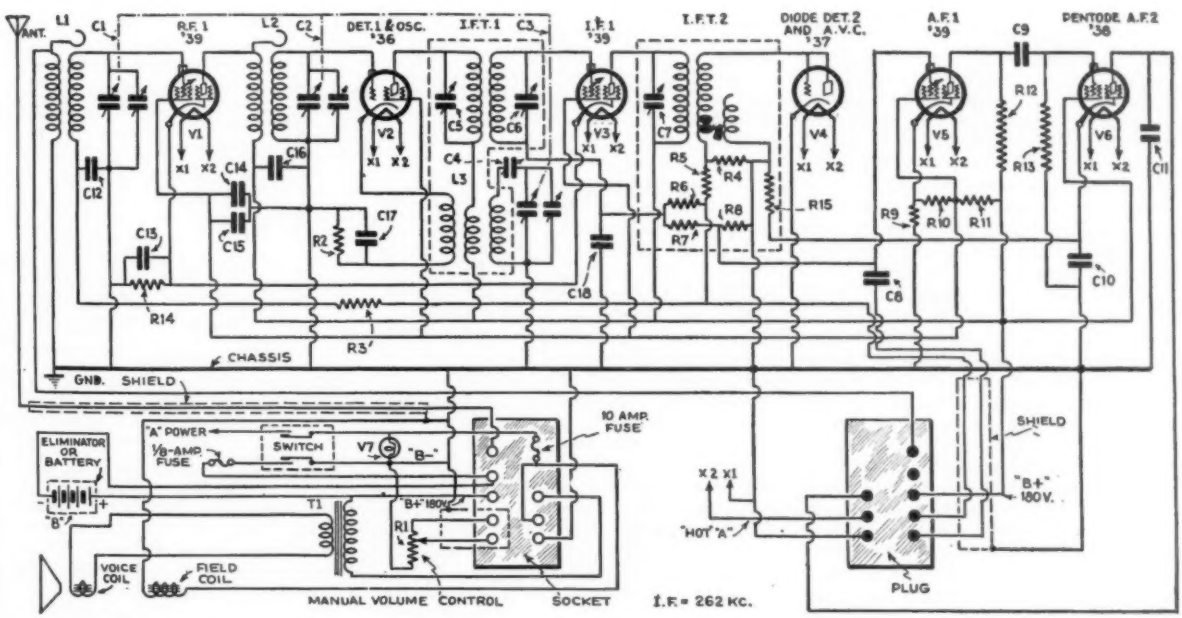
Noisy operation is also caused in some instances by loose parts in the car body or frame. These loose parts rubbing together affect the grounding and cause noises, due to the rubbing or wiping action. Tightening up the frame and body at all points and, in some cases, the use of copper jumpers, will eliminate noise of this nature.

Loose lamps or wiring are also a common cause of noisy operation.

Compensate for the increased battery drain occasioned by the radio set, by increasing the charging rate of the car generator. After a week of car operation, check the condition of the battery and re-adjust the charging rate accordingly.

Two fuses are used on this receiver. The 10 A. "A" fuse is located on the multi-point socket; the 1/8-A. "B" fuse, inside the control unit. The pilot lamp is a standard 6 volt No. 40 unit.

Poor operation may be due to moisture, from car-washings or storms, seeping into undesired places.



RADIO-CRAFT KINKS

Practical Hints From Experimenters' Private Laboratories

Prize Award

SELF-VULCANIZED RUBBER IN RADIO WORK

By G. L. Garvin

IN Fig. 1 are illustrated several ways in which the experimenter may use self-vulcanizing, tire-repair rubber in his radio work.

A twenty-five cent can contains sufficient material to make about 84 sq. ins. of rubber.

Out of this I am able to make rubber "spaghetti," rubber grommets, test-prod and pliers insulation, and insulating plates of any size.

Glue is not used—the secret lies in tightly pinching the surfaces together with pliers, then cutting along the seam with scissors to make a neat-appearing job. This rubber juncture cannot be separated—the rubber will tear elsewhere first.

REMOTE-RADIO OPERATION

By W. B. Matthews

IN Fig. 2 are shown the circuits of a system which I have set up between my laboratory and my regular household set which is located in another part of the building. The circuit was first designed in order that I might make use of the rather powerful A.F. amplifier with which the set is equipped without either moving the set to the lab., or impairing its value for ordinary use.

By plugging the output of a short-wave tuner, phonograph pickup, microphone transformer, or any other A.F. source into jack A, it may be fed into the detector or A.F. stage of the set. The jack contacts close the circuit from the battery, operating the relay (Rel. 1) which opens the last R.F. secondary winding and applies the input to the grid of the detector. In this way, short-wave programs may be picked up in the lab. and be heard from the regular set speaker. If a plug is inserted into jack B, the relay (Rel. 2) is operated and the output of the set switched back to the lab. to a speaker connected to the plug in jack B. If the plug is removed from jack A the armature of relay (Rel. 1) falls back and the regular broadcast program may be heard from the speaker in the lab. The set can be turned on or off by operating the switch S which controls the power-control relay.

The jacks and the control switch are mounted in a panel along with other jacks and equipment located in the lab. The relays (Rel. 1 and Rel. 2) are mounted in a small box located immediately back of the chassis in the radio cabinet, as is the power control relay.

$5 for a Practical Radio Kink

As an incentive toward obtaining radio hints and experimental short-cuts, "Radio-Craft" will pay $5.00 for the best one submitted each month. Checks will be mailed upon publication of the article.

The judges are the editors of "Radio-Craft" and their decisions are final. No unused manuscripts are returned.

Follow these simple rules: Write, or preferably type, on one side of the sheet, giving a clear description of the best radio "kink" you know of. Simple sketches in free-hand are satisfactory, as long as they explain the idea. You can send in as many kinks as you wish. Everyone is eligible for the prize except employees of "Radio-Craft" and their families.

This contest closes on the 15th of every month, by which time all the Kinks must be received for the next month.

Send all contributions to Editor, Kinks Department, c-o "Radio-Craft," 98 Park Place, New York City.

For the connection between the two locations, I use a small cable, but on a long run it may be necessary to use a separate pair for the output of the set in order to prevent capacity feedback into the input pair. Also, it is necessary to keep the capacity between the two sides of the output pair to as low a figure as possible to prevent distortion.

Now for a few notes about the apparatus. All three of the relays must be wound to suit the voltage of the battery supply. If they are properly designed their current consumption can be kept quite low. For the power-control relay, I rewound an old automobile generator relay which I found had large enough contacts to carry the set load. These contacts are connected directly across the regular set switch. For Rel. 1 and Rel. 2 any relays with suitable contact arrangements will do. The contacts of Rel. 1, which are normally closed, are cut into the detector control-grid circuit. Care must be taken to make sure that the input is cut into the grounded end of the R.F. secondary circuit so that the receiver's tuned circuits will not be thrown out of balance. Slightly different arrangements may be necessary for different sets at this point, or it may be wired so as to apply the input to the first A.F. stage. In any case, it is connected the same as a phonograph pickup would be. Any type jacks with an extra pair of contacts may be used as shown in the diagram.

Many interesting uses of this or similar arrangements will suggest themselves to the experimenter.

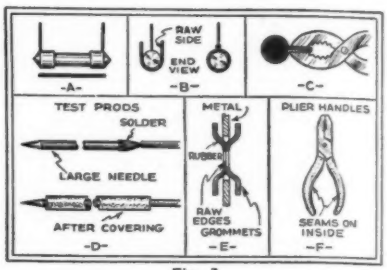


Fig. 1

Here's real efficient insulation.

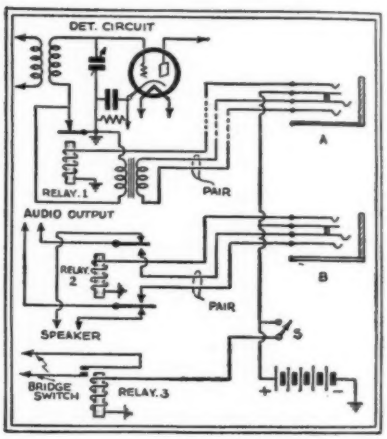


Fig. 2

You can make extra money with this idea.

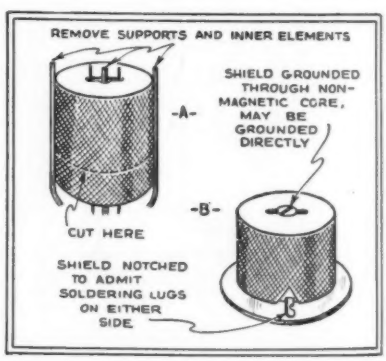

REMOVE SUPPORTS AND INNER ELEMENTS
SHIELD GROUNDED THROUGH NON-MAGNETIC CORE, MAY BE GROUNDED DIRECTLY
-A-
-B-
CUT HERE
SHIELD NOTCHED TO ADMIT SOLDERING LUGS ON EITHER SIDE

Fig. 3

Shielding R.F. coils—an important thing.

SHIELDING FOR SMALL COILS

By Wm. R. Thames

THE writer has successfully used screen-grids from old '22 and '32 tubes as shield cans for small R.F. choke coils. Perhaps this idea will interest some of the readers of RADIO-CRAFT.

The details of this arrangement are illustrated in Fig. 3, at A and B.

THE RADIO CRAFTSMAN'S PAGE

The Bulletin Board for Our Experimental Readers

"THE RADIO 'TREASURE' FINDER"

Editor, RADIO-CRAFT:

In the June issue of RADIO-CRAFT, on pg. 716, you give the construction details concerning a "treasure" finder. I had one built just the way the article specified, using the parts that were supposed to be used, but the resulting machine will only work a little bit. It seems to me that it is *too* short-wave. In other words, it will work only when it is setting on metal—*It had to be setting right on the metal.*

What could be the matter with it? I sure would like to get it working as I have spent quite a lot of money on it. If there is anything you can give me light on I sure would appreciate it.

R. L. RIDDLE,
The Kiva,
Santa Rosa, N. M.

(Below is the comment of the author, Mr. Clyde J. Fitch. Note that a slight change in the wiring is necessary to obtain satisfactory results from this practical metal locator; this circuit modification is illustrated in these columns.—*Technical Editor.*)

CONCERNING WHICH MR. FITCH REPLIES —

Editor, RADIO-CRAFT:

The treasure finder described in the June 1932 issue of RADIO-CRAFT has aroused so much interest that the editors have literally been swamped with inquiries—and to answer these in a group—this means has been taken so that others, who may be interested, will not be confronted with the same difficulties.

It is unfortunate that almost all who have built the apparatus are disappointed with the results (the letter of Mr. R. L. Riddle is a good example). In many instances we believe this is due to the fact that an error was made in reproducing the diagram of the *receiver* (Fig. 1 on page 749) and therefore *we are publishing the corrected diagram herewith.* We were not aware of this error until quite recently and were at a loss to know why so many could not get the apparatus to work.

In the original diagram, the plate supply for tubes V1 and V2 was taken from the plate terminal of V3, instead of the plus "B" lead; this caused the plate current of all three tubes to pass through the headset. Under this condition the receiver would not work as well as it should. *The corrected diagram published herewith must be followed in building the instrument.*

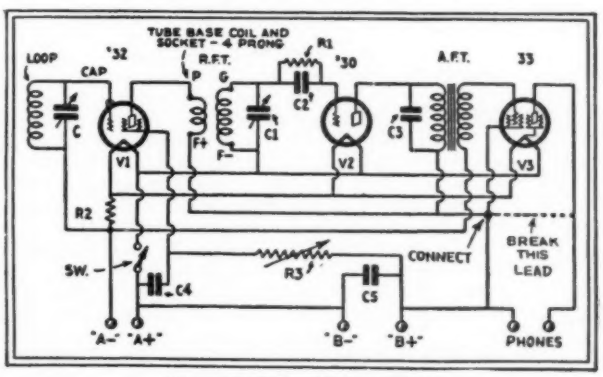
The apparatus as built by the writer was tested by placing a piece of metal about one foot square near the receiver loop, and while a perfect balance of *zero sound could* not be obtained by adjusting the angle of the loops, *the presence of the metal could be detected by means of a change in the quality of the sound.* (There is, of course, a change in *volume*, but it is very slight and the ear is not very sensitive to slight changes in volume.) The metal sheet could be detected at a distance of four feet from the receiver loop; a typewriter was detected at a distance of six feet.

Some who have used similar apparatus state that it works better outdoors during rainy weather. The reason for this is not definitely known. Perhaps the buried objects, having better electrical contact with moist earth, produce a greater effect. In one case that came to our notice a couple of swords were unearthed; they were about four feet below the surface.

One writer states that he could get a good balance of zero sound indoors, but could get no zero balance outdoors. Perhaps body capacity between his head and the earphones unbalanced the apparatus when his feet were in contact with the ground. Wearing rubbers should prevent this if this is the cause of the trouble.

Check *all* potentials with an accurate, high-resistance voltmeter. Make certain that grid-leak resistors R1 in Fig. 1 and R in Fig. 4 have the correct values. Try several tubes in their respective positions; this admonition holds true particularly for V1 in Fig. 1—and to nearly the same degree, V3 in the same figure. Try reversing the connections to the primary of the R.F. transformer, R.F.T., in Fig. 1. Test all of the condensers for short-circuit; also, if possible, check them for their rated capacities—or, at least, for open-circuit. Note that unless resistor R in Fig. 4 has the correct value it will not be possible to obtain a satisfactory audio modulation note from the transmitter; in fact, it may not be possible to obtain any audible modulation of this section of the metal-locator.

CLYDE J. FITCH.



"Treasure" Finder circuit correction.

IMPORTANT NOTICE

In the interest of those readers who do not like to mutilate this magazine, we have asked our advertisers not to place coupons in their advertisements.

Instead of the usual coupons, you will find a number of convenient post cards inserted between the last page and the back cover of this magazine.

This new service will save you time and work. No need to cut coupons, nor is it necessary to hunt for and address envelopes. Moreover, the space for your name on a coupon is usually so small that the advertiser is often not able to make out your writing and then you wonder why you do not get the literature sent for.

Then, last but not least—the postage for a postal card is only 1c whereas a letter now costs 3c.

Read the advertisements and then turn to the page containing the special postal cards. Detach, fill out and mail the card of the advertiser whose literature or offers you want to have sent to you.

Mail your card today! Show the advertisers that you appreciate their cooperation and thoughtfulness.

MORE TUBES? (!)

Editor, RADIO-CRAFT:

As one who has been interested in radio for many years (I was one of the subscribers to the splendid magazine "Amateur Work" at the time it passed and recall that your own "Modern Electrics" was sent to subscribers to the first-named publication by way of completing the periods for which they had paid their respective subscriptions to "Amateur Work"),

(*continued on page* 238)

A PORTABLE LABORATORY

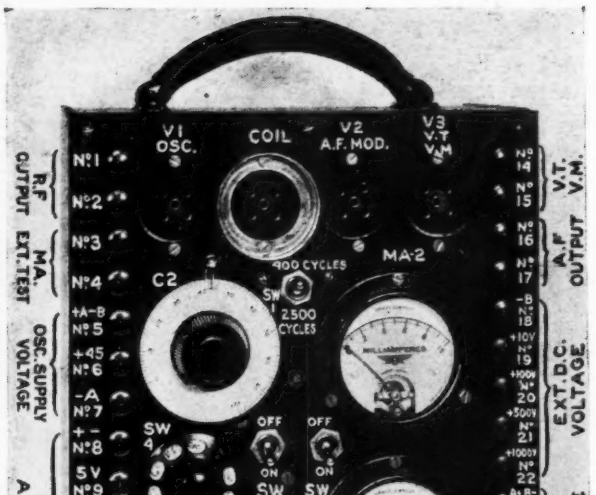


Fig. A
Front view of the very versatile short-wave tester.

● For some unaccountable reason, short-wave receivers do not seem to enjoy the same up-to-date testing that is being given to broadcast receivers. That this may be due, in part, to lack of efficient test equipment seems to be a generally accepted fact. To overcome some of the difficulties involved in testing short-wave apparatus, the author of this interesting article has designed and built the portable testing laboratory described in this article.

This device consists of a vacuum-tube voltmeter; a short-wave oscillator, which may be made to include the broadcast band by the addition of the proper coils; an audio oscillator, which may also modulate the R.F. oscillator if so desired; continuity tester; capacity tester; and may also be used for the measurement of inductance. Aside from the above, the instrument may also be used for an ohmmeter; as an external voltmeter (either A.C. or D.C.) and as a D.C. milliammeter.

The entire unit may be built in a box 9¼ x 10¾ x 3¾ inches.

SOMEHOW, to a large number of people, the idea of anything connected with the word "laboratory" immediately suggests expensive and delicate apparatus which is far beyond their financial and technical means. There is no basis for this state of mind. Good standards can be built inexpensively, ruggedly, and quite accurately by almost anyone. The ultimate test of an instrument is not always cost, but its practical workability. The simplicity of a circuit, the ease with which it can be built and used, the care with which it must be made are the factors that should be kept in mind when building any piece of equipment.

Expensive parts and "trick" circuits can never equal good workmanship and proven design. The test laboratory to be described in this issue was built to these ideals. It will perform equally as well for anyone, so long as care is taken in construction and adherence to the values given. This unit, Figs. A and B, was built specifically for one of the smaller manufacturers by whom it has been used both in the shop and in the field.

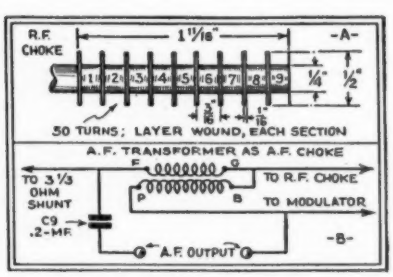


Fig. 2
At A, details of the choke; at B, method of connecting T1.

The Variable R.F. Oscillator

Keeping in mind that the oscillator must be as simple and trouble proof as possible, the features emphasized in the design were:

(1) One coil in the oscillator circuit. This would automatically eliminate bothersome feed-back arrangements and simplify the wiring;

(2) The circuit must be absolutely dependable and must oscillate easily, since the maximum available "B" supply is but 45 volts;

(3) Since the oscillator is to be calibrated, the frequency should be spread over the entire dial, as the error in calibrating and reading is less.

The circuit that was finally adopted is a modification of the Colpitt circuit shown in Fig. 1. The stability of this circuit depends upon both the value of the grid-leak R1, and more particularly on the size of the R.F. choke, R.F.C. It was found that with values of R1 varying from 5,000 to 100,000 ohms, the circuit oscillated sharply, but only at certain points. Using a 10,000-ohm leak, the oscillations centered about 4,000 kc.; a 100,000-ohm leak brought the circuit oscillations down to 2,500 kc.; but the tube refused to oscillate above 5,000 kc. A final value of 2 megohms was selected as giving the best output for the different frequencies.

The same trouble was encountered in the selection of an R.F. choke. While a choke may seem to be excellent at one frequency, it may be very poor at another because of either too large a distributed capacity or, if this is not an objectionable feature, too litle inductance. It was necessary to make a choke having high enough inductance at the lowest working frequency and negligible distributed capacity at the highest frequency. The final choke was a nine section layer-wound affair having very low distributed capacity and an inductance of 30 millihenries. This is illustrated in Fig. 2A.

The "band spread" effect was accomplished by having a fixed 60 mmf. condenser, C1, in parallel with the variable 50 mmf. condenser C2. This arrangement necessitates the use of more plug-in coils than are usually needed to cover the desired frequency band. However, the advantage of this method of tuning fully compensates for the labor involved in winding the coils. The economy of the circuit is shown by the fact that not more than 2 ma. is drawn by the plate of the tube at any frequency.

Modulation of the oscillator is accomplished by supplying a pulsating voltage to the plate of the R.F. oscillator tube V1. The plate lead of the R.F. oscillator connects to a tap on the A.F. choke, A.F. 1, in which is present the A.F. current due to the A.F. oscillations generated by the modulator tube V2. The oscillator plate voltage is then modulated by the A.F. voltage drop across the portion "a" of A.F.1, and hence the R.F. output is modulated.

The output of the oscillator is taken through L1, C3 and R2 which form a "dummy" antenna. The output is coupled directly to the receiver under test. Values for L1 and L2 are given in Table I.

The A.F. Oscillator

There is nothing difficult in the building of an oscillator to generate audio frequencies. The requirements are an A.F. transformer, a socket, a few condensers, a grid leak, and an A.F. choke. The transformer T1 has a turns ratio of 3:1; the primary is connected to the grid and filament of the tube. The secondary corresponds to the usual "tickler

SHORT-WAVE

By H. HARRISON

coil" which provides the feed back from the plate to the grid. The precaution here is to reverse the secondary leads if the tube does not oscillate.

The chosen frequencies of 400 and 2,500 cycles are obtained by switching either to a .02- or .005-mf. condenser respectively, by throwing SW1 to position 1 or 2. The A.F. output may be secured separately by connecting leads to the terminals marked "A.F. output." The common coupling choke A.F.1 may be the secondary of a standard input push-pull transformer, or a 6:1 A.F. transformer with the primary and secondary windings connected in series. The connection should then be as shown in Fig. 2B. Switch SW2 controls the A.F. into the R.F. oscillator.

Vacuum Tube Voltmeter

The vacuum-tube voltmeter used in the unit is simple to operate. The input terminals must be connected through a continuous-current-carrying circuit, which may be either a resistor or coil in parallel with the input terminals. If there is no such input circuit, the input terminals should be connected to a resistor of 1. megohm. The plate milliammeter should also be properly adjusted before applying the input voltage to be measured. To make these measurements, it is necessary to connect the terminals under test (assume it to be from the output winding of a transformer) to the input terminals of the V. T. voltmeter. Without operating the amplifier, that is, so there is no alternating signal, adjust the plate-current reading (MA2) to the smallest readable value. This adjustment is made by varying the potentiometer R3. Note the plate-current reading and the grid-voltage reading on the meter (MA1) across the potentiometer. Now start the amplifier so that there is alternating voltage across the output winding. The plate current meter reading is changed. Keeping the signal voltage on, adjust the potentiometer until the original plate current reading is obtained; read the grid voltage. Subtract the original grid voltage reading from the new reading. Their difference is approximately equal to the peak input voltage. Divide this reading by 1.41 or multiply by .707 to obtain the R.M.S. (or effective) value. During the test, SW3 must be closed.

The desired range of voltage measurements may be obtained by selecting the proper tap on switch SW4. Multipliers R4, R5, R6, and R7 have values of 1,000,

(*Continued on page 248*)

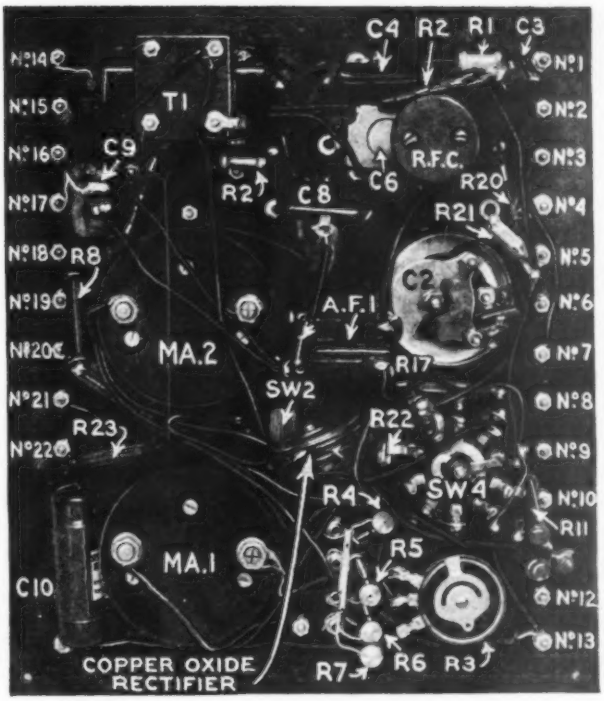


Fig. B

TABLE 1

Frequency in kc.	No. of turns of L1	No. of turns of L2	Wire size D.S.C.
1,600 2,660	30	17	26
2,660 3,250	20	13	29
3,250 4,160	20	10	26
4,160 5,350	18	9	26
5,350 6,460	18	9	24
6,460 7,180	13	7	26
7,180 8,300	9	4	24
8,100 10,000	10	7	22
10,000 11,900	13	6	20
11,900 12,800	10	4	20
12,810 13,680	7	3	20
13,680 14,220	5	2	22

NOTE: Coils are not lacquered; are wound on standard UX forms, 1-1/8 inches in diameter; L1 and L2 are spaced 1/8 to 1/4 inch. See text for further details.

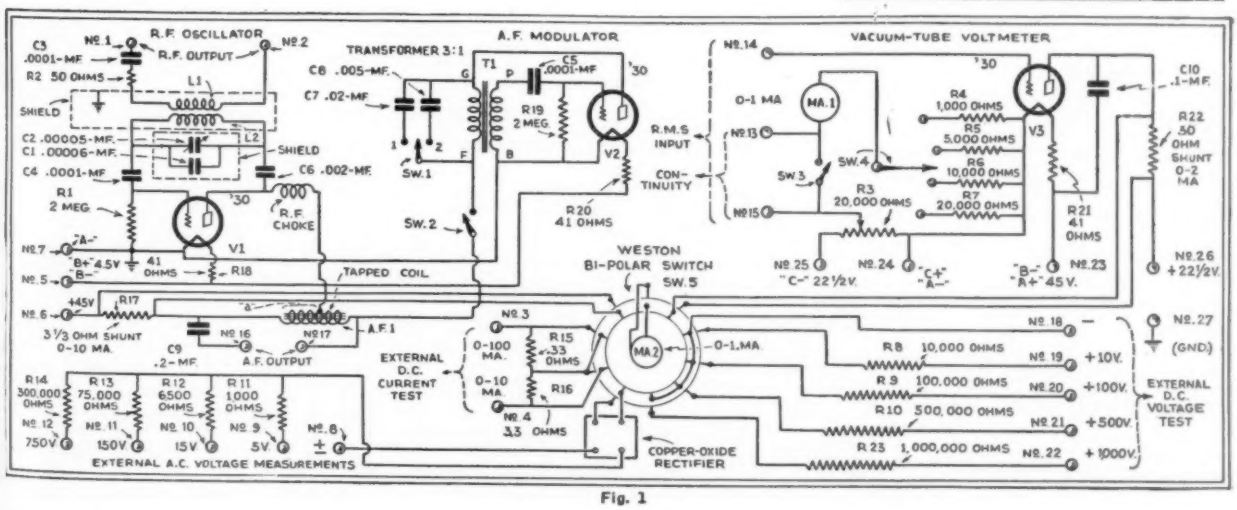


Fig. 1

RADIO-CRAFT'S INFORMATION BUREAU

SPECIAL NOTICE TO CORRESPONDENTS: Ask as many questions as you like, but please observe these rules:

Furnish sufficient information, and draw a careful diagram when needed, to explain your meaning; use only one side of the paper. List each question.

Those questions which are found to represent the greatest general interest will be published here, to the extent that space permits. At least five weeks must elapse between the receipt of a question and the appearance of its answer here.

Replies, magazines, etc., cannot be sent C. O. D.

Inquiries can be answered by mail only when accompanied by 25 cents (stamps) for each separate question.

Other inquiries should be marked "For Publication," to avoid misunderstanding.

SOUND-CONTROLLED RELAY

(173) Mr. H. Young, Allentown, Pa.

(Q. 1.) Please print the schematic circuit of an amplifier system suitable for operating a relay by whistling or hand clapping.

(A. 1.) By courtesy of Good News we reproduce in Fig. Q. 173 a schematic circuit of this nature.

The following parts were specified in the original layout:

One single-button microphone;
One microphone battery (two large-size flashlight cells);
One microphone transformer with a ratio of 60 to 1, approx.;
One A.F. transformer, ratio 4 to 1, approx.;
Three sockets (two type UY and one type UX);
One power transformer having a 600 volt secondary, center-tapped, one 2.5 volt winding and one 5 volt winding;
Two snap switches;
Two condensers, 2 mf., 400 V., D.C.;
One choke, 30 hy., 40 ma. rating;
One potentiometer, 10,000 ohms, 40 ma. rating;
One potentiometer, 1,000 ohms, 40 ma. rating.
One resistor, 5,000 ohms, 20 watt rating;
One resistor, 4 watts, 1,500 ohms;
One resistor, 4 watts, 1,000 ohms;
One resistor, 4 watts, 200 ohms;
One resistor, center-tapped, 30 ohms;
Two Western Electric, Type B-26 holding relays, (G. E. Type CR-2810-1245A);
One G. E. Master Relay, type CF-2930-1080-C2;
Three Radiotrons (two type '27; one type '80).

The relays are so adjusted that a low intensity of sound will cause one relay to act. This in turn causes a power switch to close.

The second sensitive relay breaks the current through the interlocking power switch and opens the 110 V., A.C. circuit.

Adjust the slider on the 10,000 ohm resistor until the drop across the 5,000 ohm resistor is about 135 volts, then adjust the slider of the 1,000 ohm potentiometer to obtain a bias potential within 4.5 to 22 volts on the control-grid of the second-stage '27. Make provision for reading the plate current of either tube during adjustments. With the microphone out of the circuit, the plate current of the first '27 should be about 5 ma.; the second then should read about 1. ma. The latter value should increase to about 5 ma. when the microphone is connected into circuit and actuated.

Relay No. 1 is adjusted to close at about 2 ma.; No. 2, at about 4 ma. The latter value is obtained by varying the control-grid bias of the second tube until the plate current is 2 ma.; then 4 ma. After adjustment at the first and then second values, reduce the plate current to 1 ma.

An arrangement of this nature, which will light an illuminated sign at a whistle, and put it out at the clap of hands, is a good money-maker as a novelty attraction.

7-PRONG TUBE SOCKETS

(174) Mrs. L. G. Jensen, Weehawken, N. J.

(Q. 1.) Is there an approved construction design for bases and sockets designed for 7-prong tubes?

(A. 1.) In Fig. Q. 174 are illustrated the base and socket designs approved by the Radio Manufacturers Association. The original illustration was made available by courtesy of Hygrade Sylvania Corp.

MODERNIZING THE FLEWELLING SIMPLIFIED SUPER-REGENERATOR

(175) Mr. John Smith, Chillicothe, Ohio.

(Q. 1.) What would be the circuit arrangement for the old Flewelling "simplified" super-regenerative receiver as redesigned to use the newer types of tubes—for instance, the screen-grid?

(A.1) We have not had any practical experience with this interesting circuit modified in the manner suggested. However, we reproduce in Fig. Q. 175 the schematic circuit proposed by Mr. Ricard West, writing in a recent issue of the radio section of the New York *Sun*.

A standard 3-circuit tuner, L, is used. The most important portion of the circuit is the grid leak, the value of which must be varied until satisfactory operation is secured with the particular tube used. Note that the .006-mf. fixed condenser must be of mica-insulation type. Coupling transformer T must have a very high-impedance primary. Position 1 of Sw. is the usual connection; position 2, super-regeneration.

We will be interested to hear from experimenters who may try out this circuit arrangement.

"MOBILE P. A. SYSTEM"

(176) Mr. R. Costabile, Chicago, Ill.

(Q. 1.) In the August issue of RADIO-CRAFT is described an interesting public address system amplifier consisting of a type '24 tube feeding a pair of '50's in push-pull; a direct-coupled circuit arrangement is used.

Please furnish the following data regarding components of this amplifier: Rating of power transformer; type of resistors used, and; method of connecting more than one reproducer into the circuit.

(A. 1.) The power transformer used in this powerful, high-fidelity public address system amplifier is rated at 200 watts; the high-voltage secondary delivers 750 volts each side of the center-tap. Note the use of separate windings for each of the filaments (this is necessary to obtain correct bias potentials).

The two 0.125-meg. and the two 0.1-meg. resistors are rated at 1. watt; the remaining resistor units are rated at 75 watts (in order to accurately maintain correct values under load).

Additional reproducers will require the use of individual output transformers.

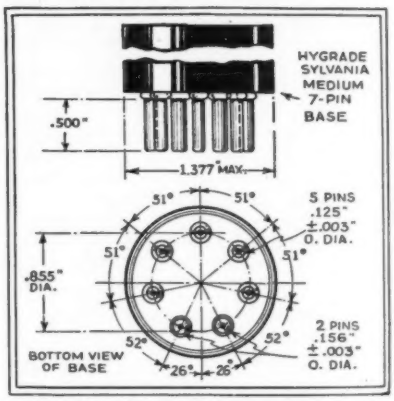


Fig. Q-174. An approved 7-prong socket design.

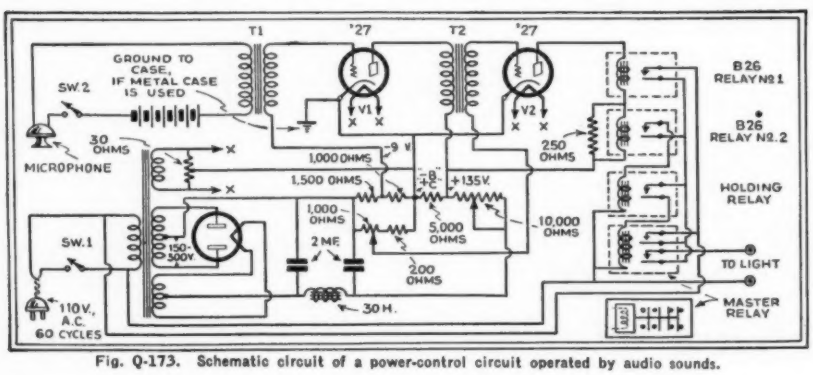


Fig. Q-173. Schematic circuit of a power-control circuit operated by audio sounds.

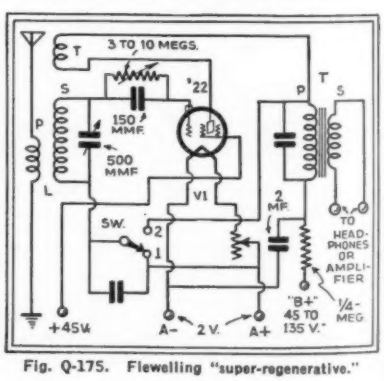


Fig. Q-175. Flewelling "super-regenerative."

SERVICING NEW RECEIVERS

By FLOYD FAUSETT*

IN order to increase his income, the active radioman must keep himself equipped to render better service to more customers. The radioman who fully appreciates this fundamental factor of profitable radio servicing is interested in testing equipment which will assure less time-per-job for service which is better than the service rendered by his competitors.

Statistics recently compiled by RADIO-CRAFT indicate that only about 9% of the owners of manufactured radio tube socket testers are equipped with manufactured oscillators, and it is probable that most of the oscillators included in this 9% do not provide the necessary frequencies for all the newer superheterodyne radios. It may, then, be conservatively concluded that not more than 9% of the owners of radio analyzers are equipped for performing better service with less time-per-job. Those of the 9% classification ought to complain less about cut-rate competition and let the public know of their better facilities for rendering better service.

Many radiomen rightfully consider themselves as professional men, but professional pride or ethics should not prevent their resorting to other than very modest advertising. For instance, chiropractic doctors believe in advertising their training and clinical facilities, and why should "radio doctors" be less modest? Is there anything wrong with a radioman's advertising that his is the only local service establishment equipped for meter adjustments of all types of superheterodyne radios when his cut-rate competitors are not so equipped?

The Supreme 56 Analyzer

The newer types of radios with R.F. pentodes, class A and class B amplification, muting circuits, etc., will introduce service complications which make it necessary that radiomen keep up with the rapid developments of the new radio technique, and keep his equipment up-to-date. In adding to his equipment, he must be interested in the equipment which is designed with the least likelihood of obsolescence. The manufacturers of test equipment are endeavoring to design apparatus which will not become obsolete in a short time. Most of the obsolescence is due to the limitations imposed by the switching arrangements employed. Furthermore, the moving parts of complicated switches constitute a possible source of trouble.

The Supreme Model 56 is an example of an analyzer which eliminates these disadvantages by eliminating the use of rotary and push button switches for connecting the meter to the various circuits. Each range of the meter, and each extended radio circuit has its own set of insulated terminals, and any desired connection can be conveniently made by means of the test leads provided with the analyzer. The cathode circuit, the filament circuit, the top cathode connection of the new 5-pin Wunderlich tube, or any other required circuit may be taken as the reference circuit for voltage measurements.

In the usual analyzer with its complicated switching arrangements, a two-way switch provides the possibility of only two reference circuits; that is, cathode and negative filament. Heretofore, analyzers have provided means for connecting a self-contained 4.5-volt battery into the (1) control grid and (2) the normal grid circuits for tube testing. The latest 6-pin and 7-pin tubes have introduced complications which may require battery connections into any one of five tube circuits for tube tests. In the Model 56, the battery may be connected into any one of the six grid and plate circuits.

Although the Supreme Models 70 and 60 oscillators were designed before superheterodyne radios became the rule rather than an exception, there has not been a single superheterodyne designed with an intermediate frequency which cannot be accommodated by the original Supreme oscillators which pioneered the harmonic principle of commercial oscillator tuning.

Since many radiomen already have hundreds of dollars invested in testing equipment, they are anxious to have such equipment re-constructed to meet the testing requirements of the new 6-pin and 7-pin tubes and circuits. Arrangements are being made for such remodeling of older Supreme products. In addition, new ohmmeters, volt-ohmmeters, resistance decade boxes, capacitor leakage testers, and meterless tube checkers are being added to the Supreme line. It is believed that the meterless tube checker will meet a long-felt need for a tube checker which can be operated in conjunction with the meters of any good analyzer. This meterless tube checker will incorporate the cathode-heater leakage test and the capacitor leakage test of the AAA-1 Diagnometer, but will be available at a cost much less than that of tube checkers of comparable quality.

Resistance Measurements

The analyses of radio receivers by resistance measurements will, undoubtedly, be recommended in future radio service manuals, and the Supreme products are designed to utilize all of the advantages of these methods as well as affording the usual potential and current measurements during analyses. For service on radio receivers on which resistance data is not published, a decade resistance box has been designed at a low cost. This unit can be connected across an open resistor in a receiver and adjusted for correct potential and current values, when the adjustment of the decade box will indicate the proper resistance value which should be permanently connected in place of the defective resistor. Terminals are provided for measuring the potential across the decade resistors, so that the power requirements of a required resistor can be determined as well as the ohmic value. This box covers the entire range of resistance values found in commercial radios. It is obvious that, with the proper testing apparatus, resistance analyses will greatly simplify and expedite the handling of radio service jobs. It is almost impossible for new radio developments to make ohmmeters and decade boxes obsolete.

Many radiomen who do not already have good testing units are interested in obtaining all of the essential equipment in a single portable unit, such as the AAA-1 Diagnometer with its improvements for accommodating the new 6-pin and 7-pin tubes without adapters. All of the testing functions required in practical radio servicing are incorporated in this unit, including a stabilized oscillator with 100% modulation for covering all intermediate and broadcast frequencies. The oscillator, analyzer, tube checker, and ohmmeter functions of the Diagnometer are separated so that it is practically impossible to harm any of its parts by inadvertent connections.

The ohmmeter ranges extend to 500,000 ohms with a self-contained 4.5-volt battery, and higher ranges are available with external power. The Diagnometer power unit supplies a D.C. potential of 250 volts for leakage tests, etc. Some users of the AAA-1 Diagnometer keep records of overall measurements made on each serviced receiver at several tuning frequencies before and after adjustments, and these records become valuable for making comparisons between receivers. This procedure also enables the radioman to determine the percentage of improvement in sensitivity adjustments in daytime and whether or not a receiver has sufficient sensitivity for distant reception at night. It is believed that such servicing practices with the best of equipment will continue to enable active radiomen to make money because of better service rendered to more customers.

* Chief Engineer, Supreme Instruments Corp.

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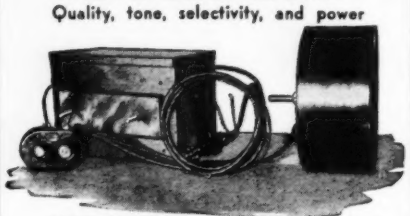
Quality, tone, selectivity, and power



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QUASI-OPTICAL HOME EXPERIMENTS

By JOHN B. BRENNAN, JR.

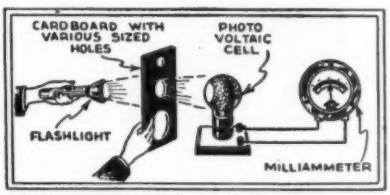
Down below that end of the frequency spectrum where the short waves are to be found, there is a band of frequencies so short in wavelength that they are classed as either light or heat waves and are more correctly termed *quasi-optical frequencies.*

In consequence of the present popularity of experiments and investigations into the realm of television and the transmission of intelligence over short distances by extremely short waves, much attention has been focused on the various and unexplored usages of special types of radio apparatus and equipment such as photo-electric cells, crater lamps, neon tubes, straight-line frequency audio amplifiers, photo-voltaic and other forms of light-sensitive cells, parabolic reflectors, sound-operated light modulators and the whole galaxy of new and unheard of equipment associated with this field of investigation.

To that dyed-in-the-wool and ever growing group of serious minded experimenters who are fast becoming sated with the rehash of present day offerings in the line of radio receiver designs, any new field of endeavor which shows promise of providing an outlet for their pent up energies is eagerly embraced.

It is with the idea in mind of providing such an outlet through the agency of planned home experiments in the quasi-optical field that this department, from this issue of RADIO-CRAFT on, is offered for your serious consideration.

Each month, in this space, one simple experiment will be described and explained in detail, so that those of you who have the time, equipment and inclination may find out for yourselves the marvelous commercial possibilities in the quasi-optical field. The experiments will use only apparatus which can be easily and economically constructed on the kitchen table. As an example of what may be expected the following tentative experiments, to be described completely in later issues of this department are listed and partially described.



A simple experiment to demonstrate current variations with light intensity.

(1.) Various types of light-sensitive cells and how they work; a simple photoelectric cell and amplifier to demonstrate its ability to change light variations into variations of electrical current.

(2.) A typical photo-electric cell amplifier and how to build it.

(3.) The uses of various kinds of lenses to concentrate, magnify or focus sources of light on photoelectric cells.

(4.) How light may be bent through a quartz tube to actuate a photoelectric cell.

(5.) Forms of light modulators: crater tubes, neon tubes, their uses.

(6.) Mechanical light modulators.

And so the list might go on almost indefinitely. Suffice it to say that the main purpose of this department will be to explain, describe and demonstrate only those experiments which the home experimenter can duplicate at a minimum of expense and labor, to the end that his store of knowledge and experience in this particular field of radio may be enhanced.

As an example of what can be done, the following rather simple experiment is outlined. The parts required for it are: a flashlight, a photo-voltaic cell, a milliammeter (0 to 5 ma.) and some bits of connecting wire.

Connect the two terminals of the photo-voltaic cell to the two terminals of the milliammeter, as shown in the accompanying diagram. It is possible that even the daylight, depending upon the light conditions under

(*Continued on page* 247)

MODERNIZING THE HI-Q 29

(*Continued from page* 218)

voltmeter with a scale deflection of at least 500 volts, D.C.; connect the negative lead to clip No. 12 and the positive lead to clip No. 11. A type '80 tube is inserted in the tube socket and the unit is plugged into the A.C. line. The "no load" voltage reading should be approximately 418 volts. Note: Do not leave the power supply unit turned on longer than necessary to make this test.

List of Parts

Five Hammarlund No. ML 17 .00035-mf. Midline condensers, C1, C2, C3, C4, C5;
One Hammarlund No. Hi-Q 29 coil set, L1, L2, L3, L4, L5, L6;
Three Hammarlund No. RFC 85 radio frequency chokes, L7, L8, L9;
One Sangamo .00025-mf. fixed mica condenser, C6;
One Sangamo .00025-mf fixed mica condenser, C12;
One Yaxley wire-wound potentiometer 1,500 to 100,000 ohms, R8;
Two Electrad 500-ohm flexible resistors, R1, R2;
Two Thordarson type R-300 audio transformers, AFT1, AFT2;
One Thordarson type R-196 choke coil, AFC;
Six Parvolt .5-mf. series 200 condensers, C4, C7, C8, C9, C10, C11;
One Parvolt 2 mf. series 400 filter condenser, C16;
One Parvolt 2 mf. series 200 filter condenser, C17;
One Parvolt 4 mf. series 200 filter condenser, C18;
One Parvolt 3. mf. bypass block, series 200, C19;
One Parvolt 1 mf. series 200 filter condenser, C20;
One Durham metallized resistor 2 megohm, R4;
One Yaxley 20-ohm center tapped resistor, R7;
One Yaxley No. 012 cable connector and cable, Y.B.
One RX. Electrad Truvolt HI-Q 29 type resistor;
One Thordarson No. R-171 power compact, PT;
One* Federated Purchaser filament transformer, catalog No. 2610, FT;
One Parvolt 4 mf. series 200 bypass condenser, C15;
Two Yaxley No. 422 insulated phone tip jacks, L.S.;
Two Eby engraved binding posts, ANT. GND;
Two Benjamin No. 9040 UX socket, VT.6;
Four* Na-Ald base mounting type UY sockets, VT1, VT2, VT3, VT4;
One* Tru-test 10,000-ohm ½ watt resistor, R3;
One* C midget variable condenser, .0005-mf. USL;
One* Screen-grid cap connector and wire;
HI-Q 29, master foundation unit containing drilled and engraved panel three complete aluminum shields. Drilled and formed steel chassis. Fahnestock clips. Brackets, clips, wire, screws, nuts, washers, and all necessary hardware to complete this receiver.
One* steel shaft 22 inches long, ¼ inch in diameter.

* Indicates additional parts other than the regular kit of parts necessary to construct this receiver.

NEW TUBES

(*Continued from page* 205)

cury-vapor type is recommended because it has a low and practically constant space-charge-voltage drop within its operating limitations. As a further means of obtaining good regulation, the filter chokes and transformer windings of the "B" eliminator should have as low resistance as possible. In the design of a power supply for a class B amplifier, consideration should be given to the peak current demand of the amplifier.

As may be expected, the grid (No. 1 and No. 2) of the 89 is operated sufficiently positive to cause grid current to flow in its input circuit. This feature imposes a further requirement on the preceding amplifier stage. It must supply not only the necessary input voltage, but it must be capable of doing so under conditions where appreciable power is taken by the grid of the class B amplifier tube. Since the power necessary to swing the grid positive is partially dependent on the plate load of the class B tube, and since the efficiency of power transfer from the preceding stage is dependent on the transformer design, it is apparent that the design of a class B audio, power amplifier requires that more than ordinary attention be given to the effects produced by the component parts of the circuit. These effects may be produced in the first-stage amplifier by the design factors of the power-output stage. For this reason, the design of a class B audio amplifier with its driver stage is somewhat more involved than for a class A system, and must be checked for each change in the component parts.

The following ratings and characteristics obtain: Heater voltage, 6.3 volts; heater current, .4-ampere; for class A triode operation, plate voltage 160; control-grid bias, -20 volts; plate current, 17 ma.; load resistance for undistorted power output, 7,000 ohms; amplification factor, 4.7; plate resistance, 3,000 ohms; mutual conductance, 1,570 micromhos; undistorted power output, 300 milliwatts.

For class A pentode operation, the following characteristics obtain: Plate voltage, 180; screen-grid voltage, 180; control grid bias, -18; plate current, 20 milliamperes; screen-grid current, 3 milliamperes; power output, 1.5 watts; load resistance, 8,000 ohms; amplification factor, 135; plate resistance, 82,500 ohms; mutual conductance, 1,635 micromhos.

Combination Amplifier and Power Rectifier

The first advantage of the unique tube to be described is to provide an electron emitting device having an electron producing area and a plurality of electron emitting cathodes adjacent to this area. Second, to provide a device that can be used in radio circuits and in which the filament or heater current, as well as the plate, can be taken direct from a source of alternating current. Third, and this is a rather unique advantage, to provide a tube having a multi-potential grid, that is, different parts of the same grid to have different potentials with respect to the cathode. Fourth, to provide a self-biasing grid.

The sketch of Fig. 6 illustrates the external view of this rather unique tube. Fig. 7A is a plan view of the internal assembly; 7B is an end view of the assembly shown in Fig. 7A; 7C is a cross section view along the line 4—4 of Fig. 7B; Fig. 7D is a section showing the relative positions of the elements; 7E is a cross section of Fig. 7D along the line 6—6. 8A shows the insulating frame that supports the elements; 8B shows one method of constructing the elements. Two schematic circuits showing the application of this tube are given in Fig. 9A and 9B.

(*Continued on page* 236)

CORRECTION NOTICE

Due to an error in printing, the cost of constructing the Radio-Craft Universal Analyzer was inadvertently printed as $28.00. A comprehensive investigation by Radio-Craft indicates that the average price for which the parts may be secured is about $48.00.

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Point-To-Point Resistance and Capacity Tester

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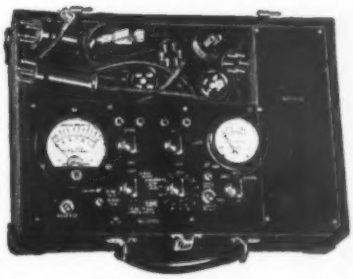
No. 1000



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Two cable plugs connect set sockets through selector switch to the precision volt-milliampere-ohmmeter,—an exclusive Readrite feature. Speeds up tests without removing chassis. Measures resistances up to three megohms — D.C. Volts up to 600 and milliamperes up to 300. Also A.C. line volts and capacity of dry and electrolytic condensers.

A fine instrument in leatherette case. Complete with batteries, test cords and full instructions.

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So many of our customers have taken advantage of our bonus that we decided to extend this offer . . . .

Take advantage of our offer—25 UY-227 Tubes FREE with every purchase of 100 assorted tubes.

(This offer is good until October 7th)

Give to your customers the advantage of buying first-quality licensed R.C.A. tubes at prices that are exceptionally low. We carry the largest variety of tubes in the world, many for special purposes, and every tube is replaceable within three months, providing the filament is not burnt out. We either have the tube you require or we will make it for you.

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Potassium Photo Cell, 4" overall. $2.10

TYPE V
Television Tube 1½" square cathode, overall size 2"x6". $3.85

TYPE R
Caesium Photo Cell, overall length 3-1/16". $5.90.

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UX-201A	—Amplifier or detector	$0.30
UX-226	—AC amplifier	.30
UY-227	—AC amplifier or detector	.30
UX-171	—Same characteristics as 171A on tungsten filament, ⅛ amp.	.30
UX-171A	—Power output amp. for AC or DC operation, ¼ amp.	.30
UX-240	—Designed for impedance and resistance coupling	.40
UX-120	—Power amp. used in last stage of audio freq.	.40
UX-199	—Detector and amp. tube, long prongs	.40
UV-199	—Standard base, has a 201A base, same characteristics as 199	.40
UV-199	—Same characteristics as UX-199 only short prongs	.40
UX-112A	—Power amp. tube for low cur. consump., ¼ amp.	.40
UX-142	—Power Amp. ½ amp.	.40
UX-200A	—Detector tube recommended for weak signals	.40
UY-224	—AC screen grid amplifier	.40
UX-245	—Power amplifier	.40
UX-201B	—1/8 Amp. Amplifier and Detector	.60
UY-246	—Power Amplifier	.60
UY-247	—Power pentode	.60
WD-11	—Detector Amp.	.60
WD-12	—Detector Amp.	.60
UX-230	—Dry cell amp. and detector, 2 volts	.60
UX-231	—Dry cell amp.—last audio stage, 2 volts	.60
UX-232	—Dry cell screen grid amp., 2 volts	.60
UY-233	—Power Amplifier Pentode, 2 volts	.85
UX-234	—Screen Grid Pentode R.F. Amplifier	.85
UY-235	—Super control screen grid amp.	.60
UY-236	—Screen Grid Radio Freq. Amp.	.85
UY-237	—Detector Amplifier	.85
UY-238	—Power Amplifier Pentode	.85
UY-239	—DC R.F. Pentode Amplifier	.85
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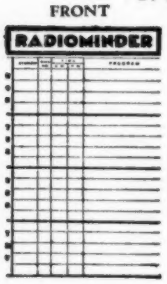

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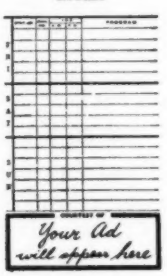

NEW TUBES

(Continued from page 235)

In Fig. 6 a glass envelope is shown at 1, the upper end of which is provided with a screw type of socket connection having terminals 2 and 3, which connect to filament or heater leads 4 and 5 respectively. These leads serve to support the elements of the tube by being fastened to sleeves 6, the latter forming an integral part of the insulating frame which may be made of lava rock, porcelain or any other suitable material.

The horizontal positions of 7, Fig. 8A, are provided with slots 8 adapted to receive the heater elements, the ends of which terminate into leads 4 and 5. When an equipotential cathode, such as in the '27, '24, etc., is desired, another element, 9, is provided. These cathodes are in the form of plates as illustrated by 10 of Fig. 7B held into longitudinal slots 11 as illustrated in Fig. 8A. The outside faces of these plates are coated with an electron emitting substance.

Part 7 illustrated in Fig. 8A is provided with 4 vertical rows of holes arranged to receive the control grid. This grid may be in the form of a continuous wire arranged in parallel positions close to the cathode. However, where it is desired to place a negative potential on this grid, the grid may consist of a number of thermo-couples each consisting of two dissimilar wires as illustrated in Fig. 7C.

In this figure the positive and negative metals are shown connected alternately and indicated as P and N respectively. Thus, when the central row of junctures (points at which the dissimilar wires connect), *h* in the figure, becomes heated because of radiation by the heater 9, the outer rows of junctures at C of Fig. 7C will remain relatively cool thereby generating a direct current in the grid. Hence, each grid wire will have a potential different from its adjacent wire.

The plates illustrated in Fig. 8B surround the grid structure in the conventional manner. These plates are connected by straps 13 and rest against surface 14 of Fig. 8A.

When the tube is to be used in the circuit which is to rectify the A.C. applied, more than two anodes are required as shown in the schematic circuits of Fig. 9. Likewise, when more than one cathode is desired, the sections may be insulated from one another or separate leads brought out by each cathode. The advantages to be gained are quite obvious. First, this tube may be used in circuits requiring more than one grid, each grid having a different potential with respect to the cathode but all of them having a common input circuit. Second, this device may be used in connection with well-known type of battery coupled A.F. amplifiers in which "C" bias is required. With the device as described here, it can be seen that an *internal* thermo-electric potential is derived for use in such circuits.

Of special interest to us are the diagrams of Fig. 9. A source of alternating voltage is applied to the heater 9, as well as to the primary winding 16 of a transformer, the secondary 17, of which, has a center tap 18, connected to one lead 19, of the D.C. terminals 19 and 20.

The tube drawn in Fig. 9A has two anodes 12, connected in parallel as shown, which lead into a series of condensers 21 and choke coils 22, before connecting to the points 19 and 20. It may be seen, therefore, that the two outside terminals of the secondary 17 apply voltage to the two inside plates, causing a pulsating D.C. to flow through the filter unit which in turn is applied to the two plates 12 which function in the normal manner. Fig. 9B discloses the somewhat similar full-wave rectifier circuit with the exception that the terminal 20 is derived from a tap between resistors 23 connected in series. The anodes are connected to the outside terminals of the secondary 17 and the cathodes to parts 23.

It would be extremely interesting to note the application to which this tube will be put, for it will evidently become a commercial product. It illustrates, as stated previously, that the trend in modern tube design has been toward a minimum of tubes and an increase insensitivity.

AN ULTRA-MODERN SUPER.

(*Continued from page 203*)

Wiring

The wiring should be done with a good grade of push-back wire and all connections carefully soldered. Wire the filament circuits of all tubes first, twisting these leads together. Then wire the plate-voltage supply, including the rectifier tube, electrolytic condensers, choke, the socket used for connections to the speaker field, and the voice-coil transformer. After this portion of the wiring is finished, complete all plate and plate-return leads to their respective socket and high-voltage supply terminals.

The grid circuits are wired last; *and be sure that the connections of this portion of the circuit are as shown in the schematic circuit. Any deviation may prevent the proper operation of the automatic volume control tube.* Study this portion of the circuit carefully and follow the connections as shown.

Speaker Selection

As the speaker field is a portion of the filter circuit, it is important that the field have the proper resistance. This circuit has been designed so that considerable variation in the resistance of the field is permissable without upsetting the balance of the various voltages used throughout the set. Do not use a field with a resistance of less than 1,000 ohms or more than 2,500 ohms. The plate voltage on the plates of the I.F. tubes, with a speaker field-resistance of 2,500 ohms, will be just under 200 volts; speaker fields with lower values of resistance will raise this voltage to about 250 volts. If the D.C. resistance of the speaker field is dropped below 1,000 ohms, the filtering action will be impaired and the applied plate voltage will be too high.

The speaker must be provided with an output transformer to match the two pentodes used in the output stage. No provision having been made in the set for placing the output transformer, it must be mounted on the speaker.

Testing and Adjustments

After the set has been completely wired and checked, it is ready for test. Use an oscillator and tune the intermediates to 175 kc. When this has been done, the set can be connected to an antenna, and stations tuned-in.

Adjust the oscillator padding condensers for maximum sensitivity over the broadcast band. Condenser C3A is used to trim the oscillator tuning at the high-frequency end of the tuning scale and C4 trims this circuit at the low-frequency end. Use the slotted plates of the oscillator tuning condenser, C3, for points between the high- and low-frequency settings compensated for by C3A and C4. The care used in tuning the oscillator tuning condenser will reflect in the over-all sensitivity of the receiver.

When the set is properly adjusted, there should be no difficulty in picking up at least 60 channels without any interference from adjacent stations.

Parts List

One Radio Condenser Corp. four-gang tuning units, with trimmers, type 6654, C1, C2, C3, C3A, C5;
Two Micamold condensers, 750 mmf., type 6636, C4A, C13;
One Micamold variable condenser, 5 to 100 mmf., type 2883, C4;
Six Acratest variable condensers, .0002-mf., type 5358 (these are part of the I.F. transformers), C6, C7, C8, C9, C10, C11;
Six Acratest tubular condensers, .1-mf., C12, C14, C16, C17, C26;
One Acratest tubular electrolytic condenser, type 6650, 50 volts, 10 mf., C18;
Two Micamold condensers, 350 mmf., C19, C28;
One Micamold condenser, .006-mf., C20;
Two Acratest dry electrolytic condensers, type 5308, 8 mf., C21, C22;
One Acratest bypass condenser, type 2837, .25-mf., C23;
One Micamold condenser, 50 mmf., C24;
Two Flechtheim bypass condensers, type GF-100.—C25, C27;
One Acratest band-pass tuning unit, as described, type 6654, R.F.T.1 (consisting of L1, L2, and L3);
One R.F. choke, L4 (This may be constructed by winding 700 turns of No. 36 enameled wire on a wooden spool ½-inch in diameter and ½-inch long);
One Acratest oscillator-detector tuning unit, type 5308, L5, L6, L7;
Three Acratest I.F. transformers, as described, type 5358, I.F.T1, I.F.T.2, I.F.T.3 (consisting of coils L8, L9, L10, L11, L12, L13);
One Acratest 1.-watt resistor, 20,000 ohms, R1;
Two Acratest .5-watt resistors, 400 ohms, R2, R8;
Three Acratest .5-watt resistors, .5-megohm, type 5860, R3, R7;
One Acratest .5-watt 6,000-ohm resistor, R4;
One Acratest .5-watt, 2,000-ohm resistor, R5;
One Acratest 1.-watt resistor, 25,000-ohms, R6;
One Acratest .5-watt, 500-ohm resistor, R9;
One Frost .5-megohm volume control, R10;
One Acratest .5-watt, 750-ohm resistor, R11;
One Acratest .5-watt, .1-meg. resistor, R12;
One Frost .5-megohm potentiometer with switch SW.2, R13;
One Acratest 5-watt resistor, 200 ohms, R14;
One Acratest .5-watt resistor, 40,000 ohms, R15;
One Federated cadmium-plated chassis;
One Acratest power transformer, type 2527, PT.;
One Acratest filter choke, 30 hy., CH.1;
One rotary snap switch, SW.1;
One Acratest push-pull input transformer, type 5834, A.F.T.1;
One Acratest tuning dial, type 4047;
Four knobs for the controls;
Five screen-grid clips;
Five Acratest tube shields, type 58;
Five Cinch 6-prong wafer sockets, four marked "58" and one marked "55";
One Cinch 4-prong wafer socket marked "280";
Two Cinch 5-prong wafer sockets marked "47";
One Cinch 5-prong wafer socket marked "56";
One Cinch 5-prong wafer socket, no marking;
One Alden 5-prong plug with short cap, type 5869;
Five feet of lamp cord and plug;
One Eby antenna-ground strip, A, G;
Four ER type 58 tubes;
Two ER type '47 tubes;
One ER type '80 rectifier;
One ER type 55 Duplex-Diode Triode;
One ER type 56 tube.

A SERVICE LAB.

(*Continued from page 219*)

may be connected "A," "B" and "C" potentials required to test battery-operated receivers; (the connections from these batteries are brought out to binding posts at the lower left-hand corner of the panel); there is also incorporated a rotary switch for checking the plate voltages at the various terminals of the "B" supply. There are two switches; one is a T.P.D.T. for the filament and "B"; a 5-P.D.T. switch controls the "B" plus potential. When the two switches are in the upward position they control the current to panel No. 8; when they supply the necessary current to Panel No. 10.

Panel No. 10 is an R.F. oscillator which tunes from 550 to 1,500 kc. This panel is also used for the transmission of phonograph records picked up in other parts of the building.

This covers a general description of all the mounted panels.

The portable equipment consists of a shielded oscillator; microphone speech amplifier used in public address work; a standard Jewell oscillator; a G.R. output meter; two Weston 565 tube checker-analyzers. The 565's are used by the outside Service Men.

The two heavy bare copper wires, running parallel underneath the panels, connect to an RCA antenna system on top of the roof. This arrangement enables a set to be connected at any position on the bench.

The steel cabinets are used to house a complete stock of resistors and other accessories for which we are jobbers.

The public address equipment which consists of ten Racon dynamic units with output transformers and trumpets is not shown in the photograph.

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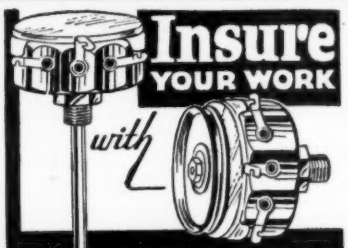

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Be sure to read the announcement on Page 256 which tells about the 1933 Official Radio Service Manual.

A LAPEL MICROPHONE

(*Continued from page* 210)

or less current through the microphone, *we not only change the volume, but at the same time the pitch and quality of the sound impulses.* This is important from the reproducer's viewpoint.

One Man Control

As mentioned in our first article, this lapel microphone can be operated independently from the main controls by having a small control device in the pocket of the speaker. It is thus possible for the speaker, from his own location, to operate the output of the loudspeakers without being compelled to remain near the amplifier unit, thus giving him greater freedom and making a second man unnecessary.

Considering the curves as shown in Figs. 2A and 2B, it will be readily understood that even if a resistor is used as a volume control (and a small variable resistor can be easily placed in a box and carried in one of the pockets), not only can the volume be changed voluntarily by the speaker, but in addition, means are given to him to change the pitch of his voice as may be necessary.

In connection with the electro-acoustical characteristics of this small microphone, it might be well to mention the fact that the quality of sound is also influenced by the direction in which the current passes. It makes a difference whether the current comes from the metal membrane to the carbon grains or the contrary—the resistance changing with the *direction* in which the current passes.

Taking Sound Curves

These curves have been made by exposing the microphone to the sound of a loudspeaker which was fed with the same amount of current at all frequencies. After the electrical impulses coming from the microphone are amplified by an amplifier of practically linear characteristics between 20 and 10,000 cycles, they are impressed upon a meter.

These curves show clearly that the general transmitting microphone is less sensitive to the general frequency range and almost exclusively pitched to one single frequency which, in our case, is between 2,000 and 3,000 cycles. It may be remarked that curve 1 of Fig. 3 corresponds to a two-button microphone, while curve 2 of the same figure is taken with only one button. All these curves have been actually measured and are independent of any that may have appeared in commercial publications.

The Amplifier

The microphone has polarized prongs so as to make sure that the current flows through the microphone in the right direction. The prongs are plugged in the control-box carried in the pocket and connected from there to the amplifier shown in Figs.4 and A. As may be seen, a concentrated setup has made possible the placement of the loudspeaker and the all-electric amplifying equipment in one easily transportable carrying case.

The microphone part of the equipment is attached at the upper left corner of the box. This equipment is small and has thus the advantage of easy transportability which means, for instance, that at an election talk, the speaker is able to have his public-address system with him, as if carrying a suit case, and operate if all by himself. Arriving at his place of destination he simply pushes the plugs in at the lower-right of the case and the apparatus is ready for work.

Figure 4 shows a wiring diagram of this amplifier. The rectification of the current takes place in two '81 tubes. The input from the microphone is first amplified by a screen-grid tube which is resistance coupled to one '45 tube. The latter is coupled by a transformer to two power tubes of the 250 type which are wired in push-pull, the output of which is fed over a transformer to the loudspeaker in the case.

Instead of one loudspeaker being attached to the carrying case, this amplifier can be connected to a number of loudspeakers which can be distributed throughout the hall and the sound of the speaker's voice carried to the farthest corner.

Thus, simple as it seems in its final solution, a vastly improved means has been given to the speaker for amplifying his voice and making his speech clear to every person in the audience.

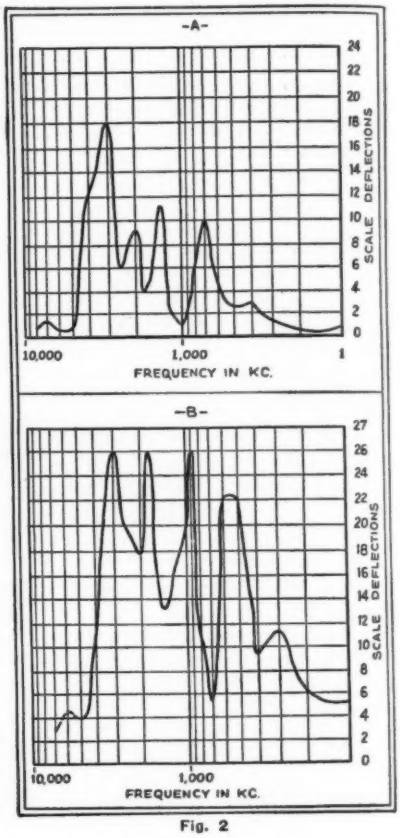

Fig. 2

CRAFTSMAN'S PAGE

(*Continued from page* 229)

May I submit the following for your comments as well as for the thoughtful consideration of tube manufacturers?

There are on the market today numerous receiving tubes, to be sure; the filament and heater voltages vary from 1½ to 15 and in certain cases even more. In the earlier days of "all electric" receivers a common combination of tubes in a set included a type '27 detector, type '26 in the R.F. and first A.F. stages, and sometimes two '26 tubes in push-pull in the second A.F. stage; (if indeed two stages of transformer-coupled audio frequency amplification did not round out the combination!); later on the '71 and '71A tubes came into use in the output stage.

To say, definitely, as to how many sets there are in use, today, that use type '26 tubes, is beside the point; in any event there are thousands upon thousands of them. In addition there are thousands of engineers, experimenters, amateurs and Service Men who use tubes of the '26 type in some circuit or another. The hum feature, of course, has been the drawback to the more universal use of this tube.

With the advent, recently, of the types '56 and '57 indirect-heater tubes, which operate with a heater voltage of 2½ and with a current intensity of only one ampere, a forward stride has been made by our friends the tube manufacturers.

(*Continued on page* 245)

A. V. C. SYSTEMS

(*Continued from page* 211)

where a diode detector was followed by four stages of audio (including the output stage), remarked that if the efficiency were to be judged by the number of audio stages, it *had* to be able to handle a high input.

The manner of operation is simple. Reference to Fig. 5 will show that one side of the tuned secondary coil is connected to the anode; the other side is connected to the cathode through two resistors, the same side of the coil is also bypassed to the cathode which is grounded to the chassis. Since the tube functions as a rectifier, current can flow in only one direction, from the anode *to* the cathode.

The unrectified half of the wave is bypassed through the small condenser C1, just mentioned. The rectified half of the wave flows through the two resistors developing a potential across them, which varies with the strength of the signal. One of these resistors (Rc) is bypassed with a fairly large condenser (C2) and its voltage does not vary with the signal modulation. This voltage is used to control the bias on the radio-frequency tubes. The other resistor (R1) has only a small condenser across it (C1) serving as a radio-frequency bypass, and the voltage across it varies with the modulation and is fed to the audio stage. The resistor R2 serves to block radio-frequency currents and keep them out of the audio-frequency end of the channel.

The potentiometer in the grid circuit of the A.F. tube serves the dual purpose of a grid leak and a volume-level control. So far now, we have shown that a constant signal level was maintained but have said little about how great this level was; usually, it is equal to the maximum power that the output stage can handle. Since it is very seldom that such high volume is required, some means of reducing it must be employed. To achieve this end, one of the familiar types of A.F. volume control systems are used. This applies any desired portion of the total output to the output stage, and when it is set for any desired level, it is maintained for all stations. The only exceptions being those too weak to produce the normal output when a state of maximum sensitivity exists.

Another type of volume-level control is shown in Fig. 4, which consists of a variable resistor across the secondary of the input transformer.

The Tuning Meter

The tuning meter shown in Fig. 4 is merely a milliammeter mounted upside down. It is mounted in this manner so that as resonance is approached and the bias applied to the controlled tubes begins to increase, with a consequent reduction of plate current, the needle will swing to the right. Thus, resonance is indicated by the point of maximum deflection. It also gives some indication of the strength of the signal being received, for the more powerful the signal, the further to the right the needle will swing, and vice versa.

The meter may be placed in the common cathode lead as shown, or it may be inserted in the "B" common lead of the plate supply. Another location is in either the plate or cathode lead of any one of the controlled tubes. While there is no objection to the use of a moving coil instrument, one of the inexpensive repulsion type instruments is usually used. When a meter of this type is used in the cathode circuit, it will be necessary to reduce the resistance in the cathode circuit, by an amount equal to the resistance of the meter, to prevent the minimum bias from being increased to more than the rated 3 volts due to the extra resistance of the instrument. The meter shown in Fig. 4 requires 15 ma. for a full-scale deflection; if it is desired to be used with only one tube, a meter that will show a full-scale deflection with a current of about 5 ma. should be used.

This should offer an opportunity for the alert Service Man to pick up quite a few dollars. There are thousands of receivers in use that employ automatic-volume-control systems without a tuning meter. Most of the owners of these sets will be glad to have their sets modernized by the addition of this device.

Alignment Procedure

Alignment procedure on receivers using A. V. C. has puzzled a great many Service Men. Obviously, an output meter cannot be used, because if a powerful signal is applied, the A. V. C. keeps it at the same level regardless of ordinary adjustment of the trimmers. It is possible to align a receiver by ear, using a signal too weak to operate the automatic volume control, but this method is notoriously inaccurate. If the receiver is equipped with a tuning meter simply adjust for maximum swing of the needle. If a tuning meter is not used, it will be necessary to insert a milliammeter in the plate circuit of one of the controlled tubes ahd adjust the compensators until it shows minimum current. In some cases it will be found that one of the trimmers does not affect the meter. This occurs when the control tube is coupled to the plate of the last R. F. tube, and the condenser that does not affect the tuning meter tunes the input to the detector. It will be necessary to use an output meter to accurately adjust this one, or it may be adjusted by ear; in most cases this is sufficiently accurate.

It is hoped that this article will clear up some of the mystery that shrouds the whole subject of automatic volume control, and that some of the readers of this article will feel like tackling the job of building an automatic volume control into their next receiver.

If any points have not been made sufficiently clear I will try to answer any letter in which a stamped, self-addressed envelope is enclosed.

A MODERN REFLEX RECEIVER

(*Continued from page* 212)

grid terminal is the one isolated from the other four. Looking at the socket from the underside and working clockwise from the grid terminal, the other terminals are cathode, filament, filament, and plate, respectively.

After the wiring is completed and checked, the only adjustments necessary are at the trimmer condensers, 7A and 17A.

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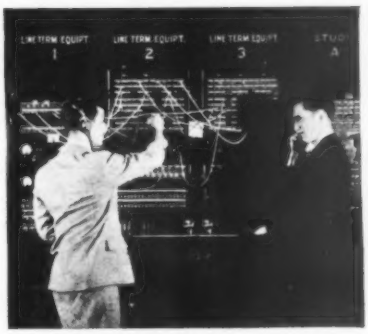
One Electrad Tapered Potentiometer, type R1-240-P, 12 with switch, 32;
Two Electrad Flexible Grid Resistors, 200 ohms, 3, 11;
One Electrad 1500-ohm Flexible Resistor, type 2G 1500, 30;
Two .006-mf. type 1450 Aerovox Mica Condensers, 5, 22;
One Aerovox Mica Condenser, .00015-mf., type 1450, 18;
One Aerovox Metal Case Resistor, .1-mf., type 260, 16;
One Aerovox Duplex Metal Case Resistors, .1-mf. each section, type 260-21, 8, 9;
One Aerovox Metal Case Resistor, .5-mf., type 260, 21;
One Aerovox Dry Electrolytic Condenser, 10 mf. type E-50-10 with mounting ring, 29;
Two Aerovox Dry Electrolytic Condensers, 4 mf., type E5-4 with mounting rings, 25, 25A;
Two Find-all R.F. Coils, 6, 15;
Two Trutest R.F. Chokes, 8 millihenries, type 2H10100, 2, 20;
One Trutest Power Supply Transformer, De-Luxe Model, type 2C-1512, 31;
Three Trutest 4-prong wafer-type sockets, 28, 33, 26;
Two Trutest 5-prong wafer-type sockets, 4, 23;
One Trutest 6-prong wafer-type socket 10;

(*Continued on page* 250)

WHAT YOU NEED TO SUCCEED IN RADIO



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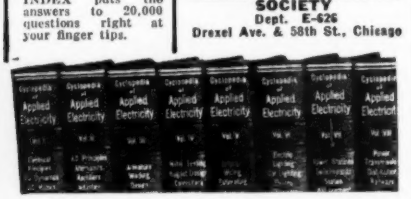

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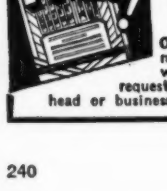

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TUNING METERS

(*Continued from page* 214)

glance seems to contradict the statement made previously.

If it is recalled that the fundamental function of any A.V.C. system is to increase the sensitivity of the amplifier for weak signals and decrease it for strong signals, the reason for the connection becomes apparent. (See the article on A.V.C. in this issue.) To decrease the sensitivity of an amplifier, either one of two methods are employed: (1) increase the 'control-grid bias when variable-mu tubes are used, and (2) lower the screen-grid potential if non-variable-mu tubes are used. Either method of control reduces the plate current flowing in the amplifier stages connected in the A.V.C. circuit as the applied signal increases. This decrease in plate current reduces the reading of the D.C. meter which is connected in the plate circuit of these tubes and furthermore, the greater the signal strength the more the A.V.C. action reduces the plate current and the less the meter reads. This explains why the tuning meter reads full scale with no signal and reads less as the signal increases.

Regardless of the system used, let us suppose for the instant that we are turning the dial of a radio set equipped with a tuning meter. As we approach the extremity of the side band of a carrier, very little action results, but as soon as we receive an appreciable signal (the signal voltage necessary for action varies with different receivers) the current through the tuning meter decreases and therefore the meter reads less. When we reach the peak of the signal, the meter reads a very low value which increases as we pass exact resonance. The lowest reading of the meter represents the best setting of the dial for that particular station.

If the meter is connected in the plate circuit of the detector, the best position of the tuning dial is indicated by the greatest reading of the tuning meter as explained above.

Various Forms of Tuning Meters

It is not necessary to have a milliammeter connected in the plate circuit in order to have a tuning meter. The Tune-A-Lite described in the November, 1931 issue of this magazine is an example of a so-called visual tuning meter. This particular device consists of a glass tube filled with neon gas and two electrodes. It is connected in the same manner as a milliammeter would be connected. As the signal increases, the gas ionizes, but in this device the *height* of the ionized layer varies. This device is used in some commercial receivers and may also be secured separately.

The Tuning Meter as a Trouble Indicator

Aside from the obvious advantages of the tuning meter as an indicating device, it may be used, with a little thought, as a trouble indicator. If the A.V.C. tube of a receiver be removed, the reading of the meter will not change as the strength of the applied signal changes. Thus the reading of the meter is an indication of the condition of the A.V.C. circuit.

A poor tube in the A.V.C. circuit may be determined by tuning-in stations of varying intensity and noting the change in volume. If there is a radical difference in the volume of two loud stations, then the A.V.C. tube is not functioning. However, the method outlined above is far more efficient. If, upon test, the A.V.C. tube is found to be in good condition, but the tuning meter does not function properly, in all probability the resistors in the A.V.C. circuit are defective and should be tested.

Ranges of Tuning Meters

Although tuning meters are simply milliammeters used in a particular manner, nevertheless meters are made especially for this purpose usually having an inverted scale as shown in the photograph. When used in the plate circuit of a power detector, the range of the meter used should be about 5 ma. If the pointer swings off scale, then a shunt is necessary. However, with a 5 ma. scale no shunts will usually be required.

(*Continued on page* 250)

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INTERESTING LOUDSPEAKER FACTS

(*Continued from page 213*)

that the amplitude of each resonant position is proportional or equal to the necessary and required values.

Case Design

Much progress has been made in the design of aperiodic cones from which the majority of resonant positions have been eliminated. This is the opposite extreme from that just mentioned; for in this case, all possible things that might possibly cause resonant positions are carefully eliminated. A few years ago it was quite common to hear a distinct and annoying paper rattle in the majority of cone loudspeakers. Analysis showed that the seam in the cone where the cone was glued together was one of the most important reasons why this rattle was prevalent. Therefore an expensive process was developed that pressed the entire cone out of a singl piece of material. This immediately eliminated the resonance caused by the seam and almost all cones in speakers today are made in this manner.

It was also found that the kind of material used in the cone itself had a direct bearing on the number of resonant positions. Metals used as diaphragms almost always responded better on the highs than on the lows and therefore resulted in a "tiny" response. Also the amount of resonance was high so that this is undoubtedly the reason why metal cones are not used today. It was found that a comparatively dead substance was the best thing if it could be made into a substance hard enough to use. The research departments of various organizations set to work, and now one of the best cones is that made from ordinary burlap bag material.

It was also found that corrugating the cone had a direct effect on the response characteristic obtainable, and although this is still a question of dispute, it is generally agreed that corrugation does materially aid the tone quality, but the main questions are, where, how much, and what shape? Even the wire used on the voice coil of the dynamic speaker affects the response characteristic. The wire used in this coil is usually wound on the coil form and is of copper. Recently, experiments have been made in which the wire is aluminum and is molded directly on the form. It has been found that a light metal is good but increases the hissing type of background noises. The aluminum just mentioned has been used quite successfully.

The size of the cone used on any type of speaker is of major importance. If we set the reed of a magnetic speaker into vibration without any cone or load connected to it, then it most certainly would be badly underloaded. Due to the small amplitude through which the reed moves there would be practically no air displaced and no sound of any consequence would be heard. If a small cone were applied to the reed, it would still be underloaded and there would be no low frequencies. As the size of the cone is increased, the air displaced will also increase and with it the low frequencies. Now, if an extremely large cone were used, then because the fundamental frequency of the cone is lowered with the increase in the size of the cone, and also because of the inherent inertia of the cone itself, too much energy will be expended in the lower frequencies with the result that the higher frequencies will be overshadowed.

What, then, must we do to obtain a balanced response? The answer is that we must match the acoustic impedance of the cone to that of the reed. This is done in most cases by the law of trial and error. A large cone is used in conjunction with a long reed. A series of response tests are made and then the reed is shortened. More tests are made with different combinations until the law of trial and error has definitely proved to the individual conducting the test just which combination gives the best response.

The amount of energy to be radiated also has a direct bearing on the size of the cone. If we have a condition where only a small room is to be supplied, then a cone of about 14 to 18 inches in diameter will supply a pleasing and natural sound provided, of course, that the motor unit is well designed and the reed matches the cone. This size cone is well able to handle up to about 12 decibels with good response. For greater power levels the cone should be increased in size to about 22 inches.

In the dynamic type loudspeaker it is not necessary to use such a large cone to get good low-frequency response and the chief reason is that the action of this type of speaker is different than that of the magnetic. In the magnetic the reed supplies the motive power to the cone, whereas in the dynamic the cone is pushed in such a manner as to actually pump air. This feature in itself would not be especially advantageous if it were not for the greater amplitude through which the dynamic speaker moves in comparison to that of the magnetic. The direction of movement of the powerizer in each type of speaker is also different. The reed in the magnetic speaker actuates the cone with a flexing or sweeping movement, Fig. 3. In the dynamic speaker, the voice coil is in a constant magnetic field and the variation in flux caused by the signal in the voice coil produces motor action in accordance with Lenz's Law. The voice coil, therefore, is either sucked into the field or repelled outwards. Thus the cone itself follows the pulsations of the voice coil and the emanating sound is produced through a piston-like action. The amplitude is so great as to produce a much larger amount of displaced air than that which would be obtained with the same size cone in the magnetic type of loudspeaker.

In conjunction with this more efficient method of radiation, the normal and usual impedance of the voice coil is such as to place the speaker in the low impedance category. This means that it is possible to connect the voice coil directly to a transmission line and transmit the necessary energy to the speaker over this line without a worry about crosstalk if more than one or two lines are used in the same cable or run in the same conduit and operated at similar volume levels.

The principal reason why the dynamic speaker has become so universally prominent and well liked, is undoubtedly due to the low variation in impedance with frequency variation. This would sound as unconnected as the well-known chicken walking across the street because it wanted to get on the other side, yet both reasons are quite correct and exact.

The impedance of the usual magnetic speaker at 1,000 cycles is roughly about 7,500 ohms (Fig. 4). The impedance of this same speaker at 6,000 cycles may be as high, if not higher, than 47,000 ohms! Can you imagine designing a matching transformer or similar device that will also vary its impedance in exactly the same manner as the speaker? In order to produce linear response the matching transformer would have to do just that.

On the other hand, the dynamic speaker has only a very low impedance variation over the frequency spectrum. At 1,000 cycles its impedance is about 14 ohms (Fig. 4), while at 6,000 cycles the impedance rises to approximately 40 ohms—this is certainly different from that of the magnetic speaker. It is fairly easy to design a matching transformer that will operate within these limits. Of course there is the matter of the iron in the speaker being a little different than that of the transformer, but in the dynamic type of speaker the transformer will only mismatch within a few ohms of the actual true value while in the magnetic speaker the transformer may mismatch by several thousand ohms on the high-frequency end. Due to the iron influencing the inductance of the speaker windings and also the inductance of the transformers and thereby, in turn, affecting the impedance, it is seldom that a perfect match is made between the magnetic speaker output matching transformer and the speaker into which it is working. Thus this is one great handicap that is not present with the dynamic speaker.

The Use of Baffles

Let us assume that all the necessary requirements have been complied with and now we desire to install our equipment in various locations. The first thing to consider is the fact that our low-frequency response will de-

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pend directly on the size of the baffle that must be used in conjunction with our loudspeakers. Perhaps one wonders just why a baffle is so necessary in order to obtain good low-frequency response.

Baffling is by no means a new idea, for as early as 1868, Lord Rayleigh, one of the most proficient of the old masters, showed, in an extremely interesting experiment, tat by preventing the air on one side of a tuning fork from circulating with that of the other side, the intensity of the sound from the tuning fork could be tremendously increased. Consider Fig. 5. As the tuning fork moves towards "B," the pressure of the air in this direction will be increased. At the same time the pressure at "A" will be decreased due to the displacement of the tuning fork in the direction of "B." Thus there will be created a difference of pressure between "A" and "B" and the air of the higher pressure will move to that of the lower, or the air at "B" will move to "A." Now this movement of air from "A" to "B" is spoken of as the "Lateral Motion of the Air."

Lord Rayleigh found that by placing a piece of cardboard in such a position so as to prevent the lateral motion of the air, the sound from the tuning fork was greatly increased. Lord Rayleigh, however, was not alone in his experiments for, as a matter of fact, most of his work was based on the results of the experiments of the late Professor Stokes. To indicate the importance of the elimination of the lateral motion, Professor Stokes estimated that in connection with middle "C" of the piano, it would be possible to increase the output 40,000 times by the complete elimination of the lateral motion of the air, thereby causing the air in front of the unaided string to travel in a thin strip.

The above however does not explain why it is necessary to have a baffle for good low frequency response. However, it is the prevention of the lateral motion of the air that brings in those deep lows that are so hard to obtain.

Referring to Fig. 6 in which is shown the radiation curves of the various frequencies, viz; high, medium and low. It will be noticed that the very high frequencies are projected in the form of a beam while, as the frequency is decreased, the angle of radiation becomes larger until at the very low frequencies the waves literally "creep around the cone." Now the cone is similar to the tuning fork shown in Fig. 5 in that it vibrates between two points. If a signal of any frequency is generated by the cone, it will move first in a direction towards "B." The pressure at "B" will be increased while that at "A" will be decreased. Sound travels at 1,100 feet per second. The wave generated by the movement towards "B" will therefore travel at 1,100 feet per second in either a beam, a short angle or else will move around the face of the cone depending on whether the frequency is high, low or medium.

At the same instant that this pressure wave is created, the cone also creates a wave on its other side which is in exact opposite phase to that of the first wave. If the generated frequency is high, then it will be projected in a beam and the positive and negative half waves will not meet. As the frequency is decreased the waves will be radiated in accordance with Fig. 6 so that on the lower frequencies the positive wave will have time to get around to the back of the cone and cancel the wave toward "A." When this happens, of course, we hear nothing and this is exactly what happens when we do not use a baffle.

Likewise, it is evident that the effect of the baffle is limited to the lower frequencies; for on the higher frequencies the radiation is in an opposite direction to that of the negative wave due to the beam effect at these frequencies.

SUGGESTIONS

The problem of what to charge for service calls is as great today as ever. RADIO-CRAFT would like to hear from various Service Men who have tried different methods of charging the customer and found them to be successful.

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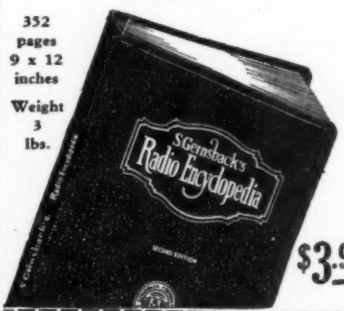

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A. F. CHOKE DESIGN

(*Continued from page 215*)

ductance of a transformer or choke. The factor $u_{a.c.}$ is defined as being equal to $\Delta B/\Delta H$ or the increment in B divided by the corresponding increment in H for a given A.C. flux. It may be readily seen that the value for $u_{a.c.}$ is something quite different from the arithmetical value for u obtained from the original B/H curve. This value is not really a constant as it varies over a range of A.C. flux densities. The relation of $u_{a.c.}$ to the A.C. flux is shown in Fig. 1D for two values of D.C. magnetizing force. These values are for a particular core material and are consequently not for design purposes.

With $u_{a.c.}$ determined for a given D.C. flux density the inductance of the winding may be calculated from the equation.

$$L = \frac{4\ \text{pi}\ A\ N^2\ u_{a.c.}}{10 \times 1}$$

It will be noted in Fig. 1D that the value $u_{a.c.}$ increases slightly for the higher values of A.C. flux density but the increase has the effect of slightly adding to the inductance of the winding and is therefore beneficial in its effect except where sharply tuned circuits are involved.

If an air-gap be provided in a magnetic core the D.C. *flux density* will fall off but an increase in the A.C. *permeability* will obtain. There is an optimum value for this relation which depends upon the characteristics of the particular iron chosen for the core material. That is to say: for some length of air-gap,—for some substitution of a material having unit permeability for a portion of the magnetic path the A.C. permeability will be at a maximum. Not only this, but the D.C. flux density may be varied over a wide range with but slight effect upon the inductance of the winding.

In most A.F. transformers a certain percentage of air-gap is necessary (although it is provided in the case of the highest quality transformers only). The air-gap may be dispensed with only when the D.C. flux density is at zero through the use of parallel feed systems. In push-pull transformers the D.C. components are assumed to cancel but there are high values of A.C. in the core and a certain amount of D.C. is always present due to tube irregularities so that an air-gap is still essential to an economical design.

In order to offer a series of chokes for the average experimenter to construct with a full knowledge of the characteristics that will obtain *it is first necessary that a particular lamination be chosen which will be available to him.* A core form which will offer a range in the neighborhood of 30 henries for all normal values of D.C. flux and which will at the same time lend itself to an economical design in so far as D.C. resistance is concerned may be achieved around the EI-11 lamination of the Lamination Stamping Company (located at Brackenridge, Pa.)

The dimensions of this core are as shown in Fig. 2. The characteristics of a core having a square cross-section are as follows:

Window area, .575-sq.-in.;
Volume, 4.03 cu. in.;
Core weight, 1.13 lbs.;
Area, 4.94 sq. cm. or .766-sq. in.;
Length of path, 15.61 cm. or 6.13 in.

Winding data from the charts is based on the use of enamel wire of the size indicated and with glassine paper between layers. The ends of windings should be straightened with pieces of fiber and the core itself should be protected with several layers of brown paper.

In forming the windings a piece of wood should be cut to slightly less than the core dimensions and wrapped with one layer of string. After winding, this string may be pulled out and the coil removed from the wood form. The values for the air-gap given in the chart refer to the total gap and in the case of EI laminations the gap at the center and in each leg should be just half the value read from the chart. Any machinist will lend you a micrometer to assist in building up wafers of paper or of fiber to insert in the legs. The holes provided in the core offer ready means for mounting and for clamping in position by means of iron or brass straps in methods which will readily suggest themselves to the experimenter.

Variations from the data in the charts may be achieved by varying the "stack" of the laminations. For double the stack (2 x .875-in.) the inductance will be doubled for a given current density and number of turns; while the resistance of the windings will be increased by but one-third. For one-half the stack specified, the inductance will be halved; and the resistance will be reduced by about 16 percent.

We may thus achieve with this single type of lamination any value of inductance up to 60 hy with D.C.; or up to over 200 henries where no D.C. is employed. The core material employed in calculating the charts was Allegheny Super Dynamo Grade Steel Sheets and the calculations are based on the assumption that 1 volt of A.C. is superimposed on the D.C.; for high A.C. flux densities a slightly higher inductance will prevail.

As an example, let us suppose that we desire a choke having an inductance of 20 henries at a current of 120 ma. on the core specified. Tracing through from the 20 henry line at the left edge of the curve we find that this cuts the 120 ma. line at 6,700 turns. We find from the wire size—resistance curve, that the wire size for this number of turns is No. 35 enamel; and that the resistance of the choke will be 1,100 ohms. Under these conditions the second chart shows that the air-gap is about .045-in. This means that .0225-in. of insulating material will be required in each leg of the choke core.

Now, an efficient choke should not have a resistance so high as this unless designed for a specific purpose—for instance, let us suppose that a receiver employing a 2,500-ohm dynamic field in its filter system must be used with an external speaker and we desire to replace the original speaker field with a choke. In this case we would work *backwards* from the specifications that our choke must have a 2,500-ohm resistance. On the basis of economy, let us see what we can do with our choke design in order to reduce the resistance.

As we noted before, increasing the height of the core stack by 100% increases the inductance of any form of winding by a like degree. On the basis of a square stack we start with an inductance of 10 henries at 120 ma. This calls for 5,000 turns of No. 33 wire, a total resistance of 450 ohms. Increasing the stack by 100% likewise increased the resistance by one-third. Our total resistance is therefore 600 ohms. Reference to the second chart indicates a total air-gap of .035 in. or .0175-in. in each leg.

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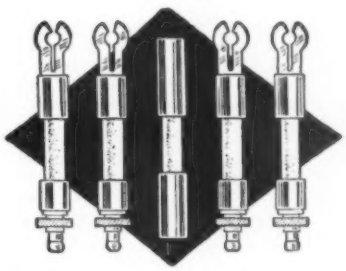

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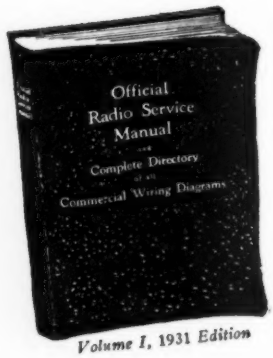

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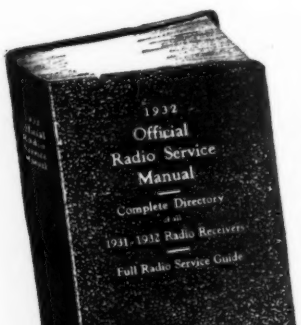

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ADAPTERS FOR TEST EQUIPMENT

(*Continued from page* 217)

214, 533, 534, 535, 536, 538, 540 and 597. The adapter is to be inserted in a 2.5 volt five-hole socket.

No. 955G-2, Fig. 17, is for testing the Majestic G-2-8 and G-4-8 tubes in the following tube checkers: Models 209, 210, 214, 533, 534, 535, 536, 538, 540, 597 and 675.

No. 975, Fig. 18, is for testing the types '36, '38 and '39 tubes in the 209 and 210 Jewell tube checkers.

No. 949PT, Fig. 19, is an output meter adapter for use in connection with Jewell 560 and 563 oscillators. Also used with the 444 and 660 set analyzers.

No. 954DS, Fig. 20, changes over a four-prong plug to a five-prong plug. Used with Jewell 444 or any set analyzer having a basic five-prong plug using the Na-Ald latch-lock feature.

No. 944FFL, Fig 21, is for testing the Kellogg "overhead heater" type tubes in the Jewell set analyzer comprising models 198, 199, 408, 409, 444, 577, 578, 581 and 660.

No. 954KGL, Fig. 22, is for testing the first section of a Triple-Twin tube. This adapter is to be placed in a four-hole socket of the tube tester; the '45 socket for the 295, and the 71A socket for the 293. May be used with analyzer models 209, 210, 214, 533, 534, 535, 536, 538, 540, 597 and 675.

No. 954KG, Fig. 23, tests the second section of the triple-twin tube. May be used in all the analyzer models listed for the No. 954KGL; operating directions are the same as for the No. 954KGL adapter.

No. 944PLCR, Fig. 24, is for testing the type 866 tube in tube checkers. This adapter incorporates a resistor to limit the plate current to prevent damage to meters. May be used with models 209, 210, 214, 533, 534, 535, 538, 540, 597 and 675. Insert in 2.5 volt socket.

No. 965PG, Fig. 25, will also test the types 41, 42, PA and PZH tubes. However, it is necessary to use the socket of the tester ordinarily used for testing the type '37 tube, or any five-hole socket wired for 6.3 volts.

No. 965-55, Fig. 26, is used for testing the 55 and 85 "duplex-diode triode" tubes. Place in a '27 socket to test the 55 and and in a '37 to test the 85. May be used with the following Jewell analyzers: Models 209, 210, 214, 533, 535, 536, 538, 540, 597 and 675, to check the 55 tube.

No. 955GGKL, Fig. 27, is for testing the five-prong Wunderlich "A" tube in the type '27 socket of testers. May be used with all Jewell instrument models described in connection with the No. 965-55 adapter.

No. 965GG, Fig. 28, is used for testing the types 29 and 69 tubes. When testing the 29, use a 2.5 volt (filament) socket; when testing the 69, use a 6.3 volt, five-prong socket. May be used with the following Jewell Models: 210, 214, 533, 534, 535, 536, 538, 540, 597 and 675.

No. 944BRR, Fig. 29, is used to test the Raytheon type BR tube which is made for use in auto "B" eliminators. It is to be placed in the type '80 socket of tube checkers. Included in the adapter is a current-limiting resistor of correct size. May be used with Jewell instruments: Models 209, 210, 214, 533, 534, 535, 536, 538, 540, 597, and 675.

This practically completes the data on adapters suited to the needs of Service Men who own Jewell test instruments; also, it concludes the first part of our article on Adapters. The author will be pleased, indeed, to hear from radio men who may wish to question any of the statements that have been made, or who would like to obtain any additional information regarding the design or use of these very useful units.

THE CRAFTSMAN'S PAGE

(*Continued from page* 238)

The writer knows of a number of instances wherein the one-ampere detector tubes have been substituted for the older types designed for the 1¾ ampere drain; furthermore, in nearly every instance where such procedure has been followed in the case of an old receiving set it has been possible to substitute a second tube of the new type in place of one of the old type '26 tubes (usually the first A.F. tube), and operate the heating feature of same from the filament supply winding of the detector tube without causing perceptible excess heating in the transformer. (Adapters do the trick; or the sockets may be changed, where necessary.)

With the above mentioned facts as a basis for further consideration of the subject I submit that it would be a profitable venture for the tube interests to offer, at an early date, a tube of the '27 type, designed to operate from the 1½ volt winding of transformers intended for type '26 tube supply, and with a drain of not to exceed the one-ampere demand of the present tubes, and also a screen-grid tube operating on the same current drain.

As a matter of information the writer had, at one time, some tubes of the heater type which were designed for operation on a 1½ volt supply circuit and which operated very satisfactorily for a long time.

C. M. Delano.
Lincoln, Neb.

(Although the proposal of Mr. Delano will strike a responsive chord in the hearts of many, we question whether mathematical computation could stand the strain of supplying identification figures!

At any rate, Mr. Delano has opinions—and that, certainly, is worth writing home about.

Since the ice has been broken, let us put on record the following suggestions: A line of direct-heater tubes to be known as the "portable series." It should include tubes of screen-grid, pentode and general-purpose type, but with an envelope no larger than is used for the Western Electric type N (No. 215-A) or "peanut" tube, which is the size of one's little finger; their filament is rated at 0.25-amp. and 1.1 volts—the proposed series should more nearly approximate our present "2-volt" series (60 ma. and 2 volts), or perhaps be designed for 1.1 volts supply but a current consumption of only 60 ma., or less.

What say—manufacturers, experimenters, engineers and Service Men?—*Editor.*)

THE "K" TUBE

(*Continued from page* 206)

milliamperes when ionization took place, and for this reason a limiting resistor must be used to prevent the self destruction of the tube.

The connections for the operation of a relay in a photoelectric cell circuit are shown in Fig. 3 using a single stage set-up of the type "K" tube. A limiting resistor of 3,200 ohms was used in the plate circuit while a 50,000-and a 15,000-ohm potentiometer were used to obtain bias for the grid. The potentiometer was used to set the operating bias. The grid leak to be used depends entirely on the photoelectric cell used, although, as in all gas content tubes, one to two megohms appear to be the maximum practical values. The relay used need not be sensitive, as from 20 to 30 milliamperes is available for operation. If more is necessary, it is obtainable by increasing the plate voltage. It is necessary to shunt the relay with about one microfarad of capacity in order to operate on the pulsating current.

Operation was secured on light inputs to the photoelectric cell of much lower intensities than with any previous amplifier tried. In fact, the single stage here proved to be the equal of the two stage direct-coupled amplifier described in the October, 1931 issue of Radio-Craft. The action was very positive, the relay closing with a snap.

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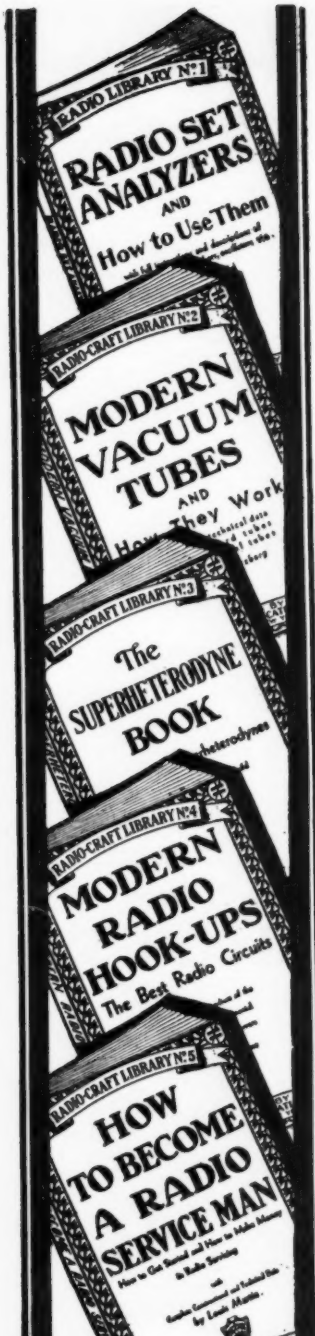
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The circuits cover the following: Broadcast Receivers, All-Wave Receivers, Short-Wave Receivers, Converters and Adapters, Television Receivers, Home Recording Apparatus, Automobile Receivers, Audio and Power Amplifiers, Power Units and Miscellaneous Equipment.

Book No. 5
HOW TO BECOME A RADIO SERVICE MAN
How To Get Started and How To Make Money in Radio Servicing
By LOUIS MARTIN

The ambition of many men in radio today is to become a first-grade Service Man. It is not as difficult as one might believe, but it cannot be done in a few short months. Following very carefully the advice of Mr. Martin, who has dealt with the problems of thousands of Service Men, this book deals very carefully with the essential stages in the preparation for qualifying as a Service Man.

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Read in this book by Mr. Denton, how easily you can modernize any obsolete set, and with little additional costs.

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If there is one subject that is fascinating to every radio man, it is that of Home Recording. Of course, this volume is not all on "Home" recording, but the information contained therein is important to commercial radio men, studio operators, engineers and others interested in this phase of radio.

The art of recording and reproducing broadcast selections is becoming more important every day to radio men, experimenters and Service Men. Equipping dance halls, auditoriums, churches, restaurants and homes with public address and amplifiers brings many extra dollars and often an excellent income.

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This supplementary laboratory course will prove invaluable to the radio student regardless of the branch of the business in which he intends to specialize, as the basic principles underlying the fields of servicing, television, sound pictures, aircraft radio and commercial operating are the same. For example, the sound projectionist, just as well as the Service Man, must understand the operation of vacuum tubes and the testing of circuits, and both must have a thorough understanding of the generation and control of electricity.

The course is divided into five sections of ten experiments each. The first section starts with the simple detection of electric currents by means of an earphone, and proceeds through the following further tests: to show how the presence of an electromotive force can be detected by the eye; to show that an e.m.f. has polarity; to show that a magnetic field is always associated with current flow; to show that a greater magnetic field is produced by direct current flowing through a coil wound on an iron core than by the same current flowing through an air core coil, to show the different conductivities of different materials; to illustrate the effect of electro-magnetic induction; to show how electricity is generated by chemical action; to show the action of the storage battery; to show mechanical generation of electricity.

The second booklet deals specifically with vacuum tubes, how to test them and how to record their behavior under various conditions of filament, grid and plate voltage. The construction of several simple but useful tube testing instruments is described and illustrated.

In the third instalment the student is taken into the more advanced subjects of the generation of low-frequency oscillations, the control of frequency, the dynatron oscillator, the modulated R.F. oscillator, the phenomenon of resonance, etc.

The subject of detection or "demodulation" is treated at length in the fourth booklet. Various types of detectors and detector circuits are considered and the features of each described. An excellent modulated oscillator can be made with the apparatus furnished to the student of the course.

The fifth and last booklet expands generally on the subjects previously covered. The usefulness of the vacuum-tube voltmeter is explained, as are numerous methods for using it. The experiments cover the following subjects: the action of the A.F. output transformer; audio amplification with transformer and resistance-capacity coupling; building a regenerative receiver; adding R.F. amplification; and the effect of tuning the aerial circuit.

This course is well prepared and clearly written, and is altogether devoid of the mathematics that confuses many practical radio workers.

HOME EXPERIMENTS

(*Continued from page* 234)

which you are conducting the experiment, will be sufficient to cause a minute electrical current to flow and register upon the meter. However, when the full force of the rays of a flashlight are flashed on the cell, a decided deflection of the meter needle will be observed, denoting the fact that due to a chemical action within the cell an appreciable electrical current has been caused to flow in the photoelectric cell circuit. To demonstrate how variations in the size and intensity of the beam of light can change the current flow in the cell circuit, it is suggested that the experimenter cut various sized holes in a piece of cardboard and then shine the light from the flashlight through them, noting the while, the change in current indicated on the meter. Also, pieces of smoked glass of varying density may be held before the beam of light to cause changes in the flow of current. Other experiments will suggest themselves to the experimenter. The practical knowledge to be gained from this rather simple experiment is in the parallel which can be drawn between it and the rather elaborate processes involved in the production of sound from film as used in talking movies.

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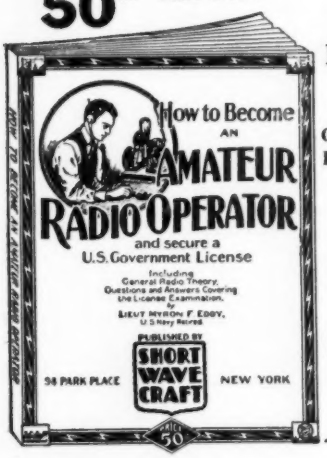

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CHAPTER 2. Concise, authoritative definitions of radio terms, units and laws, brief descriptions of commonly used pieces of radio equipment. This chapter gives the working terminology of the radio operator. All graphic symbols used to indicate the various parts of radio circuits are shown so that they may be readily recognized when studied in the following chapters.

CHAPTER 3. General radio theory particularly as it applies to the beginner. The electron theory is briefly given, then waves—their creation, propogation and reception. Fundamental laws of electric circuits, particularly those used in radio, are explained next and typical basic circuits are analyzed.

CHAPTER 4. Descriptions of modern receivers that are being used with success by amateurs. You are told how to build and operate these sets, and how they work.

CHAPTER 5. Amateur transmitters. Diagrams with specifications are furnished so construction is made easy.

CHAPTER 6. Power equipment that may be used with transmitters and receivers, rectifiers, filters, batteries, etc.

CHAPTER 7. Regulations that apply to amateur operators.

CHAPTER 8. Appendix, which contains the international "Q" signals, conversion tables for reference purposes, etc.

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S. W. LABORATORY

(*Continued from page* 231)

5,000, 10,000, and 20,000 ohms, thus providing an excellent reading range of .1- to 20 volts. The variation of the proper grid voltage for each resistor is accomplished by the 20,000-ohm potentiometer R3. Needless to say, the accuracy of the reading depends upon the accuracy of the voltmeter resistors to a large extent.

Shown with the diagram is the switch SW3 which is closed during operations. When this switch is open, terminals 15 and 13 are used to test for resistance values. This will be explained in the following section.

The 30-type tubes are used in all the individual units because of their low plate- and filament-current consumption. The "A" battery is a 4.5 volt "C" battery. Resistors of 41 ohms are in series with each tube filament.

Tests and Measurements

RESISTANCE: The usual resistance test is good for one particular range of resistance, depending on the size of the battery. A 4.5-volt battery will give useful indications up to 100,000 ohms, but not above that, and none below 1,000 ohms. Increasing or decreasing the battery voltage will result in different satisfactory ranges; but for any one battery voltage there will be upper and lower limitations.

The feature of this unit is that almost any range from 50 ohms to 1 megohm can be easily calculated. The potentiometer R3 provides a means of obtaining a variable voltage from 1, to 20 volts. The meter, MA1, may be connected into any range by connecting, for example, the 10,000-ohm resistor R6 so that the full-scale deflection of MA1 is 10 volts. The potentiometer is now adjusted to give a full-scale deflection. The resistance testing terminals are Nos. 13 and 15 (switch SW3 open).

Suppose that while testing a resistor, the meter MA1 shows a deflection of .5-ma. The applied potential being 10 volts, the total circuit resistance is 20,000 ohms. However, in the voltmeter circuit there is a resistance of 10,000 ohms (R6) and the value of the resistance under test is therefore 20,000 minus 10,000 or 10,000 ohms. The same procedure is followed for any range. Select the range by plugging into the proper resistance. Adjust the potentiometer R3 to give full-scale deflection at that particular range. Read the scale reading of MA1 in milliamperes when testing, and determine the value of resistance as described above.

CAPACITY: The capacity test is not as simple as the resistance test as it will be necessary to first determine the capacity of several condensers and then make a permanent record or graph of the readings.

In this method it will be necessary to know the current drain of the copper-oxide voltmeter (now composed of MA2 and either R11, R12, R13, or R14) at any point. This may be obtained by dividing the applied A.C. voltage reading at any point by the series resistance; for example: The A.C. indication is 75 volts. If the 150-volt A.C. range is desired, a series resistance of 75,000 ohms is required (R13) with the particular rectifier used; therefore, the current taken by the voltmeter is

$$I = \frac{E}{R} = \frac{75}{75{,}000} = .001\text{-ampere}$$

at that point.

The voltage drop across the condenser is obtained by first reading the line voltage *without* the condenser, then reading the voltage *with* the condenser in series, and finally subtracting the second reading from the first; for example, the voltage Ec across the condenser to be tested may be determined from the relation

$$Ec = \frac{I \times 1{,}000{,}000}{6.28 \times f \times C}$$

and the capacity in mf. equals:

$$C = \frac{I \times 1{,}000{,}000}{6.28 \times f \times Ec}$$

where Ec is the difference in voltage reading; f, the line frequency; C, the capacity in mf.; and I, the current at the *second* reading.

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Although so many calculations may seem tedious, it must be borne in mind that once they are made, they are permanent. The capacity is, of course, plotted against the *second* reading. This procedure is the same for any set of frequencies and voltage. The total voltmeter resistance is easily known because if the rectifier is self-constructed, the required series resistor can be measured. Where the rectifier is of commercial manufacture, the series resistors are specified for each range. A manufacturer's curve usually accompanies the rectifier. The particular rectifier used was made so as to utilize the resistances indicated in the wiring diagram. The resistance sizes will vary with each type rectifier, so do not accept the values given as standard.

INDUCTANCE: Having, in all probability, already calculated the current drain of the voltmeter when making the capacity curves, the computations for inductances will be considerably shortened. The procedure for testing inductances is the same as that for capacities. Obtain readings before and after inserting the choke; then calculate the difference. In the case of inductance, the drop across it is due to *both* the *reactance* and *resistance of the* winding. Measure the D.C. resistance of the choke before testing. The formula used to calculate the inductance is:

$$L = \sqrt{\frac{EL^2 - I^2 R^2}{I^2 \times f^2 \times (6.28)^2}}$$

Where EL is the difference in voltage readings; I, the current at any point (second reading); R, the D.C. resistance of the choke; f, the line frequency; and L, the inductance in henries.

POWER: Power measurements may be made by taking a known resistance and obtaining the voltage drop across it. This is an excellent way of computing power output of R.F. and A.F. circuits. For example, let us calculate the power output of the modulator: Connect a 2,000-ohm resistor across the output terminals. Connect the input terminals of the V.T. voltmeter across the resistor. Actually, then, the A.F. output, the resistor, and the V.T.V.M. input terminals are in parallel. First adjust the V.T.V.M. as outlined in its operation procedure; start the modulator and then obtain the D.C. voltage change by varying R3; multiply this by .707 to get the R.M.S. value. The power output P equals $\frac{E2}{R}$, where E is the R.M.S. value of voltage.

General Construction

The components have been so arranged that many connections can be made directly to the lugs on the parts themselves. The sockets are placed so direct connections are possible to the condensers or resistors. Since there are a large number of parts within a small space, exact mechanical and physical spacing is necessary. Take the time and care necessary to build a good piece of apparatus. If the copper-oxide rectifier is to be constructed at home, test the individual junctions and choose those having the highest forward-backward resistance ratio.

The simplest manner to make the shunts R15 and R6 is to wind a length of fine, copper wire, about No. 30 or No. 32, the resistance per foot, for which, may be found from wire tables. Measure the proper length of wire and wind it on any form. Errors of one or two feet in length of wire will be negligible, whereas high-resistance wire must be absolutely calibrated before being used.

Individual shields are placed around the oscillator coil and the tuning condenser. The shield for the condenser is short and constructed so as to fit into the case. Both shields are grounded to the "A—" terminal. The output coil should be wound next to the plate end of the oscillatory tank coil. The spacings for all coils is not critical, being between ⅛-in and ¼-in. Contact to the external test posts can be made only by the bi-polar switch. Before testing be certain that the switch is in its proper place.

(Continued on following page)

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Also accessories such as High Frequency Buzzers, Audio Oscillators, Differential Current Transformers, Two Stage Audio Amplifiers, etc.

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Write for pamphlet giving history and theory on this interesting subject. All above circuits are described and explained. List of parts and prices included. (Price of pamphlet 10c.)

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Address .....

State .....

S. W. LABORATORY

(*Continued from preceding page*)

Parts List

Two megohm grid leaks, R1, R19;
One 50-ohm resistor, R2;
One 20,000-ohm potentiometer, R3;
Two 1,000-ohm resistors, R4, R11;
One 5,000-ohm resistor, R5;
Two 10,000-ohm resistors, R6, R8;
One 20,000-ohm resistor, R7;
One 100,000-ohm resistor, R9;
One 500,000-ohm resistor, R10;
One 6,500-ohm resistor, R12;
One 75,000-ohm resistor, R13;
One 300,000-ohm resistor, R14;
One .333-ohm resistor, R15;
Two 3 1/3-ohm resistors, R16, R17;
Three 41-ohm resistors, R18, R20, R21;
One 30-ohm resistor, R22;
One 1.-megohm resistor, R23;
One audio transformer, any make, ratio 3 to 1, T1;
One input push-pull transformer as described; A.F.1;
One R.F. choke as described, R.F.C.;
One copper-oxide rectifier;
One .00006-mf. fixed condenser, C1;
One .00005-mf. variable condenser, C2;
Three .0001-mf. fixed condensers, C3, C4, C5;
One .002-mf. fixed condenser, C6;
One .02-mf. fixed condenser, C7;
One .005-mf. fixed condenser, C8;
One .2-mf. fixed condenser, C9;
One .1-mf. fixed condenser, C10;
One S.P.D.T. Toggle switch, SW1;
Two S.P.S.T. Toggle switches, SW2, SW3;
One 10 point bi-polar switch, SW4;
Two Jewell, type 301, 0 to 1. ma meters, MA1, MA2;
Two tube shields;
Four tip jack plugs;
One tip jack;
One carrying handle;
Three type '30 tubes;
Three 22½-volt B batteries;
Two 4.5-volt C batteries;
One dial and escutcheon;
Twenty-seven binding posts with insulated top.

REFLEX RECEIVER

(*Continued from page 239*)

One Trutest 4-prong plug, 26A;
One Trutest Full-Vision Dial;
Two Trutest Equalizers, capacity 2 to 35 mmf. 7A, 17A;
One Trutest Binding Post, 1;
One Amperite Self-Adjusting Line Voltage Control, type 5A-5, 33;
One I.R.C. (Durham) 500-ohm Metallized Resistor, ¼ watt, type M.F. 4½, 14;
Two I.R.C. (Durham) 25,000-ohm Metallized Resistors, 1. watt, type M.F. 4, 13, 24;
One I.R.C. (Durham) 1. meg. Metallized Resistor, 1, watt, type M.F. 4, 19;
One Arcturus type 56 Tube, 23;
One Arcturus type 58 Tube, 10;
One Arcturus type PZ Pentode Power Output Tube, 4;
One Arcturus type 180 Full Wave Rectifier Tube, 28;
One Aluminum Chassis, 1/16-in. material, 10 x 6 x 2-in. high;
One Wright-DeCoster No. 255 Dynamic Reproducer with 2,500-ohm field and output transformer to match PZ or '47 Power Tube, 27.

REACTIVATOR

(*Continued from page 221*)

in your tube checker. If it is not then O. K., return it to the rejuvenator and flash high voltages to the grid and plate *while still using 5 volts on the filament.* First flash the grid by pressing switch SW2 a number of times, then flash the plate by pressing switch SW1 a number of times. When the desired color has been produced, press both switches simultaneously a number of times and then test the tube in the tube checker. If the tube does not now show a desired improvement, then flash at higher voltages. Flashing must be done by very short, snappy contacts until the desired color is obtained, hence the need of especially quick-acting switches.

As a general rule, the 600-volt tap is used only on type '50 tubes, except after having definitely proven that 400 volts will not do the work; of course, 1,200 volts are used only on '50 tubes. Higher voltages can be used on transmitting tubes, from two to three times their maximum plate operating voltages being required. This is being done at some amateur stations.

If flashing is continued after the blue-green color has arrived, this color will grow weaker and so will the tube.

A large percent of '26 tubes will come back to full life faster than probably any other type of tube, and this is well, for they are the "weak sisters" among the A.C. tubes.

The '12 and '71 tubes will not rejuvenate by this method on account of the type of filament, but '12A and '71A respond readily.

As may be noted by reference to the photographs, only a single 4-prong socket is mounted in the rejuvenator; an additional adapter is used for 5-prong tubes, as shown in Fig. 4. The circuit diagram of Fig. 1 shows connections for the use of two sockets. Merely leave one out if an adapter is to be used.

(The following results were obtained in tests conducted by RADIO-CRAFT:

Practical test of		Before (in ma.)	After (in ma.)
1 '80		25	50
(one read lower than the other; now, both alike)			
8 '24		3-5	5-10
1 '12A	(not '12)	3-4.5	5-10
1 '71A	(not '71)	15-17	20-25
1 '35		dead	6-10
4 '26		dead	par
(as specified on a Weston test unit)			
6 '45		20-0 (par)	20-6 (average)
1 '10		40-3	40-8

List of Parts

One power transformer, either special wound or any having 2.5, 5, and 7.5-volt filament windings and 750 or more volts on the H.V. secondary, at 100 ma.;
One four-hole tube socket;
One five-hole tube socket;
Two four-point rotary switches, SW3 and SW4;
One A.C. toggle switch, SW5;
Two special switches SW1 and SW2;
One A.C. voltmeter, 0-7½ volts (not absolutely necessary);
Two phone tip jacks;
One metal or wood box;
Five feet of double light-cord and plug;
One bakelite panel, 7 x 8 in.

TUNING METERS

(*Continued from page 240*)

When used in R.F. or I.F. amplifier circuits, they may be connected as in the sketch of Fig. 2. They are connected in the plate return lead of four tubes, the normal plate current of each being about 1. ma. If such is the case, the full scale reading of the meter should be about 5 ma. so as the signal reduces the plate current, the pointer travels toward zero. In other words, *the full scale deflection of the meter should equal the sum of all the currents of all the tubes which pass through the meter.*

In the case of the grid-leak and grid-condenser detector, the range of the meter should be about 1. ma. since the reading decreases as the applied signal increases, similar to the case of amplifiers with A.V.C.

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(While every precaution is taken to insure accuracy, we cannot guarantee against the possibility of an occasional change or omission in the preparation of this index.)

OPERATING NOTES

(*Continued from page 225*)

which was covered with rosin. The tip of a hot iron soon remedied this condition.

This bad joint rendered the control-grid circuit of the A.V.C. tube open, thus causing the tube to draw so much plate current as to pass an excessive control-grid bias on the tubes which it controlled, producing the effect as described.

General Motors 252, etc.

A large number of General Motors model 252 receivers, a ten-tube superheterodyne, have been serviced because of an annoying, intermittent buzz-buzz that rode in, at regular intervals, above the signal level. As the tubes checked perfectly and the interference ceased when the aerial and ground wires were disconnected and the local distance switch was set at local, shorting aerial and ground, it appeared that the noise was external. But after listening to several receivers in the same building, where no interference of this nature was picked up, it was evident that the buzz-buzz was caused by a tube or some part of the set. Each tube was substituted with one from the kit and when the first-detector '24 was replaced, the interference was ended! The original tube had responded satisfactorily to all meter tests, yet, had caused the receiver to be noisy.

Of the four '24 tubes carried in the kit that were tried in this stage, another was found that produced the same noise. Because of the design of this receiver, it is essential that a tube of the correct characteristics be used as the first-detector.

Brunswick 42

The automatic mechanism of the phonograph part of the Brunswick 42 combination, though relatively simple, still furnishes cause for complaint. In some cases, it was found that as soon as the record had been lowered into position and the pickup head had come down upon the records, the mechanism rejected the record before it was played and then another record was lowered, after which this sequence of actions was repeated.

This is usually caused by the cycle switch, whose blades or contacts do not open when the cycle is ended. It is only necessary to loosen the two screws holding this switch in position and adjust the switch so that the contacts will disengage at the correct moment.

This same defect will not permit the phono mechanism to be switched off, without pulling out the line plug and the mechanism will continue to reject until this is done.

On other occasions, the mechanism will not reject the records, but repeats the same one. This is due to the fact that the contacts on the small switch (that travels along with the tone-arm) do not open because they have been set too close together. Here, these contacts must be so adjusted that when the tone-arm and pickup head reach the end of the record, the plunger or trigger of the switch will open the contacts when the former engages with the stop. Either the stop will have to be brought in further by loosening the screw and re-setting, or the tension on the spring contact must be reduced by loosening the two nuts on the screw that controls the tension.

Automobile Interference

Many paragraphs have been written upon the suppression of noise in automotive radio receivers, but there still arise cases, where everything known has been tried with little success. It is quite possible that this interference may be eliminated by having the motor and all wires and parts beneath the hood thoroughly cleansed and washed, especially the wires. Just this, has eliminated noise that for many weeks had been combatted without any results.

The grease and oil as well as road-dust, covering the wires and motor, contain a certain percentage of carbon and graphite which will cause many high-frequency leaks and losses, increasing noise and impairing motor performance.

3 Valuable New Books

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Radio — Mathematics

Presented here are three brand new books, of diversified nature, which are most educational and instructive. Each book carefully treats a subject in full, and is prepared by an expert in the field. The books are well written and thoroughly illustrated to make the contents easily learned. Get these books promptly—mail coupon below.

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This book has been compiled by S. Gernsback, a well-known author of practical instructional manuals in various scientific fields. It is extremely helpful and will show you how to save money by making in your own home, at a fraction of the regular cost, the hundred and one preparations which you now buy for use at home or business.

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2. FUNDAMENTAL PRINCIPLES OF RADIO
Radio Simply Explained—Its Origin, Nature and Functions

The book, written by Louis Martin, has been prepared with special consideration given to young members in the radio profession, and those who have gained their experience in a haphazard fashion. This radio primer is a handy fundamental aid for "checking up" and systematizing your knowledge of radio. Regardless of how much you know about the subject, you should read this book.

HERE ARE PARTIAL CONTENTS:
Chapter I—Fundamentals of Radio: Electricity, Resistance, Batteries, The Magnetic Circuit, The Magnetic Field, Inductance, Condensers, A.C. Circuits, Propagation of Radio Waves; Chapter II—The Simple Radio Set, Single, Two, and Three-Circuit Tuners, The Battery Set, Vacuum Tubes, Electric Sets, Loud Speakers; Chapter III—Diagrams, How to Read Them; Chapter IV—Amateurs and Broadcast Stations, Talking Pictures, Television.

3. ELEMENTARY MATHEMATICS
for the Technician and Craftsman

This manual has been especially prepared for the man who wishes to acquire a working knowledge of the elementary principles of mathematics. A complete treatment of the subject is given by the author, Mr. C. Shainmark, with special attention to the use of mathematics in Radio and other technical work for those who employ its formulas daily.

HERE ARE PARTAIL CONTENTS:
I—Arithmetic: Addition, Multiplication, Subtraction, Division, How to Use Decimals; II—Fractions, Percentages, Ratio and Proportions; III—Powers and Roots; IV—The Metric System; V—How to Measure Surfaces and Volumes; VI—Mathematics for the Manual and Technical Craftsman; VII—Special Mathematics for the Radio Technician; VIII—Commercial Calculations, Short-cut Arithmetic, Interest Calculation, Discounts; IX—Weights and Measures; X—Useful Tables.

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67. RICH REWARDS IN RADIO

No one can read this heavy 64-page book, printed throughout in two colors, without realizing that the radio field is a rich and large one. It points out in convincing fashion that there are many opportunities in set servicing, broadcasting, commercial radio operating, aviation, radio and television, and it describes the excellent courses of home instruction in these subjects that the National Radio Institute has available to ambitious men of all ages. Dozens of testimonials scattered throughout the book indicate that the instruction is highly successful in training capable radio experts. *National Radio Institute.*

58. THE UPPCO A.C. CAR GENERATOR

The problem of furnishing power to the automobile radio receiver or portable public address amplifier has been attacked from several angles by interested manufacturers. The general trend seems to be toward the use of the car's regular storage battery for filament supply and of "B" batteries, rotary converters and vibrator type power units for plate supply. However, the manufacturers of the Uppco Car Generator go a step further and provide a double generator which replaces the present charging generator in the car and supplies six volts D.C. for regular automotive purposes and also 110 volts, 60 cycles A.C. for radio purposes. With this power supply, the advantages of standard A.C. receivers, amplifiers and associated equipment may be enjoyed with a minimum of operating trouble and expense. The generator is available in three models, having outputs of 100, 150 and 300 watts, respectively. The possibilities and advantages of the arrangement are obvious. *Upp Electric Company.*

69. ARCO PHOTO CELLS AND TELEVISION LAMPS

The increasing importance being assumed by photo-electric cells for a wide variety of radio, electrical and industrial applications makes it desirable for Service Men and experimenters to develop a close acquaintance with them, for eventually they will encounter them in their everyday work. For this reason the Arco bulletin is worth reading, as it contains some readable "dope" on the caesium type cell and includes a hook-up showing how it is used. The back of the sheet also includes a handy reference list of some 60 tubes, many of which are no longer considered standard and cannot be obtained from other sources of supply. *Arco Tube Company.*

Radio-Craft READERS' BUREAU

On this page are listed booklets, catalogs, pamphlets, etc., of Manufacturers, Schools, Institutions, and other organizations, which may be of interest to readers of "Radio-Craft." The list is revised each month, and it will be kept as up-to-date and accurate as possible. In all cases the literature has been selected because of the valuable information which the books contain.

This Service is absolutely free to all Readers of "Radio-Craft."

Fill in and mail the coupon below; make sure that your name and address are included and are plainly written. Order by number only.

70. NOISE-REDUCING ANTENNAS

This four-page reprint of an article on noise-reducing antennas for short-wave reception is very interesting and contains a lot of how-to-do-it material that will be appreciated by practical Service Men and short-wave fans. It deals with the use of special aerials of the so-called "doublet" type, with transposed-wire lead-ins or "feeders" which themselves do not act as aerials but serve only to connect the actual aerial to the receiver. Since much of the noise that affects short-wave reception is picked up by the ordinary lead-in (because of its proximity to electrical machinery in buildings), this arrangement is noticeably quieter than conventional antenna systems, and is therefore well worth installing in notoriously noisy districts. *Lynch Mfg. Co.*

71. BRUNSWICK REPLACEMENT PARTS

Service Men who have occasion to repair Brunswick receivers, which were widely distributed at one time, will be glad to learn that replacement parts for some sixteen models are now available from a reliable organization. This firm has published a seven-page list of conveniently indexed parts, for the following Brunswick model numbers: 14, 21, 31, S-14, S-12, S-31, 15, 22, 32, 42, 11, 12, 16, 33, 17, 24 and 25. The number of each part, a description of it and its price are all given. *United Radio Service Co.*

72. FOX IMPROVED SOUND EQUIPMENT

The three technical bulletins covered by the name above are of special interest to individuals or firms engaged in public address or theatre sound system work. The first describes a new all-aluminum trumpet horn and an adapter which permits two or three horns to be used with one dynamic unit. The second describes a heavy duty electro-dynamic speaker unit having a capacity of 25 watts. The last bulletin describes various exponential horns for theatre and other indoor uses. *Fox Electric and Manufacturing Co.*

73. AUTO DIAL RADIO

This folder describes a powerful auto radio receiver using a '47 pentode and having an output rating of 2½ watts. It uses regular A.C. tubes of the 24, 35 and 27 varieties, which are easily replacable at low cost. The speaker is of the full dynamic type. The whole outfit is unusually compact and may be installed with a minimum of tool work on the car itself. *J-M-P Manufacturing Company.*

74. THE LINCOLN DE LUXE 33

Radio men who have become slightly fed-up on tiny midgets which have just enough parts to make them produce signals will read this description of the Lincoln Model 33 receiver with relish. Here is a 12-tube all wave receiver, tuning from 15 to 550 meters, with all the features that advanced radio engineering technique has made possible at the present time. It is a superheterodyne, of course, with automatic volume control, visual tuning, twin grid detection, double push-pull audio, etc. *Lincoln Radio Corp.*

RADIO-CRAFT 67-74
Readers Bureau
96-98 Park Place, New York, N. Y.

Please send me *free of charge* the following booklets indicated by numbers in the published list above:

No. .....

.....

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Address .....

City..... State.....

"Round the World" Reception!

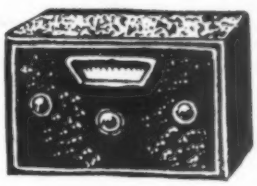
World wide reception with the new "Powertone" battery operated shortwave receiver is an accomplished fact. Letters from enthusiastic customers verify the fact that they have received more distant stations on this receiver than on any other two or three tube receiver.

It requires no exaggeration to create enthusiasm for the new "Powertone" two tube short wave receiver. The "Powertone" short wave receiver handles the broadcast band as well as the short-waves.

It's fine construction (which includes a main .00015 Hammarlund condenser) gives you complete assurance, of dependable round the world performance throughout the years to come.

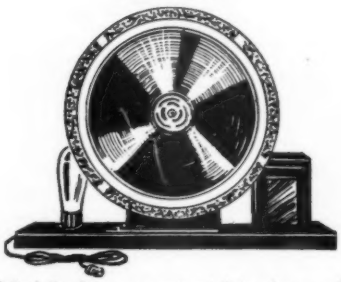
The tubes used in this receiver are 1-232 and 1-233 which are used advantageously and economically in this receiver.

The tuning range of the "Powertone" receiver is from 14 meters up to broadcast band with plug-in type coils.



Price which includes four coils.....RC 504A Our Price $9.95
Price complete with set of matched tubes.....RC 504B Our Price 12.50
Set of full sized batteries for above receiver..... 3.50

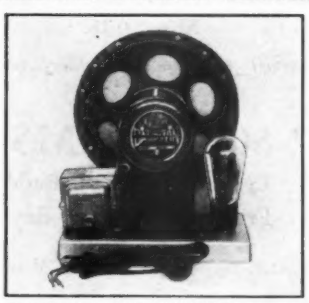
OXFORD DYNAMIC SPEAKER CHASSIS



Oxford Speakers are constructed to give reproduction as near perfect as possible—they have long been known to give radio's finest tone qualities. The most advanced principles of speaker design for dynamics have been incorporated in Oxford Speakers. Those that bear recognition are: the three-point suspension balance; permanentized controlled diaphragm; moisture-proof coil mountings; and inertia counteracted construction.

A.C. Models—14" Auditorium Speaker, using 280 rectifier.....No. RC 177—$10.00
12" Concert Dynamic, using 280 rectifier, No. RC 178— 8.50
9" Midget Dynamic, using 280 rectifier, No. RC 179— 7.50
D.C. Models — 14" Speaker, 5000 ohm field ....No. RC 180— 8.95
11" Speaker, 2500 ohm field ....No. RC 181— 4.95
9" Speaker, 2500 field ....No. RC 181— 2.45
All 9" Midgets are supplied with hum balancer.

BALDWIN DYNAMIC CHASSIS

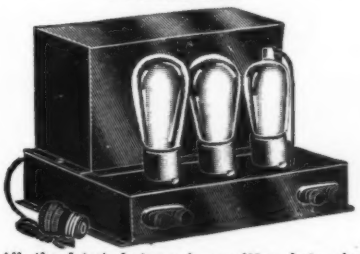


The Introduction of the BALDWIN name need not be further discussed.

From the beginning of Radio, Baldwin has been a pioneer in radio reproducing equipment. We need no further praise in recommending a Baldwin Speaker for any purpose in regard to radio.

A.C. using 280 Tube Rect., No. RC 410—Our Price $6.95
D.C. 2500 Ohm—110 Volts, No. RC 411—Our Price $4.95
D.C. 2500 Ohm—110 Volts Field, Less Stand, No. RC 412—Our Price $4.00
D.C. 2500 Ohm—110 Volts Field, Less Output Trans.....No. RC 413—Our Price $3.25

POWERTONE DIRECT COUPLED AMPLIFIERS



All the latest features in amplifier design have been incorporated in this direct coupled amplifier. Input terminals for permitting phonograph attachment, and in addition, by attaching the proper microphone transformer, also dry cell batteries, a microphone may be used for public address work. Any type of receiver can be connected to the input terminals without fear of loss of energy. These amplifiers are for use with dynamic speakers.

Amplifiers supply field current of 2500 ohms to dynamic speakers. Resistors are furnished when D.C. speakers are used. Amplifiers operate on 110 volts, A.C., 60 cycles.

Model RC 245 uses 1—224, 1—245 and 1—280 Output is rated at 3 watts.
No. RC 245—Our Price $10.95 (less tubes)

Exactly the same amplifier, but employs a power pentode in place of the 245 tube. Output is rated at 2.5 watts.
No. RC 246—Our Price $12.95 (less tubes)

Model RC 250 uses 1—224, 1—250 and 1—281. Consumption is 85 watts; maximum undistorted output, 6 watts. Gain rated at 100 cycles, 55db; gain at 1000 cycles, 52.5 db; gain at 10,000 cycles, 51.5. Input direct to 224 control grid; output direct from 250 plate.
No. RC 250—Our Price $15.25 (less tubes)

R.C.A. VICTOR HAND MICROPHONE

This microphone is furnished with the latest model Radiola 86 Home recording Super-Heterodyne Receiver and Amplifier. Do not confuse this microphone with the "toy" type as it can be used for many professional and commercial uses. It is a single button "mike" with a gun metal finish. Its total length is 6½" and has as standard equipment a four foot cord.

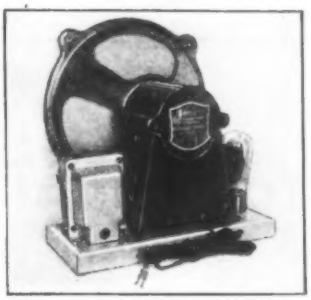
No. RC 417—Our Price $2.75

Franklin Transmitting Keys and Buzzers

Practice Set

For those who want to practice code, sending and receiving, this small practice set is recommended. A small dry cell is required to operate buzzer, otherwise it is complete.
No. RC 300—Key only, 95c
No. RC 301—Buzzer only, 69c

JENSEN DYNAMIC SPEAKER CHASSIS



The tone and quality of these reproducers are unsurpassed. Will reproduce faithfully under the most trying conditions. These speaker outputs are available for all new type tubes as well as the old.

A.C. Types, D-7 Jr. Auditorium (as pictured) uses 280 rectifying tube in conjunction with 8 mfd. electrolytic condenser which insures minimum hum.

JENSEN—
A.C.—D.8 Dry Rectifier .....No. R 184—$14.95
A.C.—D.7 Jr. Audit Tube Rectifier..No. R 185— 9.95
A.C.—D.15 Concert Jr. Tube Rect..No. R 186— 7.50
D.9—2500 Ohm D.C. Field.....No. R 187— 6.95
D.7—2500 Ohm D.C. Field.....No. R 188— 7.50
D.15—Midget, 2500 Ohms.....No. R 189— 4.75

FARRAND INDUCTOR DYNAMICS

Farrand Inductor dynamic, the finest speaker available anywhere for A.C., D.C., or battery operated sets.

Will take as much power and reproduce in tone and quality as the finest dynamic speaker will.

In buying a FARRAND INDUCTOR DYNAMIC you are making a safe investment, because these speakers will match any of the new output tubes as well as all the old type output tubes.

A special feature of the FARRAND INDUCTOR DYNAMIC is that it has a freefloating cone, which tends to bring forth the speech and music with true fidelity. Absolutely will not blast or rattle.

Please state type of power or power tubes when ordering.
9 inch model.....No. RC 192—$5.95
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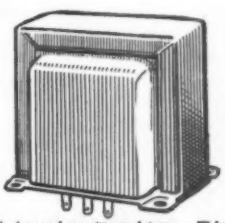

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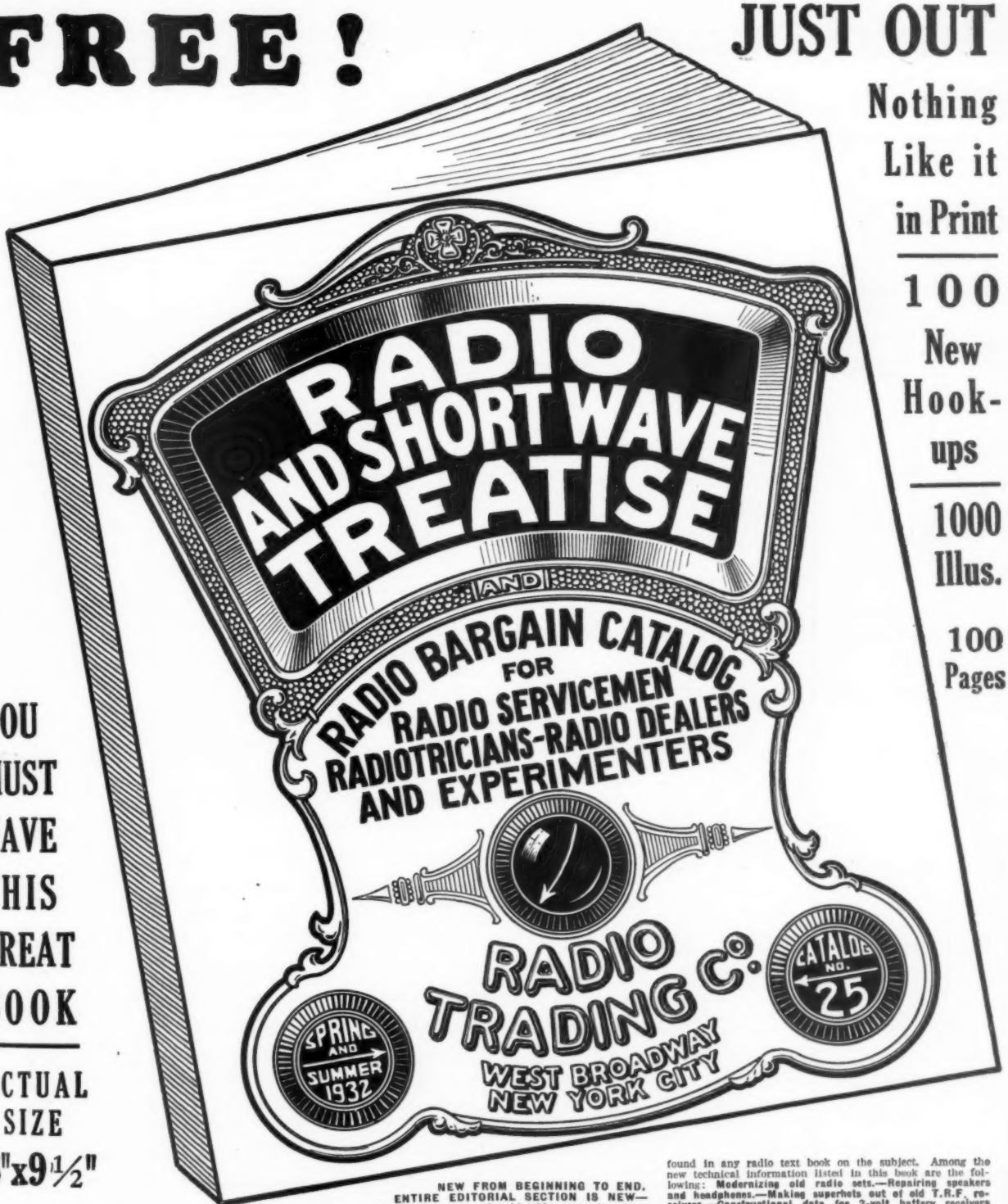

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OVER-THE-COUNTER SUGGESTIONS

By JACK GRAND

In this department Mr. Grand passes on to RADIO-CRAFT readers many new thoughts and ideas he encounters in his daily over-the-counter contact with radio technicians who come from all over the world to one of New York's oldest retail radio establishments.

The whistles blew and the bells rang,
And the door opened with a bang.
There was I, leaning "Over the Counter"
Offering suggestions to an old "Brass Pounder."

Anyhow, a customer came in for an '80 tube and I suggested an 82, because the 82 tube is becoming very popular as a replacement tube for the '80 and '81 types of rectifiers, as it is more efficient than either of these two tubes, especially on heavy drain sets, and is less costly than the latter. However, to obtain best results with this new rectifier, a tube shield is required; also, a small choke of about 2 millihenries, with a current capacity of 250 ma., must be placed in series with the common lead of the power transformer as shown in Fig. 1A. The resistance of the choke must be low; approximately 50 ohms.

When using the 82 to replace the '80 tube, I would suggest that an 8-ohm resistor be used in one of the filament leads to cut the potential to 0.5 volt. Although the current requirement of the 82 is 3 amperes, the filament winding for the '80 will stand the overload.

In the case of the '81 type tubes, the secondary winding must be checked to see that no more than 500 volts is obtained on each half of the winding; then, the filament winding must be checked. If the transformer is made by a good manufacturer, then the 7.5-volt winding will stand the overload; otherwise, I recommend a small 2.5-volt transformer rated at 6 amps. at a cost that will still effect quite a saving. Figure 1B shows how to connect an 82 to replace one '81; and Fig. 1C, two 82's to replace two '81's, both using 7.5-volt windings. If a 2.5-volt transformer is preferred, the resistors shown in the filament winding must be omitted. The 82 tube, a shield and choke were sold. The smiling report next day was, "O. K., Jack!"

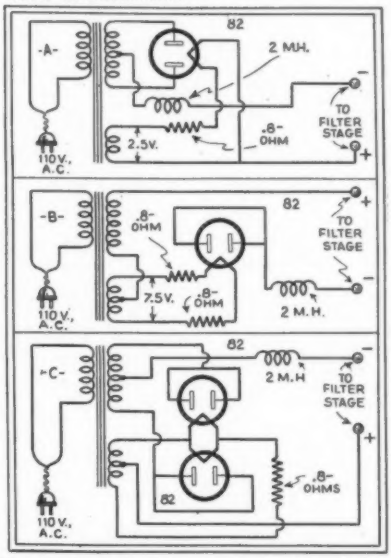

Fig. 1. At A (top) is shown the 82 replacing an '80; B (center), the connections for replacing an '81, and; C (bottom), recommended wiring for replacing two type '81 rectifiers.

Here is a good money making idea for the Service Man who owns an oscillator. Recalibrate radio receivers to read in kilocycles or meters—as most newspapers and radio magazines list the stations. Thousands of set owners, especially those fishing for distance, would like to dial for stations as listed in these periodicals.

It gets hot, a resistor does, particularly if it is placed too close to the base or other objects that will not allow it to dissipate heat, even if the power rating of the resistor is three times what it should be—correct figuring, say some of our good engineers—and others.

The flood of the best toned, longest-distance getting sets, consisting of four and five tubes "built at a price" as high as $7.95; opens the market again for wave-traps. Some salesman will admit that they are slightly broad near a powerful station. Yes sir! It appears now that all stations are powerful with this type of set.

Mr. S..... who is connected with one of the largest untilities in this country dropped in to say "hello" and at the same time to purchase a 2-watt neon tube. We all know that you can't read a newspaper by the light of a neon lamp so I asked him what he was going to use it for. He told me that an inexpensive method used to determine the efficiency of various tubes in the detector and audio systems is to use a neon lamp. The neon lamp is connected across the output of the receiver. The glow on the plate of the neon lamp will be greater when the more efficient tube is used. In some sets where the amplification is poor, the output transformer should be disconnected for this test. This ought to be good news to some of our thrifty friends.

Improving the radio-frequency stages, when your power tube already overloads, is just like adding another load to the old gray mare when she can't even pull the first load.

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EASILY BUILT 0-800 V. PEAK VOLTMETER FOR RECTIFIER WORK

By J. S. WILLIAMS*

TESTING filter circuits to determine the correct condenser to use involves a consideration of "peak voltages." To measure peak potentials requires either a vacuum tube, an electrostatic voltmeter, or an oscillograph; these are rarely found outside well-equipped laboratories. Here is a simple means of constructing a peak voltmeter which will indicate the peak voltages of the entire filter circuit of any standard radio receiver.

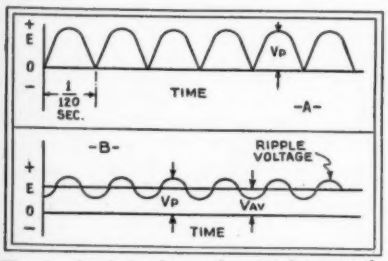

Fig. 1. At A is shown the waveform at the output of the rectifier, and at B the waveshape as it might exist on the first condenser of the filter circuit.

The output of a rectifier is pulsating direct current, as shown at A in Fig. 1. The effect of the filter condensers, imposed across the output, is to smooth out these wave pulsations. At B in Fig. 1 is shown a waveform such as might exist on the first condenser of a filter circuit.

Condensers have a maximum or peak voltage rating. This has reference to the maximum instantaneous voltage, as shown by Vp in Fig. 1A. The peak voltage of the output may be measured by a voltmeter, as shown in Fig. 2, by adding the average D.C. voltage (Vav in Fig. 1B) to the A.C. component, shown by the ripple voltage-line in the same figure.

A D'Arsonval (ordinary D.C.) meter will give the average, or D.C. working voltage. The A.C. component, the R.M.S. *ripple* voltage, is the R. M. S. value of some A.C. voltage which will give the waveshape in question (ripple voltage) when it is imposed upon a smooth D.C. voltage equal to Vav (Fig. 1B). Its frequency is 120 cycles, for a rectifier operating on a 60-cycle input.

Voltages across rectifier circuits naturally vary with the line voltages applied to the set. Line potentials reached in service are sometimes as high as 135 volts. Because of this, voltage measurements in condenser tests would be made with 135 volts applied to the set. Or, if the measurement is simply to determine what voltages the condenser must stand, lower voltages may be increased to an equivalent of 135 volts, by multiplying by the following constants: 110 V., 1.227; 115 V., 1.174; 120 V., 1.125; 125 V., 1.080; 130 V., 1.038.

The maximum voltage ratings of, for instance, Mallory-Elkon condensers are: 450 volts, D.C.; 475 volts, peak; 25 volts, R.M.S. ripple.

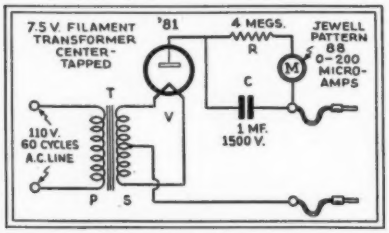

Fig. 2. The connections of the peak voltmeter. The entire voltmeter is easily mounted in a portable case, with cord, ready for service.

With 135 volts applied to the set, none of these voltages should exceed these values.

The peak voltmeter described here is sufficiently accurate for all test work of this nature. In addition, it is low in cost, easy to construct, and it is usable on either D.C. or A.C. as designed; the range is 0 to 800 volts.

Only five parts are needed, as shown, These are: One 7.5 volt center-tap, filament transformer, T; one '81 type tube, V; one 1 mf, 1500 V. condenser, C; one 4-meg. resistor, R; and a Jewel pattern 88, 0-200 micrometer, M.

(Since the voltage and current in an A.C. circuit are continually varying in value and reversing in direction, it is evident that the equivalent to an equal unit of steady direct current is really some value between the maximum and minimum values of the A.C.

In order to establish the units of alternating voltage and current on a common basis with those of D.C., an ampere in an A.C. circuit is defined as that rate of current flow in an A.C. circuit which will produce heat at exactly the same rate as one ampere of steady D.C. This is the *effective* or r. m. s. value of the A.C. and is equal to the maximum or peak value multiplied by 0.707 (for a sine-wave potential); also, the effective voltage is equal to 0.707 times the maximum voltage (for a sine wave).

The heating effect of D.C. is proportional to the square of the current (12R). If we consider the instantaneous values of current flowing at various internals during a cycle of A.C., then square them (heating effect), then find the average of these values (average heating effect) and then extract the square root of this average value, it will be equal to the D.C. that will produce an identical heating effect. The effective value of the current obtained in this manner is 0.707 times the maximum value. Since it is the "square root of the average or mean squares" of several current values during one cycle, it is abbreviated root-mean-square, or r. m. s.

This subject is very nicely covered in "Radio Physics Course," by A. A. Gherardi.—Tech. Ed.)

*Chief Engineer, Mallory-Elkon Division P. R. Mallory & Co.

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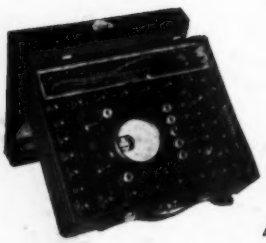

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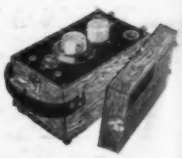

No. 60 Handy Oscillator

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